



# QUAD INTEGRATED POWER SOURCING EQUIPMENT POWER MANAGER

#### **FEATURES**

- Quad-Port Power Management With Integrated Switches and Sense Resistors
- Compliant to IEEE 802.3af Standard
- Operates from a Single 48-V Input Supply
- Individual Port 15-bit A/D
- Auto, Semi-Auto and Manual Operating Modes
- Controlled Current Ramps for Reduced EMI and Charging of PD's Bulk Capacitance
- I<sup>2</sup>C Clock and Oscillator Watch Dog Timers
- Over-Temperature Protection
- DC and DC Modulated Disconnect
- Supports Legacy Detection for Non-Compliant PD
- Supports AC Disconnect
- High-Speed 400-kHz I<sup>2</sup>C Interface
- Comprehensive Power Management Software Available
- Operating Temperature Range -40°C to 125°C

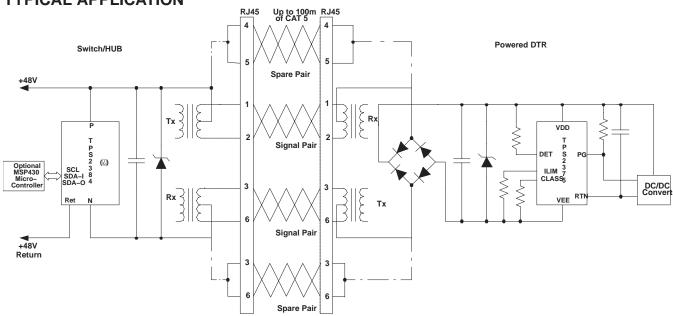
#### **APPLICATIONS**

- Ethernet Enterprise Switches
- Ethernet Hubs
- SOHO Hubs
- Ethernet Mid-Spans
- PSE Injectors

#### **DESCRIPTION**

The TPS2384 is a quad-port power sourcing equipment power manager (PSEPM) and is compliant to the Power-over-Ethernet (PoE) IEEE 802.3af standard. The TPS2384 operates from a single 48-V supply and over a wide temperature range (-40°C to +125°C). The integrated output eliminates 2 external components per port (FET and sense resistor) and will survive 100-V transient. Four individual 15-bit A/D converters are used to measure port voltage, current and die temperature making PSE solution simple and robust. The TPS2384 comes with a comprehensive software solution to meet the most demanding application which can serve as a core for all PoE system designs.

#### TYPICAL APPLICATION



Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet.



### **DESCRIPTION (CONTINUED)**

The TPS2384 has three internal supply buses (10 V, 6.3 V and 3.3 V) generated from the 48-V input supply. These supplies are used to bias all internal digital and analog circuitry. Each supply has been brought out separately for proper bypassing to insure high performance. The digital supply (3.3 V) is available for powering external loads up to 2 mA. For more demanding loads it is highly recommended to use external buffers to prevent system degradation. When the TPS2384 is initially powered up an internal Power-on-Reset (POR) circuit resets all registers and sets all ports to the off state to ensure that the device is powered up in a known safe operating state.

The TPS2384 has three modes of operation auto mode (AM), semi-auto mode (SAM) and power management mode (PMM).

- In auto mode the TPS2384 performs discovery, classification and delivery of power autonomously to a compliant PD without the need of a micro-controller.
- In semi-auto mode the TPS2384 operates in auto mode but users can access the contents of all read status
  registers and A/D registers thru the I<sup>2</sup>C serial interface. All write control registers are active except for D0
  thru D3 of port register 1 (Address 0010) for limited port control. The semi-auto mode allows the TPS2384
  to detect valid PD without micro-controller intervention but adds a flexibility to perform power management
  activities.
- Power management mode (with a micro-controller) allows users additional capabilities of discovering non-compliant (legacy) PDs, performing ac disconnect and advanced power management system control that are based on real time port voltages and currents. All functions in this mode are programmed and controlled through read/write registers over the I<sup>2</sup>C interface. This allows users complete freedom in detecting and powering devices. A comprehensive software package is available that mates the power of the TPS2384 with the MSP430 micro-controller.

TPS2384 integrated output stage provides port power and low-side control. The internal low-side circuitry is designed with internal current sensing so there are no external resistors required. The output design ensures the power switches operate in the fully enhanced mode for low power dissipation.

The I<sup>2</sup>C interface allows easy application of opto-coupler circuitry to maintain Ethernet port isolation when a ground based micro-controller is required. The TPS2384 has 5 address pins allowing individual control and monitoring of up to 128 Ethernet ports from a single master I<sup>2</sup>C micro-controller. The I<sup>2</sup>C interface is set for group broadcast on address 0 allowing simultaneous access to an entire bank of TPS2384 devices. Per-port write registers separately control each port state (discovery, classification, legacy, power up etc) while the read registers contain status information of the entire process along with parametric values of discovery, classification, and real-time port operating current, voltage and die temperature.

The proprietary 15-bit integrating A/D converter is designed to meet the harsh environment where the PSEPM resides. The converter is set for maximum rejection of power line noise allowing it to make accurate measurements of line currents during discovery, classification and power delivery for reliable power management decisions.

The TPS2384 is available in a 64-pin Power PAD™ package.



#### ORDERING INFORMATION

TEMPERATURE RANGE	PACKAGED DEVICES†
$T_A = T_J$	TQFP - 64 (PAP)‡
-40°C to 125°C	TPS2384

<sup>†</sup>The PAP package is available taped and reeled. Add R suffix to device type (e.g.TPS2384PAPR) to order quantities of 2,500devices per reel.

#### **ABSOLUTE MAXIMUM RATINGS**

over operating free-air temperature (unless otherwise noted)†‡

PARAMETER	SYMBOL	RATING	UNITS
V10 current sourced		100	μА
V3.3 current sourced		5	mA
Applied voltage on #CINT, CT, RBIAS		-0.5 to +10	
Applied voltage on SCL_I, SDA_I, SDA_O, INTB, A1, A2, A3, A4, A5, MS, PORB, WD_DIS, ALT_A/B, AC_LO, AC_HI		-0.5 to +6	V
Applied voltage on V48, #P, #N		-0.5 to +80	
Junction operating temperature	TJ	-40 to 125	
Storage temperature	T <sub>stg</sub>	-55 to 150	oC
Lead temperature (soldering, 10 sec.)	T <sub>sol</sub>	260	

<sup>†</sup> Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute–maximum–rated conditions for extended periods may affect device reliability.

#### RECOMMENDED OPERATION CONDITIONS

Parameter	MIN	TYP	MAX	UNITS
Input voltage, V <sub>DD</sub>	44	48	57	V
Junction temperature, T <sub>J</sub>	-40		125	°C

#### **ELECTRO STATIC DISCHARGE (ESD) PROTECTION**

	MAX	UNITS
Human body model	1.5	
CDM	1	kV
Machine model	0.2	]



<sup>‡</sup> All voltages are with respect to GND. Currents are positive into, negative out of the specified terminal. Consult Packaging Section of the Databook for thermal limitations and considerations of packages.

## **ELECTRICAL CHARACTERISTICS:**

V48 = 48 V, R<sub>T</sub> = 124 kΩ, C<sub>T</sub> = 220 pf, C<sub>INT</sub> = 0.027 μF (low leakage),  $-40^{\circ}$ C to  $-125^{\circ}$ C and T<sub>A</sub> = T<sub>j</sub> (unless otherwise noted)

