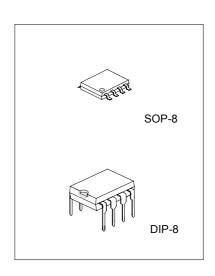
# HIGH-EFFICIENCY DC/DC **CONVERTER**

#### **DESCRIPTION**

The UTC 3563 is a monolithic control circuit containing the primary functions required for DC to DC converters and highside-sensed constant current source. The device consists of an internal temperature compensated reference, comparator, controlled duty cycle oscillator with an active current sense circuit, bootstrapped driver, and high current output switch. This device is specifically designed to construct a constant current source for battery chargers with a minimum number of external components. Bootstrapped driver can drive the NPN output switch to saturation for higher efficiency and less heat dissipation. The UTC 3563 can deliver 1.5A continuous current without requiring a heat sink.



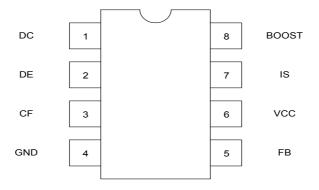
#### **FEATURES**

- \*3V to 30V input voltage operation.
- \*Internal 2A peak current switch.
- \*1.5A continuous output current.
- \*Bootstrapped driver.
- \*High side current sense capability.
- \*High efficiency (up to 90%)
- \*Internal ±2% reference.
- \*Low quiescent current at 1.6mA.
- \*Frequency operation from 100Hz to 100KHz.

#### **APPLICATIONS**

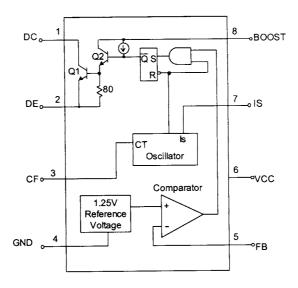
- \*Constant current source for battery chargers.
- \*Saver for cellular phones.
- \*Step-Down DC-DC converter module.

# PIN CONFIGURATION



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## **BLOCK DIAGRAM**



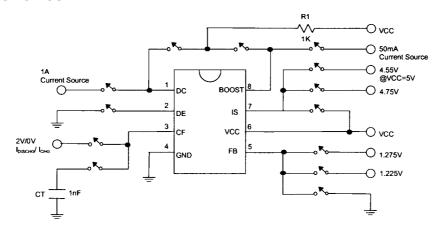
PIN	SYMBOL	DESCRIPTION	PIN	SYMBOL	DESCRIPTION		
1	DC	2A switch collector	5	FB	Feedback comparator inverting input		
2	DE	Darlington switch emitter	6	VCC	Power supply input		
3	CF	Oscillator timing capacitor	7	IS	Highside current sense input		
					(V <sub>CC</sub> -Vis=300mV)		
4	GND	Power ground	8	BOOST	Bootstrapped driver collector		

## ABSOLUTE MAXIMUM RATINGS (Ta=25°C)

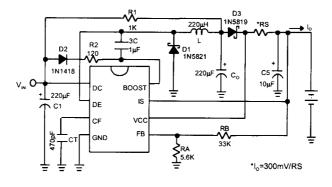
SYMBOL	VALUE	LINUT
	VALUE	UNIT
VCC	30	V
Vi	-0.3~+30	V
Vc(switch)	30	V
VE(switch)	30	V
Vce(switch)	30	V
Vc(driver)	30	V
Isw	2	Α
	1000	mW
	625	mW
	100	°C/W
	160	°C/W
Tj	125	°C
Та	0~+70	°C
Tstg	-65~+150	°C
	Vi Vc(switch) VE(switch) VcE(switch) Vc(driver) Isw	Vi         -0.3~+30           Vc(switch)         30           VE(switch)         30           VcE(switch)         30           Vc(driver)         30           Isw         2           1000         625           100         160           Tj         125           Ta         0~+70

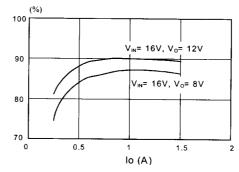
# ELECTRICAL CHARACTERISTICS (VCC=5.0V, Ta=25°C, unless otherwise specified)