PARAMETER	TEST CONDITION	MIN	TYP	MAX	UNITS	
Power Supply			-			
V48 quiescent current	Off mode (all ports)	4	9	12		
V48 quiescent current	Powered mode (all ports)		10	14	mA	
V10, internal analog supply	I <sub>LOAD</sub> = 0	9.75	10.5	11.5	.,	
V3.3, internal digital supply	I <sub>LOAD</sub> = 0 to 3 mA	3	3.3	3.7	V	
V3.3 short circuit current	V = 0	3		12	mA	
V6.3, internal supply	Iload = 0	5	6.3	7		
V2.5, internal reference supply	Iload = 0	2.46	2.5	2.54	V	
Input UVLO			26	32	]	
Internal POR time out(I <sup>2</sup> C)	After all supplies are good I <sup>2</sup> C activity is valid		8			
Internal POR time out (Port)	After all supplies are good  Port active to I <sup>2</sup> C commands		66000		Pulses	
Port Discovery			1	I.	1	
Port off #P to #N input resistance		400			kΩ	
Discovery open circuit voltage			22	30		
Discovery 1 voltage loop control	70 μA < I <sub>PORT</sub> < 3 mA	2.8	4.4		V	
Discovery 2 voltage loop control	70 μA < I <sub>PORT</sub> I < 3 mA		8.8	10	1	
Discovery current limit	P = N = 48 V	3	4	5	mA	
Auto-mode discovery resistance acceptance Band		19		26.5		
Auto-mode discovery resistance low end rejection		0		15	kΩ	
Auto-mode discovery resistance high end rejection		33				
Discovery1,2 A/D conversion scale factor	10 μA < I <sub>PORT</sub> < 3 mA	5.30	6.10	6.75	count/ μΑ	
Discovery1,2 A/D conversion time	I <sub>PORT</sub> = 120 μA		18	22	ms	



## **ELECTRICAL CHARACTERISTICS:**

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PARAMETER	TEST CONDITION	MIN	TYP	MAX	UNITS
Port Classification		-			•
Classification voltage loop control	100 μA < I <sub>PORT</sub> < 50 mA	15	17.5	20	V
Classification current limit	P = N = 48 V	50	60	100	
Class 0 to 1 detection threshold		5.5	6.5	7.5	1
Class 1 to 2 detection threshold		13	14.5	16	1.
Class 2 to 3 detection threshold		21	23	25	mA
Class 3 to 4 detection threshold		31	33	35	
Class 4 to 0 detection threshold		45	48	51	
Classification A/D conversion scale factor		375	424	475	Count/ mA
Classification A/D conversion time	I <sub>PORT</sub> = 50 mA		18	22	ms
Port Legacy Detection		•			
Legacy current limit	P = N = 48 V	2.6	3.5	4.3	mA
Legacy voltage A/D conversion scale factor	100 mV < V <sub>PORT</sub> < 17.5 V	1365	1400	1445	Count/V
Legacy A/D conversion time	0 V < V <sub>PORT</sub> < 15 V		18	22	ms
Port Powered Mode					
Port on resistance	20 mA < IPORT < 300 mA		1.3		Ω
Over current threshold	R <sub>BIAS</sub> = 124 kΩ, C <sub>T</sub> = 200 pF Time > 75 ms, $-25 \le T_i \le 105$	350	375	400	
Output current limit	R <sub>BIAS</sub> = 124 kΩ, C <sub>T</sub> = 200 pF Time > 75 ms, $-25 \le T_i \le 105$		425	450	mA
Average min current trip	R <sub>BIAS</sub> = 124 k $\Omega$ , C <sub>T</sub> = 200 pF Current pulse > 100 ms		7.5	10	
Port output UV		42.0	42.7	44.0	
Port output OV		54	55	56	V
Over current time Limit	R <sub>BIAS</sub> = 124 kΩ, $C_T$ = 200 pF	50		75	
Short circuit time Limit	R <sub>BIAS</sub> = 124 kΩ, $C_T$ = 200 pF	50		75	]
Turn-off delay from UV/OV faults	R <sub>BIAS</sub> = 124 kΩ, $C_T$ = 200 pF After port enabled and ramped up		3		ms
Port current A/D conversion scale factor	20 mA < IPORT I < 300 mA	31	36.41	40	Count/ mA
Port current A/D conversion time	I <sub>PORT</sub> < 300mA		18	22	ms
Port voltage A/D conversion scale factor	45 V < V <sub>PORT</sub> < 56 V	335	353	370	Count/V
Port voltage A/D conversion time	45 V < V <sub>PORT</sub> < 56 V		18	22	ms
Port temperature A/D conversion scale factor			300		Count/ °C



### **ELECTRICAL CHARACTERISTICS:**

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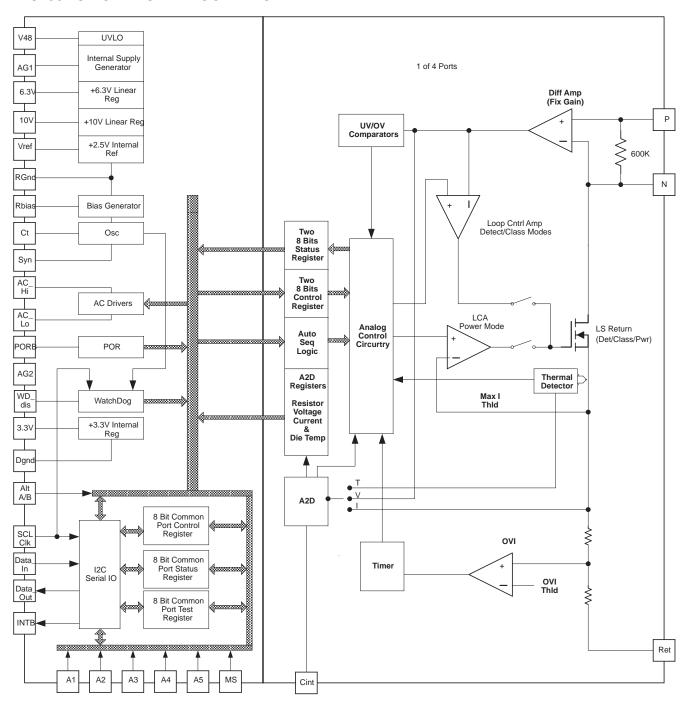
PARAMETER	TEST CONDITION	MIN	TYP	MAX	UNITS
Port Disable Mode		•			1
Port N voltage	P = 48 V	47			V
AC Low & AC High Specification		•			
AC_lo, AC_hi – low output voltage		0		0.5	
AC_lo – high output voltage		3.0		5.0	V
AC_hi – high output voltage		5.0		7.0	
Digital I <sup>2</sup> C DC Specifications		•			•
SCL, SDA_I, A1-A5 ,WD_DIS, ALTA/B, MS, PORB logic input threshold			1.5		V
SCL, SDA_I input hysteresis			250		
MS, PORB input hysteresis			150		mV
WD_DIS,ALT A/B, MS, PORB input pull-down resistance	Input voltage 0.5 to 3 V		50		kΩ
A1-A5 pull-down current			10		μΑ
SDA_O logic high leakage	Drain = 5 V		100		nA
SDA_O logic low	I <sub>SINK</sub> = 10 mA		200		mV
INTB logic high leakage	Drain = 6 V		10		μΑ
INTB logic low	ISINK = 10 mA		200		mV
Digital I <sup>2</sup> C Timing		·			•
SCL clock frequency		0		400	kHz
Pulse duration	SCL high	0.6			
Pulse duration	SCL low	1.3			
Rise time, SCL toSDA				300	
Fall time, SCL to SDA				300	
Setup time, SDA to SCL		250			]
Hold time, SCL to SDA		300		900	μs
Bus free time between start and stop		1.3			
Setup time, SCL to Start condition		0.6			
Hold time, start condition to SCL		0.6			
Setup time, SCL to stop condition		0.6			

NOTES: 1. Probe test only. TPx numbers are internal device test points only available at wafer probe

2. Ensured by design. Not 100% tested in production.



### **TPS2384 SINGLE PORT BLOCK DIAGRAM**



## **TERMINAL FUNCTIONS**

### **Power and Ground**

Pin Name	Pin #	I/O	Description
V48	60	I	48-V input to the device. This supply can have a range of 44 to 57 V. This pin should be de-coupled with a 0.1-μF capacitor from V48 to AG1 placed as close to the device as possible.
V10	58	0	10-V analog supply. The 10-V reference is generated internally and connects to the main internal analog power bus. A 0.1-µF de-coupling capacitor should terminate as close to this node and the AG1 pin as possible. Do not use for an external supply.
V6.3	59	0	6.3-V analog supply. A 0.1-μF de-coupling capacitor should terminate as close to this pin and the AG1 pin as possible. Do not use for an external supply.
V3.3	24	0	$3.3\text{-V}$ logic supply. The $3.3\text{-V}$ supply is generated internally and connects to the internal logic power bus. A $0.1\text{-}\mu\text{F}$ de-coupling capacitor should terminate as close to this node and the DG pin as possible. This output can be used as a low current supply to external logic.
V2.5	54	0	2.5-V reference supply. The V2.5 is generated internally and connects to the internal reference power bus. This pin should not be tied to any external supplies. A 0.1-μF de-coupling capacitor should terminate as close to this node and the RG pin as possible. Do not use for an external supply.
AG1	57	GND	Analog ground 1. This is the analog ground of the V6.3, V10 and V48 power systems. It should be externally tied to the common copper 48-V return plane. This pin should carry the low side of three de-coupling capacitors tied to V48, V10 and V6.3.
AG2	61	GND	Analog ground 2. This is the analog ground which ties to the substrate and ESD structures of the device. It should be externally tied to the common copper 48-V return plane. AG1 and AG2 must be tied together directly for the best noise immunity.
DG	23	GND	Digital ground. This pin connects to the internal logic ground bus. It should be externally tied to the common copper 48-V return plane.
RG	56	GND	Reference ground. This is a precision sense of the external ground plane. The integration capacitor (CINT#) and the biasing resistor (RBIAS pin) should be tied to this ground. This ground should also be used to form a printed wiring board ground guard ring around the active node of the integration capacitor (CINT#). It should tie to common copper 48-V return plane.

## **Port Analog Signal**

Pin Name	Pin#	I/O	Description			
1P	7	I				
2P	10	1	Port Positive. 48-V load sense pin. Terminal voltage is monitored and controlled differentially with re-			
3P	39	1	<ul> <li>spect to each Port N pin. Optionally, if the application warrants, this high side path can be protected with the use of a self-resetting poly fuse.</li> </ul>			
4P	42	Ι				
1N	6	I				
2N	11	Ι	Port negative. 48-V load return pin. The low side of the load is switched and protected by internal circuit-			
3N	38	Ι	ry that will limit the current.			
4N	43	Ι				
1Ret	5	Ι				
2Ret	12	Ι	40 Markon etc			
3Ret	37	I	48-V return pin			
4Ret	44	I				
1Cint	4	Ι				
2Cint	13	I	Integration capacitor. This capacitor is used for the ramp A/D converter signal integration. Connect a 0.027-μF capacitor from this pin to #RG. To minimize errors use a polycarbonate, polypropylene, poly-			
3Cint	36	I	styrene or teflon capacitor type to prevent leakage. Other types of capacitors can be used with in-			
4Cint	45	I	creased conversion error.			



## **TERMINAL FUNCTIONS**

## **Analog Signals**

Pin Name	Pin#	I/O	Description
СТ	53	I	This is a dual-purpose pin. When tied to an external capacitor this pin sets the internal clock. When the CT pin is grounded the SYN Pin turns from a output to an input (see SYN Pin description)  The timing capacitor and the resistor on the RBIAS pin sets the internal clock frequency of the device. This internal clock is used for the internal state machine, integrating A/D counters, POR time out, faults and delay timers of each port. Using a 220-pF capacitor for CT and a 124-kΩ resistor for RBIAS sets the internal clock to 245 kHz and ensure IEEE 802.3af compliance along with maximizing the rejection of 60-Hz line frequency noise from A/D measurements.
Rbias	55	_	Bias set resistor. This resistor sets all precision bias currents within the chip. The RBIAS pin is control to voltage of 1.25 V. This voltage and RBIAS generates a current which is replicated and used throughout the chip. This resistor also works in conjunction with the capacitors on CT and CINT to set internal timing values. The RBIAS resistor should be connected RG. RBIAS is a high impedance input and care needs to be taken to avoid signal injection from the SYN pin or I <sup>2</sup> C signals.
SYN	52	I/O	This is a dual purpose pin. When the CT pin is connected to a timing capacitor this output pin is a 0 to 5 V pulse of the internal clock which can be used to drive other TPS2384 SYN pins for elimination of a timing capacitor. When the CT pin is grounded this pin becomes an input pin that can be driven from a master TPS2384 or any other clock generate signal.
AC_LO	51	0	Open drain output pin for ac disconnect excitation
AC_HI	50	0	Totem poll output pin for ac disconnect excitation.

## **Digital Signals**

Pin Name	Pin #	I/O	Description
SCL	25	I	Serial clock input pin for the I <sup>2</sup> C interface.
SDA_I	26	I	Serial data input pin for the I <sup>2</sup> C interface. When tied to the SDA_O pin, this connection becomes the standard bi-directional serial data line (SDA)
SDA_O	27	0	Serial data open drain output for the I <sup>2</sup> C interface. When tied to the SDA_I pin, this connection becomes the standard bi-directional serial data line (SDA). This is a open drain output that can drive opto-coupler LEDs directly from the 48-V bus with an external, series current limiting resistor.
WD_DIS	22	I	This input pin when held high disables the watch dog timer monitoring the I $^2$ C clock and the OSC/SYN pins. The WD pin has an internal 50-k $\Omega$ resistor pull-down to digital ground. Also the WD timer is automatically shut off when the MS pin is set to 0 V (auto mode).
INTB	20	0	This is an open-drain output that goes low if a fault condition occurs on any of the 4 ports.
ALT A/B	21	I	When this input is set to logic low there is no back-off time after a discovery failure. When this pin set to a logic high there is a back-off time (approximately 2 seconds) before initiating another discovery cycle. This pin has an internal $50-k\Omega$ resistor pull-down to digital ground.
A1	28	1	
A2	29	I	Address 1 through 5. This is the I <sup>2</sup> C address select input. Select the appropriate binary address on these
А3	30	I	pins by connecting this pin to chip ground for a logic low and tying this pin to the V3.3 pin for a logic high.
A4	31	1	Each address line has an internal current source pull-down to digital ground.
A5	32	1	
MS	63	I	The MS pin selects either the auto mode (MS Low) or the power management mode, PMM, (MS High). This pin can be held low for controller-less standalone applications. When MS is Low and the POR timing cycle is complete the chip will sequentially <i>Discover, Classify</i> and <i>Power on</i> each port. When MS is set high the ports are controlled by register setting via the $I^2C$ bus. The MS pin has an internal $50$ -k $\Omega$ resistor pull-down to analog ground.
PORB		1	This pin can be used to override the internal POR. When held low, the $I^2C$ interface, all the state machines, and registers are held in reset. When all internal and external supplies are within specification, and this pin is set to a logic high level, the POR delay will begin. The $I^2C$ interface and registers will become active within 70 $\mu$ s of this event and communications to read or preset registers can begin. The reset delay for the remainder of the chip then extinguishes in 1 second. This pin has an internal 50-k $\Omega$ resistor pull-down to analog ground.



## **CONNECTION DIAGRAM**

#### (TOP VIEW) AC\_IO PORB AC\_HI **RBIAS** AG2 V2.5 SYN AG1 RG СТ 62 60 59 51 61 58 57 55 54 53 52 50 49 63 56 48 2 47 3 46 CINT1 45 CINT4 RET1 44 RET4 5 6 43 N1 7 **P**1 42 Р4 **TPS2384** 64 Pin Power Pad 8 41 **TQFP** 9 40 10 39 P2 Р3 11 38 N2 N3 RET2 12 37 RET3 36 CINT3 CINT2 13 14 35 15 34 16 33 20 8 L N 26 FYQS 17 18 19 21 22 23 24 25 27 28 29 30 31 32 WD\_DIS SDA-0 AltA/B **A**5 Ā A4 **A**2

tqfp PACKAGE (TQFP-64)



#### **Auto Mode**

Auto mode (MS = low) is a basic approach to building power sourcing equipment power manager systems without a micro-controller. When AM has been selected the TPS2384 automatically performs the following functions;

- Discovery of IEEE 802.3af compliant power devices (PD)
- Classification
- Power delivery
- Port over/under voltage detection
- Port over current detection (350 mA < I<sub>PORT</sub> < 400 mA)</li>
- Port maximum current limit (400 mA < I<sub>PORT</sub> < 450 mA)</li>
- DC disconnect (5 mA < I<sub>PORT</sub>< 10 mA)</li>
- Thermal shutdown protection (Tj > 150° C)
- Internal oscillator watch dog

In AM all the information in the read registers are available through the I<sup>2</sup>C interface along with controlling some states by writing to the appropriate registers. This gives the user the ability to use the TPS2384 in a semi-auto mode where the TPS2384 auto detects compliant PD's and a host can access the A/D registers and class information for power management, turn-on or off ports and disable dc disconnect or discovery faults.