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PARAMETER	SYMBOL	TEST CONDITIONS	MIN	TYP.	MAX	UNIT
Oscillator						
Charging Current	ICHG	5V≤VCC≤30V	10	25	40	μΑ
Discharging Current	Idischg	5V≤VCC≤30V	100	150	200	μΑ
Voltage Swing	Vosc	PIN 3		0.6		V
Discharge to Charge Current Ratio	IDISCHG / ICHG	Vis=VCC		6.0		
Current limit Sense Voltage	VCC -VIS	ICHG=IDISCHG	250	300	350	mV
Output Switch						
Saturation Voltage, Emitter Follower Connection	VCE(SAT)	IDE=1.0A, VBOOST=VDC=VCC		1.5	1.8	V
Saturation Voltage	VCE(SAT)	IDC=1.0A, IBOOST=50mA, (Forced β≈20)		0.4	0.7	V
DC Current Gain	hFE	ISC=1.0A VCE=5.0V	35	120		
Collector Off State Current	IC(OFF)	Vce=30V		10		nA
Comparator						
Threshold Voltage	VFB	Ta=25°C 0°C≤Ta≤70°C	1.225 1.210	1.250	1.275 1.290	V V
Threshold voltage Line Regulation	REGLINE	3V≤VCC≤30V		0.1	0.3	mV/V
Input Bias Current	lв	VIN=0V		0.4	1.0	μΑ
Supply Current	ICC	VIS=VCC, PIN 5>VFB, 5.0V≤VCC≤30V, CT=1nF, PIN 2=GND, Remaining pins open		1.6	3.0	mA



# TYPICAL APPLICATION CIRCUIT

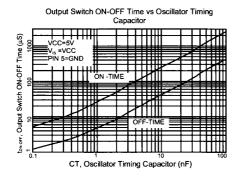


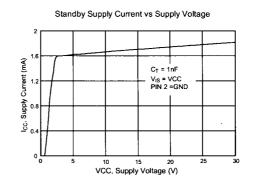


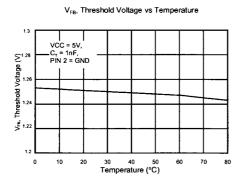
EFFICIENCY vs OUTPUT CURRENT

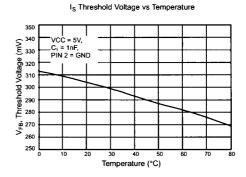
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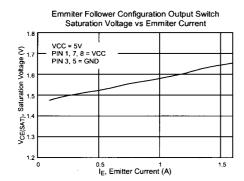
#### TYPICAL PERFORMANCE CHARACTERISTICS

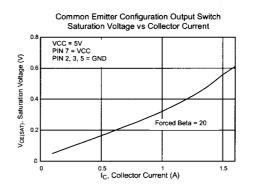












### **DESIGN FORMULA TABLE**

APPLICATION INFORMATION

CALCULATION	STEP-DOWN	STEP-UP			
t <sub>ON</sub> /t <sub>OFF</sub>	$V_{OUT}$ + $V_{F}$	$V_{OUT}+V_F-V_{IN(MIN)}$			
	V <sub>IN(MIN)</sub> -V <sub>SAT</sub> -V <sub>OUT</sub>	V <sub>IN(MIN)</sub> -V <sub>SAT</sub>			
(t <sub>ON</sub> +t <sub>OFF</sub> ) <sub>MAX</sub>	1/F <sub>MIN</sub>	1/F <sub>MIN</sub>			
Ст	4x10 <sup>-5</sup> t <sub>ON</sub>	4x10 <sup>-5</sup> t <sub>ON</sub>			
Ic(switch)	2l <sub>OUT(MAX)</sub>	$2I_{OUT(MAX)}(\frac{t_{ON} + t_{OFF}}{t_{OFF}})$			
RS	0.3/I <sub>C(SWITCH)</sub>	0.3/I <sub>C(SWITCH)</sub>			
L(MIN)	$(\frac{V_{\text{IN(MIN)}}-V_{\text{SAT}}-V_{\text{OUT}}}{I_{\text{C(SWITCH)}}}) t_{\text{ON(MAX)}}$	$\left(\begin{array}{c} V_{\text{IN(MIN)}} - V_{\text{SAT}} \\ I_{\text{C(SWITCH)}} \end{array}\right) t_{\text{ON(MAX)}}$			
Со	I <sub>C(SWITCH)</sub> (t <sub>ON</sub> +t <sub>OFF</sub> )  8V <sub>RIPPLE(P-P)</sub>	lout ton VRIPPLE(P-P)			

 $V_{\text{SAT}}$ : Saturation voltage of the output switch  $V_F$ : Forward voltage of the ringback rectifier

The following power supply characteristics must be chosen:

V<sub>IN</sub>: Norminal input voltage

 $V_{OUT}$ : Desired output voltage,  $V_{OUT}$ =1.25(1+RB/RA)