The write registers that still active in AM are;

- All ports disable common write register 0001
- Over/Under Voltage Faults common write register 0001
- Software reset common write register 0001
- Disconnect disable write register 0001
- Discovery fault disable write register 0001
- Port enable write register 0011
- Power mode class over current threshold enable write register 0011



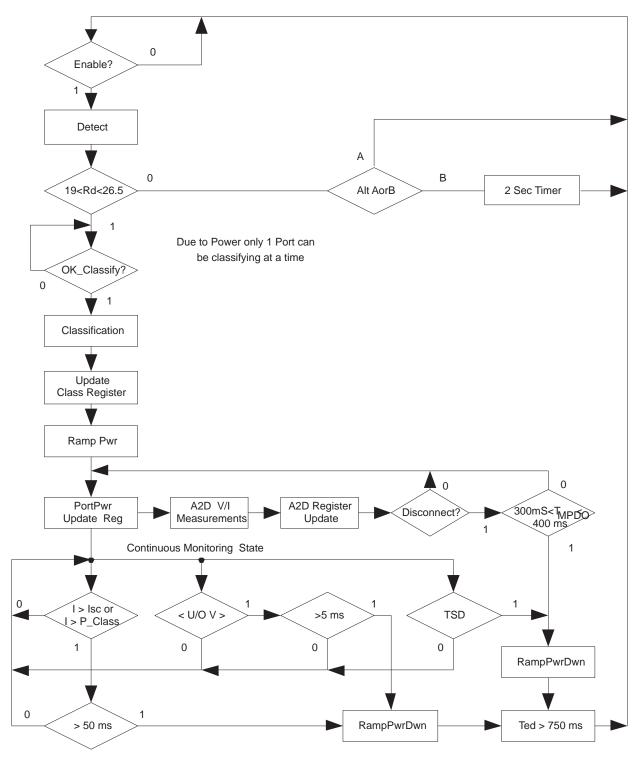


Figure 1. The Basic Flow for Auto Mode



#### **AM Discovery**

The TPS2384 uses two low level probe signals (typically 4.4 V and 8.8 V) during the discovery process to determine whether a valid PD is present. The use of multi point or slope detection method for the PD resistor measurement allows accurate detection even when series steering diodes are present. The low level probe voltages also prevent damage to non-802.3 devices. When a valid PD has been detected the TPS2384 moves to the classification. If a valid PD has not been detected the TPS2384 continues to cycle through the discovery process. The waveform in Figure 2 shows typical N-pin waveforms for the discovery of a valid PD and the failure to discovery due to a discovery resistor of 15 k $\Omega$  and 33 k $\Omega$ .

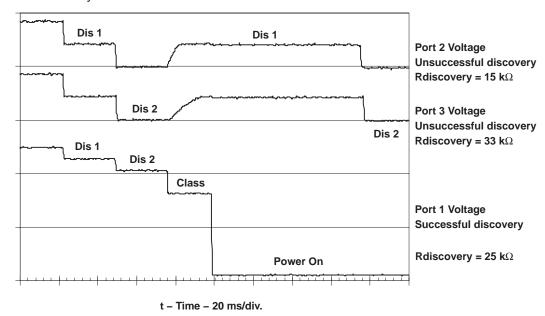


Figure 2.



#### **AM Classification**

After a successful discovery of a valid PD the TPS2384 enters the classification function that identifies the power level based on the PD's current signature. The classification current level is measured at a reduced terminal voltage of 17.5 V and classified with 15 bits of resolution. During classification the power consumption can be at its highest so to prevent over temperature shutdown in auto mode only one port classifies at a time. When multi-ports successfully discovery and enter classification at the same time the auto sequencer processes each request separately allowing only one port to enter classification. Figure 3 shows all 4 ports successfully detecting a valid PD at the same time and than the classification of each port occurring separately.

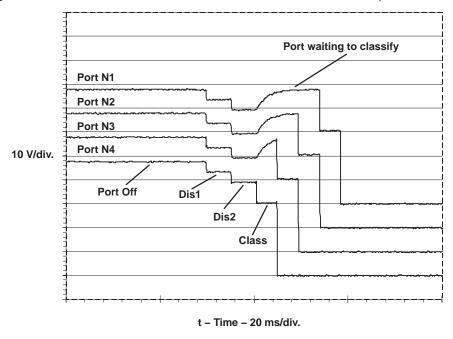


Figure 3.



Upon completion of classification the port classification register is updated. In AM mode this information is not used but for semi-auto mode the class information can be used for power management and the over current threshold can be set to correspond to the PD's classification by setting the power mode class over current threshold bit in the individual port control register (address 0011). The Figure 4 shows actual class currents and the class assignment which where stored in the register. These assignments are compliant with the IEEE 802.3af Standard

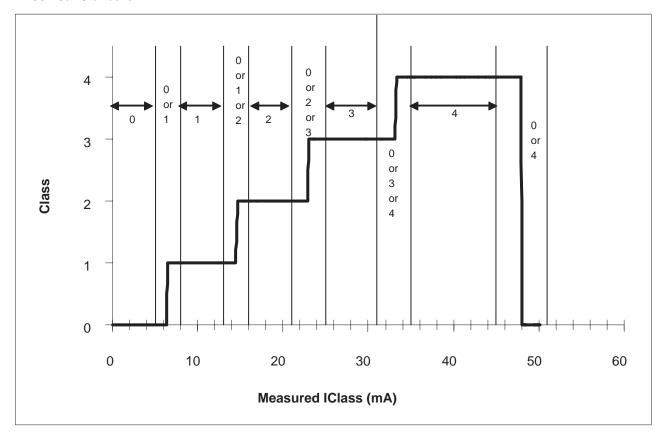
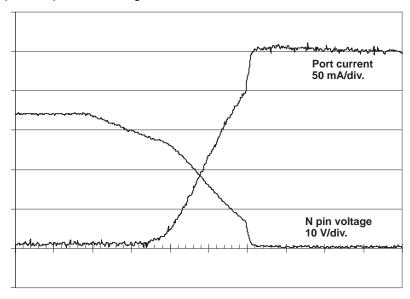


Figure 4.



#### **AM Power Delivery**

After successfully discovery and classification of a valid PD the power is delivered by controlling the current to the PD until its current requirements are met or until the internal current limit is reached (approximately 425 mA). The power switch is fully enhanced after  $500 \, \mu s$ . Figure 5 show the voltage and the current that is being applied to the PD during power-up and reaching the PD load of 250 mA.



t – Time – 100  $\mu$ s/div.

Figure 5.

After power has been applied to the PD the TPS2384 automatically enters the current and voltage sample mode. The sample mode performs 31 current measurements and a 1 voltage measurement. Each measurement takes approximately 18 ms to complete. The port remains powered and the current/voltage measurement cycle continues until a fault condition occurs. The current and voltage measurements are both stored in the A/D current and voltage registers and can be access through the I<sup>2</sup>C pins. This allows power management in the AM if it is desired.



#### **AM Faults**

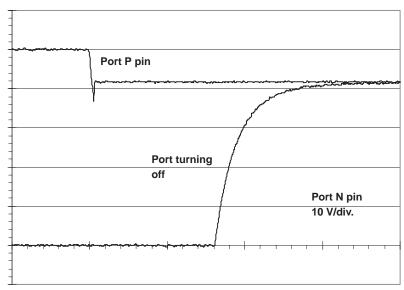
AM faults are;

- Under and over voltage faults
- Over current faults
- Under current (dc modulated) fault
- Thermal shutdown fault
- Watch dog oscillator fault

Any one of these faults will cause the port to shutdown.

## **Over/Under Voltage Fault**

Over/under voltage faults are only processed after power up has completed. The TPS2384 measures the voltage between the P and N pin and if this voltage drops below the under voltage threshold (typical 43 V) or the over voltage threshold (53 V) the voltage timer is turned on. When the voltage timer reaches its time out limit that is set between 2 ms to 5 ms the corresponding port will be turned off and the UV/OV fault bit is set in the port status register. If the over/under voltage condition goes away prior to the voltage timer reaching its limit the timer is reset and waits for the next event. Figure 6 shows a voltage fault lasting for more then 2 ms that has caused the port to shutdown.



t - Time - 2 ms/div.

Figure 6.



#### **Over Current or Current Limit Faults**

Over current or current limit faults are conditions when the load current that is being sensed trips either the over current comparator (350 mA to 400 mA) or the max current comparator (400 mA to 450 mA) and turns on the over current timer. When the over current timer reaches its time out limit that is set between 50 ms to 80 ms the corresponding port is turned off and the over current bit is set in the port status register. If the over current condition goes away prior to the over current timer reaching its limit the timer is reset and waits for the next event. Figure 7 shows an over current fault lasting more than 50 ms that has caused the port to shut off.

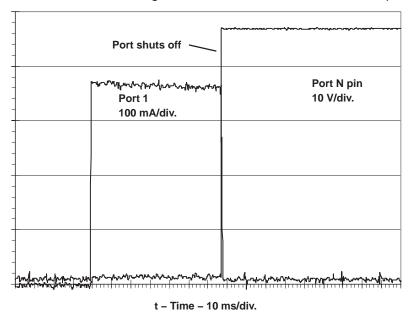


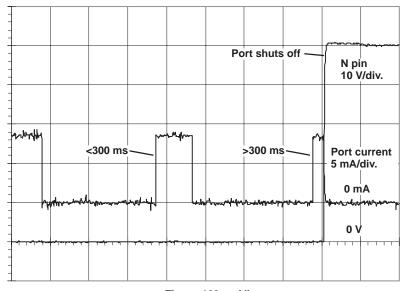
Figure 7.

## **Under Current Fault (DC Modulated Disconnect)**

Under current fault (dc modulated disconnect) is a condition when the load current that is being measured drops below 7.5 mA and turns on the disconnect timer. If the disconnect timer reaches its time out limit that is set between 300 ms to 400 ms the corresponding port is turned off and the load disconnect bit is set in the port status register. If the under current condition goes away prior to the disconnect timer reaching its limit the timer is reset and the port remains powered.



Figure 8 shows dc disconnect event. In this setup the load current was set right above the 7.5 mA threshold. The duty cycle of the load was then adjusted until the point within the disconnect window of 300 ms to 400 ms.



t - Time - 100 ms/div.

Figure 8.



### **Power Management Mode (PMM)**

Power management mode (PMM) has been designed to work efficiently with simple low-cost micro controllers such as those in the MPS430 family.

The power management mode has 13 self contain functions that are needed to perform all your power management. You simply write/read through the I<sup>2</sup>C pins and wait for the function done bit to be set. If an A/D measurement was performed during the function the results can be access by going to the read mode and addressing the proper register.

#### 13 Functions

- Disable: Disables the port.
- Discovery 1: Enables the discovery 1 condition which applies a 4.4 V across the PD and measure current and store the resulting current.
- **Discovery 2:** Enables the discovery 2 condition which applies a 8.8 V across the PD and measure and store the resulting current.
- V Sample: Measure the voltage between the P and N pins and store the result in the A/D voltage register.
- Legacy: Enables the 3.5-mA current source for measuring capacitance and measure the voltage across the P and N terminals and store the result in the A/D voltage register.
- Classify: Enables the classification condition which applies 17.7 V across the PD and measure and store
  the resulting current.
- Rup Pwr: Turn on the output switch while controlling the current beginning delivery to the PD until the PD current needs are met or the max current is reached.
- **C Sample:** Continues sample mode cycle performing 31 current measurements and 1 voltage measurement. After each measurement the contents for the appropriate register is updated.
- Rdwn: Turn off the output switch while controlling current until output current reaches 0 mA.
- AC Low: Turn on ac low driver circuit.
- AC High: Turn on ac high driver circuit.
- **ISample:** Measure the current and store the result in the A/D current register.
- TSample: Measure the internal die temperature and store the result in the A/D temperature register.



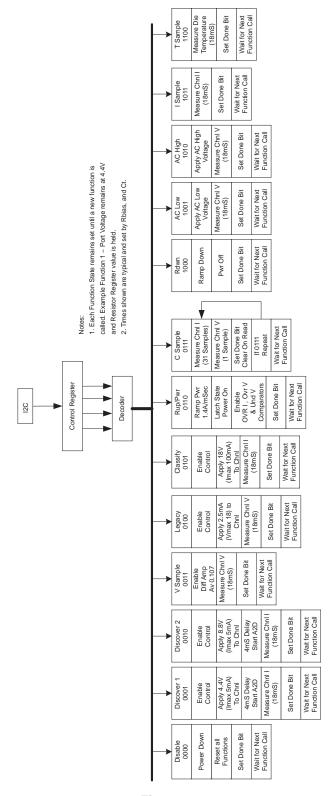


Figure 9.



### PMM Discovery 1

PMM discovery 1 function waveforms for the N and CINT pins are shown in Figure 10. The measurement is being performed using  $25~\mathrm{k}\Omega$  impedance between the P and N pin. The discovery 1 voltage is allowed to settle for approximately 5 ms before the A/D begins integrating. The voltage on the CINT pin shows the A/D cycle. There are 4 distinct regions to any A/D cycle, pre-charge to know starting voltage, charge time, coarse discharge time and fine discharge time. CINT pin is very high impedance therefore extreme care must be taken to avoid any noise or leakage affecting this pin. For the measurements where CINT voltage is shown a buffer was used to prevent performance degradation. The A/D measurement time is approximately 18 ms. The entire discovery 1 function takes approximately 22 ms to complete. The function remains set and the data in the register remains until another function is called. At the end of the fine discharge ramp the data is stored in the resistor register and the done bit is set.

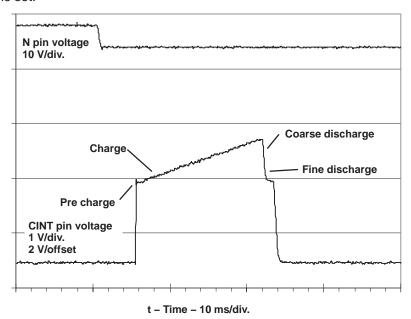


Figure 10.



## **PMM Discovery 2**

PMM discovery 2 function waveforms for the N and CINT pins are shown in Figure 11. Again the measurement is being performed using 25 k $\Omega$  impedance between the P and N pin. The discovery 2 function was call after a discovery 1 function so the voltage ramps from 4.4 V to 8.8 V below the P pin. The discovery 2 voltage is given 5 ms to settle before the A/D begins to integrate. At the end of the fine discharge ramp the data is stored in the resistor register and the done bit is set.

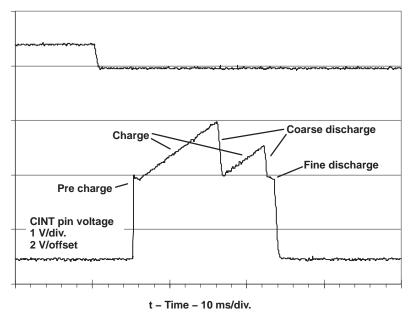


Figure 11.



#### **PMM Classification**

PMM classification function looks similar to discovery 1 and 2 except that the voltage between the P and N pins regulates to approximately 17.5 V.

### **PMM Legacy**

PMM legacy function is used to detect PDs that are non compliant. Legacy detection uses a current source (typically 3.5 mA) as a test current while the A/D measures the average voltage for approximately 18 ms. The waveform shown in Figure 12 is the legacy function charging a  $10-\mu F$  capacitor. The capacitance charges to a value that is no greater than 20 V below the P port voltage.

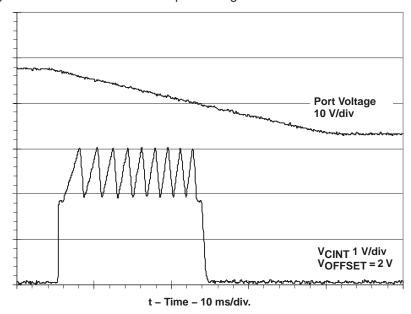


Figure 12.



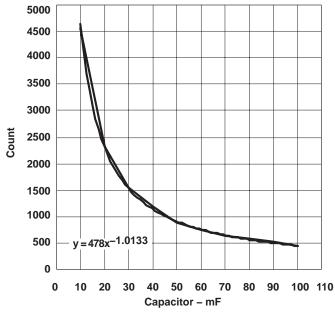
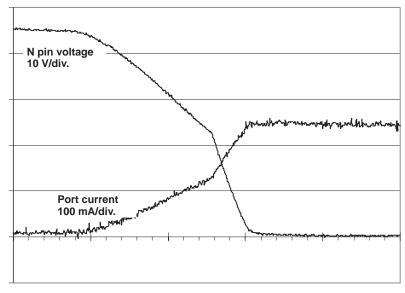


Figure 13.