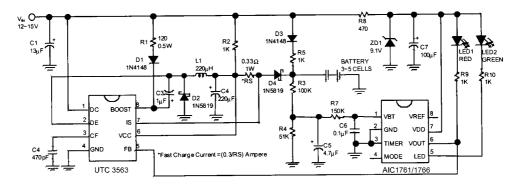
I<sub>OUT</sub> : Desired output current

 $F_{MIN}$ : Minimum desired switching frequency at selected values for  $V_{IN}$  AND  $I_{OUT}$ 

V<sub>RIPPLE(P-P)</sub>: Desired peak-to-peak output ripple voltage. In practice, the calculated value will need to be increased due to the capacitor equivalent series resistance and board layout. The ripple voltage should be kept to a low value since it will directly affect the line and load regulation.

# **APPLICATION EXAMPLES**

Fig. 1 Simplified Battery Charge Circuit for Ni-Cd/Ni-MH Battery



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Fig.2 Battery Charge Circuit for Fluctuating Charging Current Applications

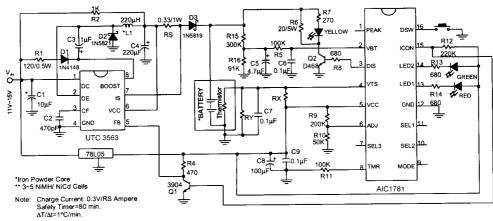
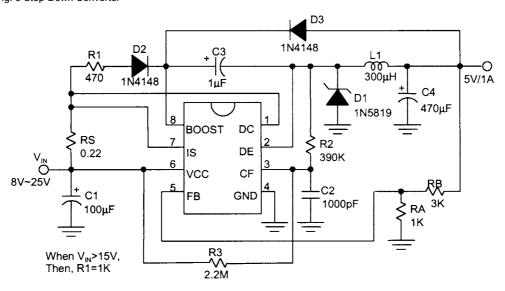


Fig. 3 Step-Down Converter



Line regulation: 40mV (V<sub>IN</sub>=10V~20V, @I<sub>O</sub>=1A) Load regulation: 20mV ( $V_{IN}$ =15V,  $@I_O$ =100mA~1A) Short circuit current: 1.3A (V<sub>IN</sub>=15V, @R<sub>L</sub>=0.1Ω)

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Fig. 4 Step-Down Converter with External 5V Bootstrap

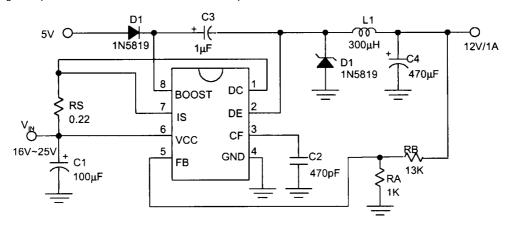
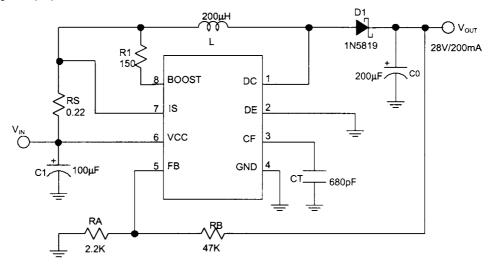


Fig. 5 Step-Up Converter



Line regulation: 100mV (V<sub>IN</sub>=8V~16V, @Io=200mA) Load regulation: 40mV ( $V_{IN}$ =12V,  $@I_{O}$ =80mA~200mA)

Fig. 6 Step-Up Converter with External NPN Switch

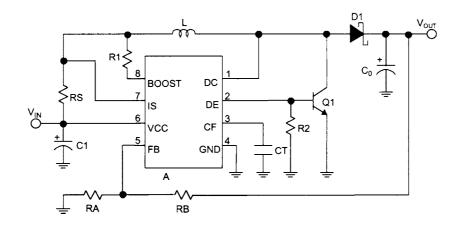
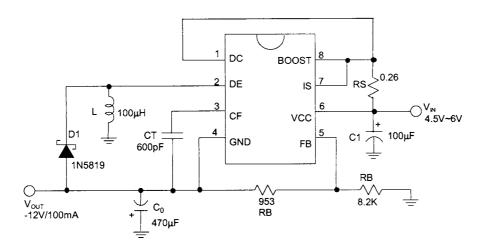


Fig. 7 Inverting Converter



Line regulation: 20mV (V<sub>IN</sub>=4.5V~6V, @I<sub>O</sub>=100mA) Load regulation: 100mV (V<sub>IN</sub>=5V, @I<sub>O</sub>=10mA~100mA)

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