## **PMM** Rup Pwr

PMM R<sub>UP</sub> power function turns on the port power by ramping up the current that is being delivered to the load in a controlled fashion. The output current ramps from 0 mA to  $I_{MAX}$  (typically 425 mA) in approximately 500  $\mu$ s. Figure 14 shows the output voltage and current turning on for a 250-mA load.



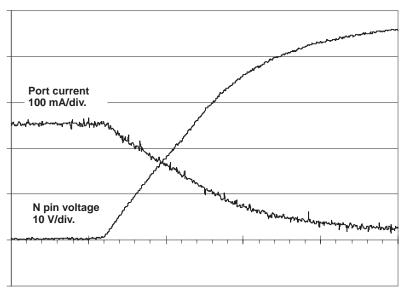
t – Time – 100  $\mu$ s/div.

Figure 14.



## PMM R<sub>DWN</sub>

PMM R<sub>DWN</sub> function turns off the port power by ramping down the current in a controlled fashion. The output current ramps from  $I_{MAX}$  (typically 425 mA) to 0 mA in approximately 500  $\mu$ s. Figure 15 shows the output voltage and current shutting down for a 250-mA load.



t – Time – 100  $\mu\text{s/div.}$ 

Figure 15.



#### **Watch Dog Timer**

TPS2384 has two watch dog timers. One monitors the  $I^2C$  clock and the other monitors the internal clock. When auto mode is selected and the watch dog timer has not been disable only the internal clock is monitored. When in power management mode and the watch dog timer has not been disable than both the  $I^2C$  and internal clocks are monitored. If there is no  $I^2C$  clock activity for approximately 2 seconds than all ports are disable. There are three means to enable ports after a  $I^2C$  clock fault and they are:

- 1. Hard power reset
- 2. PORB pulse
- 3. Writing a software reset to the common control register.

In both auto mode and power management mode if the internal oscillator is lost for more than 20 ms all ports are disabled.

Loss of these signals is considered catastrophe since the system loses its ability talk to each port. Therefore the WD timers disabling all ports protects the system from potential catastrophe failures.

This function can be easily over ridden by setting the WD\_DIS pin high.



## I<sup>2</sup>C Interface Description

The serial interface used in the TPS2384 is a standard 2-wire I<sup>2</sup>C slave architecture. The standard bi-directional SDA lines of the I<sup>2</sup>C architecture are broken out into independent input and output data paths. This feature simplifies earth grounded controller applications that require opto-isolators to keep the 48-V return of the Ethernet power system floating. For applications where opto-isolation is not required, the bi-directional property of the SDA line can be restored by connecting SDA\_I to SDA\_O. The SCL line is a unidirectional input only line as the TPS2384 is always accessed as a slave device and it never masters the bus.

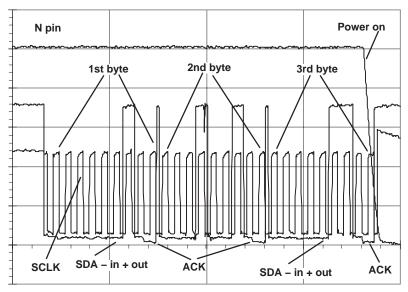
Data transfers that require a data-flow reversal on the SDA line are 4-byte operations. This occurs during a TPS2384 port read cycle where a slave address byte is sent, followed by a port/register address byte write. A second slave address byte is sent followed by the data byte read using the port/register setup from the second byte in the sequence.

The I<sup>2</sup>C read cycle consists of the following steps 1 through 14 and is shown in Figure 20:

- 1. Start Sequence (S)
- Device address field
- 3. Write
- 4. Acknowledge
- 5. Register/Port address
- Acknowledge
- 7. Stop
- 8. Start
- 9. Device address field
- 10. Read
- 11. Aknowledge
- 12. Data Transfer
- 13. Acknowledge
- 14. Stop



Data write transfers to the TPS2384 do not require a data-flow reversal and as such only a 3-byte operation is required. The sequence in this case would be to send a slave device address byte, followed by a write of the port/register address followed by a write of the data byte for the addressed port. Figure 16 is an actual 3-byte operation showing a function call to ramp on port 3.



t – Time – 500  $\mu$ s/div.

Figure 16.

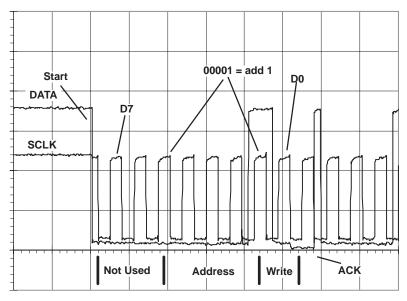
The I<sup>2</sup>C write cycle consists of the following steps 1 through 9 and is also shown in Figure 20:

- 1. Start Sequence (S)
- 2. Device address field
- 3. Write
- 4. Acknowledge
- 5. Register/Port address
- 6. Acknowledge
- Data for TPS2384
- 8. Acknowledge
- 9. Stop



### First Write Byte Magnified (Address Selection Byte, Table 1)

In Figure 17 the data line (SDA) high-to-low transition while the clock (SCL) is high generates a start condition. In this example, a device address of 00 0000 1 is then transmitted. Data low during the eighth clock pulse signals that this is a write operation. The SDA\_O line goes low on the falling edge of the 8th clock pulse, as the device acknowledges its address.



t – Time – 100  $\mu\text{s/div.}$ 

Figure 17.

## Second Write Byte Magnified (Port/Register Selection, Table 2)

Figure 18 shows the expansion of the second byte of operation. After the Port/Register acknowledge bit, the host transmits the data 0 0010 0 10 selecting write control register 1 for port 3. If a common register is selected than the port information is not used. This sets up the path for the 3rd byte during the write cycle which will be directed to call a function for control of port 3.

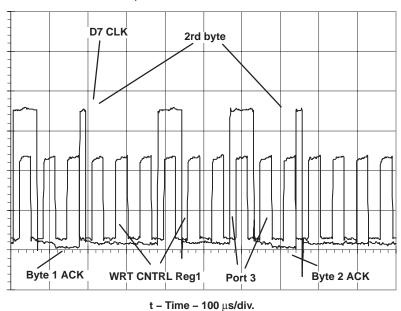


Figure 18.



## Third Write Byte (Port Control Byte, Any Write Control Register)

Figure 20 shows an expansion of the third byte of the operation. After the Port/Register acknowledge bit, the host transmits the data 00 00 0110, which is the TPS2384 Ramp-Up/Power Function. The TPS2384 acknowledges the data byte by pulling the SDA line low for one clock cycle. After the acknowledge pulse, the host generates a Stop condition by releasing the data line while the clock line is high.

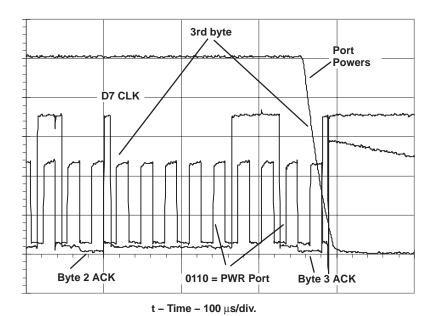


Figure 19.

The I<sup>2</sup>C interface and the port read write registers are held in active reset until input voltage is within specification and the internal POR timer has timed out.

#### Start/Stop

The high-to-low transition of SDA while SCL is high defines the start condition. The low to high transition of SDA while SCL is high defines the stop condition. The master device initiates all start and stop conditions.

A first serial packet enclosed within start and stop bits, consists of a 7-bit address field, read/write bit, and the acknowledge bit. The acknowledge bit is always generated by the device receiving the address or data field. Five of the seven address bits are used by the TPS2384. The sixth and seventh bits are placeholders for future expansion. During a write operation to the TPS2384 from the master device, the data field is 8 bits long. During a read operation where the TPS2384 is writing to the master device, the data field is also 8 bits long.



### **Chip Address**

The address field of the TPS2384 is 8 bits long and contains 5 bits of device address select and a read/write bit as and two spare bits per Table 1. The leading two bits are not used and are reserved for future port expansion. The five device address select bits follow this plan. These bits are compared against the hard-wired state of the corresponding device address select pins (A1–A5). When the field contents are equivalent to the pin logic states, the device is addressed. These bits are followed by LSB bit, which is used to set the read or write condition (1 for read and 0 for write). Following a start condition and an address field, the TPS2384 responds with an acknowledge by pulling the SDA line low during the 9th clock cycle if the address field is equivalent to the value programmed by the pins. The SDA line remains a stable low while the 9<sup>th</sup> clock pulse is high.

#### START/STOP SEQUENCE

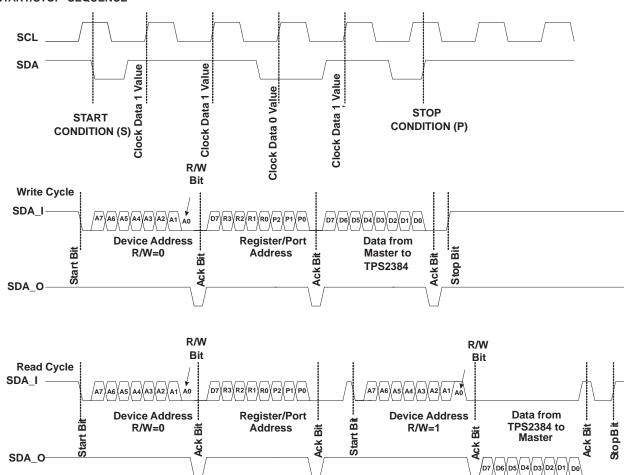


Figure 20. I<sup>2</sup>C Read/Write Cycles

#### **Chip Addressing**

Table 1 shows the bit assignments during the addressing cycle.

**Table 1. Address Selection Field** 

BIT	FUNCTION
A7	Future expansion (not used)
A6	Future expansion (not used)
A5	Device address. Compared with pin A5
A4	Device address. Compared with pin A4
А3	Device address. Compared with pin A3
A2	Device address. Compared with pin A2
A1	Device address LSB. Compared with pin A1
A0	Read/Write

#### Port/Register Cycle

After the chip address cycle, the TPS2384 accepts eight bits of port/register select data as defined in Table 2. The SCL line high-to-low transition after the eighth data bit then latches the selection of the appropriate internal register for the follow on data read or write operation. After latching the eight-bit data field, the TPS2384 pulls the SDA line low for one clock cycle, for the acknowledge pulse.

### **Data Write Cycle**

For a data write sequence, after the Port/Register address cycle, the TPS2384 accepts the eight bits of data as defined in the tables below. The data is latched into the previously selected Write Register, and the TPS2384 generates a data acknowledge pulse by pulling the SDA line low for one clock cycle. Common register functions act on all ports simultaneously. Per port registers are specific to the target port only.

To reset the interface, the host or master subsequently generates a stop bit by releasing the SDA line during the clock-high portion of an SCL pulse.

#### **Data Read Cycle**

For a data read sequence, after the register acknowledge bit, the master device generates a stop condition. This is followed by a second start condition, and retransmitting the device address as described in chip address above. For this cycle, however, the R/W bit is set to a 1 to signal the read operation. The TPS2384 again responds with an acknowledge pulse. The address acknowledge is then followed by sequentially presenting each of the eight data bits on the SDA line (MSB first), to be read by the host device on the rising edges of SCL. After eight bits are transmitted, the host acknowledges by pulling the SDA line high for one clock pulse. The completed data transfer is terminated with the host generating a stop condition.



Table 2. Common Write Register Port /Register Addressing

BIT	FUNCTION	STATE	PRESET STATE
D7	Future expansion		0
D6	Register select MSB	0000 = common read – port fault status, ID and rev 0001 = common control write – software reset, ports shutdown and ac disc 0010 = write control reg 1 – function calls and discovery, disable fault disable 0011 = write control reg 2 – classify current limit, power over I cntrl & AD abort	
D5	Register select Bit 3	0100 = read status reg 1 – fault status, class Infor, power Infor 0101 = read status reg 2 – detection status 0110 = read R result reg 1 – A/D resistor register (lower bits) 0111 = read R result reg 2 – A/D resistor register (upper bits)	0000
D4	Register select Bit 2	1000 = read V result reg 1 – A/D voltage register (lower bits) 1001 = read V result reg 2 – A/D voltage register (upper bits) 1010 = read I result reg 1 – A/D current register (lower bits) 1011 = read I result reg 2 – A/D current register (upper bits)	
D3	Register select LSB	1100 = read T result reg 1 - A/D temperature register (lower bits) 1101 = read T result reg 2 - A/D temperature register (upper bits) 1110 = spare 1111 = common write - test disable timer, discovery loop	
D2	Future expansion		0
D1	Port address MSB	00 = port 1 01 = port 2	00
D0	Port address	10 = port 3 11 = port 4	00

## Table 3. Common Register Read Register, Address = 0000 Port Fault Status & Chip ID/Rev

BIT	FUNCTION	STATE	PRESET STATE
D7	Port 4 general Fault status	0 = no fault 1 = port fault	0
D6	Port 3 general Fault status	0 = no fault 1 = port fault	0
D5	Port 2 general Fault status	0 = no fault 1 = port fault	0
D4	Port 1 general Fault status	0 = no fault 1 = port fault	0
D3	Chip rev	00 = rev - 01 = rev 1	00
D2		10 = rev 2 11 = rev 3	00
D1	Chia ID	00 = TPS2383 01= TPS2383A	10
D0	Chip ID	10 = TPS2384 11 = reserve	10

Table 4. Common Register Write Register, Address = 1111 Test Register

BIT	FUNCTION	STATE	PRESET STATE
D7	Spare		0
D6	TSD Test	0 = normal 1 = Force TSD condition	0
D5	Discovery 1 & 2	0 = normal 1 = all 4-port discovery 1 and discovery 2 – halt	0
D4	Discovery timers	0 = normal (4-ms discovery 1 & discovery 2) 1 = timers Disable	0
D3	POR disable	0 = normal POR timing 1 = force POR to a non-reset state	0
D2	DC disconnect timer	0 = normal 1 = 300-ms disconnect timer disable	0
D1	TED timer	0 = normal 1 = 750-ms ted timer disable	0
D0	Spare		0

## Table 5. Common Control RegisterWrite Register, address = 0001

BIT	FUNCTION	STATE	PRESET STATE
D7	Spare		0
D6	Spare		0
D5	TSD fault	0 = Normal 1 = Fault Disable	0
D4	AC high	0 = Off 1 = Osc Hi Driver On	0
D3	AC low	0 = Off 1 = Osc Lo Driver On On	0
D2	Over/under voltage faults	0 = Normal 1 = Faults Disable	0
D1	All Ports Disable	0 = Normal 1 = All Ports shut down (no ramp)	0
D0	Software RESET	0 = Normal operation 1 = Reset all circuits and start a POR timing cycle	0

## Table 6. Individual Port Write Control Register 1 Address 0010 (One per port)

Bit	FUNCTION	State	PRESET STATE
D7	Spare		0
D6	Spare		0
D5	Discovery Fault Disable	0 = Normal mode 1 = Disable internal Discovery Fault Limits	0
D4	Disconnect Disable	0 = Normal mode 1 = Disable the effect of the logic signal from the Disconnect detection circuits (used for test only)	0
D3	Function Bit 3	0000 = Disable Function (Power Down & reset all functions) 0001 = Discovery 1 Function 0010 = Discovery 2 Function 0011 = Port Voltage Sample Function	
D2	Function Bit 2	0100 = Legacy Detection Function 0101 = Classification Function 0110 = Ramp Up/Power Function 0111 = Continuous Sample Function	0000
D1	Function Bit 1	1000 = Ramp Power Down Function 1001 = AC Low 1010 = AC High 1011 = Port Current Sample Function	
D0	Function Bit 0	1100 = Die Temperature Sample Function 1101 = Spare 1110 = Spare 1111 = Spare	

## Table 7. Individual Port Write Control 2 Write register, address = 0011 (One per port)

BIT	FUNCTION	STATE	PRESET STATE
D7	Spare		0
D6	Spare		0
D5	Spare		0
D4	Port Enable	0 = Normal 1 = Port Disable	0
D3	A/D Start	0 = Normal 1 = Start A/D (self Clearing)	0
D2	A/D Abort	0 = Normal 1 = Abort	0
D1	Power Mode Class Over Current Threshold Enable	0 = Full Current (375mA) Over Current Threshold 1 = Deny Current Beyond Class 1 (90mA) or Class 2 (160mA) depending on Class contents in Register 4.	0
D0	Spare		0

Table 8. Port Status Port 1 through 4 Read Register, address = 0100 (One per port)

BIT	FUNCTION	STATE	PRESET STATE
D7	Discovery Status	0 = Normal 1 = Discovery Fail	0
D6	Function Done Bit	0 = Normal 1 = Function Complete (Self Clearing by new a function Write)	0
D5	Port Class	000 = Class 0 	000
D4	Port Class	010 = Class 2	000
D3	Port Class	011 = Class 3 100 = Class 4	
D2	Fault status (MSB)	000 = No Faults 001 = UV/OV fault 010 = Thermal Fault	
D1	Fault status	011 = Overload Current > 50mS fault 100 = Load disconnect	000
D0	Fault status (LSB)	101 = Reserved for future 110 = Reserved for future 111 = Reserved for future	

## Table 9. Port Mode Status Port 1 through 4 Read Register, address = 0101 (One per port)

BIT	FUNCTION	STATE	PRESET STATE
D7	Spare		0
D6	Spare		0
D5	Spare		0
D4	Watch dog timer	0 = Not Active 1 = Active	0
D3	A/D status	0 = Not Active 1 = Active	0
D2	Detection status (MSB)	000 = Disable 001 = Searching	
D1	Detection status	010 = Power Delivery 011 = Fault 100 = Test	000
D0	Detection status (LSB)	101= Other Fault 110 = undefined 111 = undefined	



## A/D Register (Resistance, Voltage, Current & Temperature)

Table 10. Analog to Digital Lower Bits Resistor Register Read Register, Address = 0110 (One per port)

Bit	Function	State	PRESET STATE
D7	A/D bit 7		
D6	A/D bit 6		
D5	A/D bit 5		
D4	A/D bit 4		
D3	A/D bit 3	A/D Lower Bits	0
D2	A/D bit 2		
D1	A/D bit 1		
D0	A/D bit 0		

## Table 11. Common Analog to Digital Upper Bits Resistor Register/Read Register, Address = 0111 (One per port)

Bit	Function	State	PRESET STATE
D7	Resistor measurement complete	0 = Measurement active (Bit set low at the start of Discovery 1 or Discovery 2) 1 = Measurement complete (Bit set high after A/D is completed during Discovery 1 or Discovery 2)	0
D6	A/D bit 14		
D5	A/D bit 13		
D4	A/D bit 12		
D3	A/D bit 11	A/D Upper bits	0
D2	A/D bit 10		
D1	A/D bit 9		
D0	A/D bit 8		

#### Table 12. Analog to Digital Lower Bits Voltage Register/Read Register, Address = 1000 (One per port)

Bit	Function	State	PRESET STATE
D7	A/D bit 7		
D6	A/D bit 6		
D5	A/D bit 5		
D4	A/D bit 4	A/D Leaves hits	
D3	A/D bit 3	A/D lower bits	0
D2	A/D bit 2		
D1	A/D bit 1		
D0	A/D bit 0		

## Table 13. Common Analog to Digital Upper Bits Voltage Register/Read Register, Address = 1001 (One per port)

BIT	FUNCTION	STATE	PRESET STATE	
D7	Voltage measure complete	0 = Measurement active (Bit set low when A/D begins a voltage measurement) 1 = Measurement complete (Bit set high after A/D has completed a voltage measurement)	0	
D6	A/D bit 14			
D5	A/D bit 13			
D4	A/D bit 12			
D3	A/D bit 11	A/D upper bits	0	
D2	A/D bit 10			
D1	A/D bit 9			
D0	A/D bit 8			

### Table 14. Analog to Digital Lower Bits Current Register/Read Register, Address = 1010 (One per port)

BIT	FUNCTION	STATE	PRESET STATE
D7	A/D bit 7		
D6	A/D bit 6		
D5	A/D bit 5		
D4	A/D bit 4	A/D Lower Bits	0
D3	A/D bit 3		
D2	A/D bit 2		
D1	A/D bit 1		
D0	A/D bit 0		

## Table 15. Common Analog to Digital Upper Bits Current Register/Read Register, Address = 1011 (One per port)

BIT	FUNCTION	STATE	PRESET STATE
D7	Current Measure Complete	0 = Measurement active (Bit set low when A/D begins a current measurement) 1 = Measurement complete (Bit set high after A/D has completed a current measurement)	0
D6	A/D bit 14		
D5	A/D bit 13		
D4	A/D bit 12		
D3	A/D bit 11	A/D upper bits	0
D2	A/D bit 10		
D1	A/D bit 9		
D0	A/D bit 8		



Table 16. Analog to Digital Lower Bits Temperature Register/Read Register, Address = 1100 (One per port)

BIT	FUNCTION	STATE	PRESET STATE
D7	A/D bit 7		
D6	A/D bit 6		
D5	A/D bit 5		
D4	A/D bit 4		
D3	A/D bit 3	A/D lower bits	0
D2	A/D bit 2		
D1	A/D bit 1		
D0	A/D bit 0		

## Table 17. Common Analog to Digital Upper Bits Temperature Register/Read Register, Address = 1101 (One per port)

BIT	FUNCTION	STATE	PRESET STATE
D7	Temperature Measure Complete	0 = Measurement active (Bit set low when A/D begins a temperature measurement) 1 = Measurement complete (Bit set high after A/D has completed a temperature measurement)	0
D6	A/D bit 14		
D5	A/D bit 13		
D4	A/D bit 12		
D3	A/D bit 11	A/D upper bits	0
D2	A/D bit 10		
D1	A/D bit 9		
D0	A/D bit 8		

## **TPS 2384 AC DRIVE APPLICATION SCHEMATIC**

## AC\_Hi & Low W/O External FET Configurations

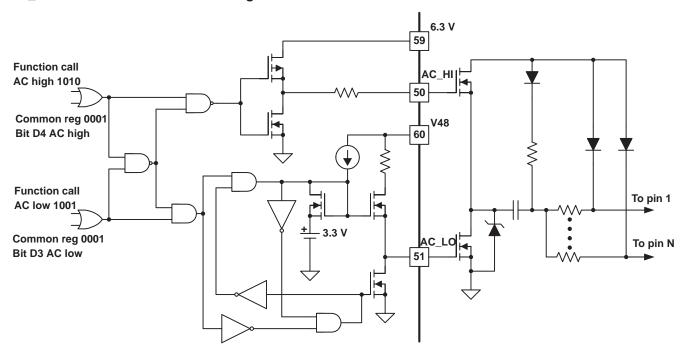


Figure 21.



## **TPS2384 SYSTEM BLOCK DIAGRAM**

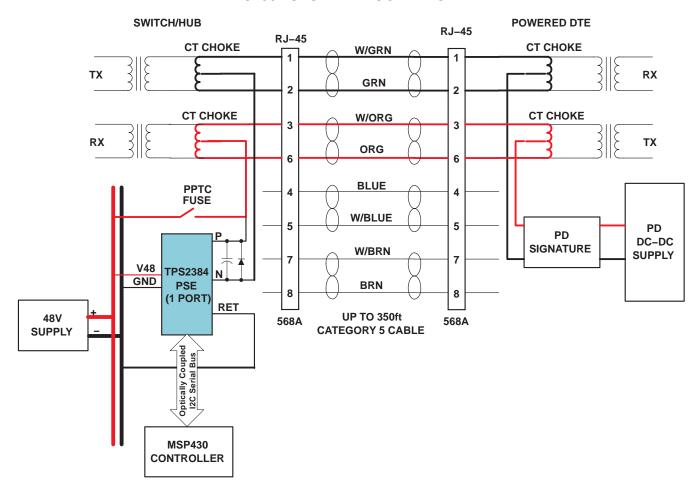


Figure 22.

## TPS2384 BASIC 4 PORT (PMM) ISOLATED CONFIGURATION WITH AC DISCONNECT

TPS2384 basic 4-port isolated configuration with ac disconnect.

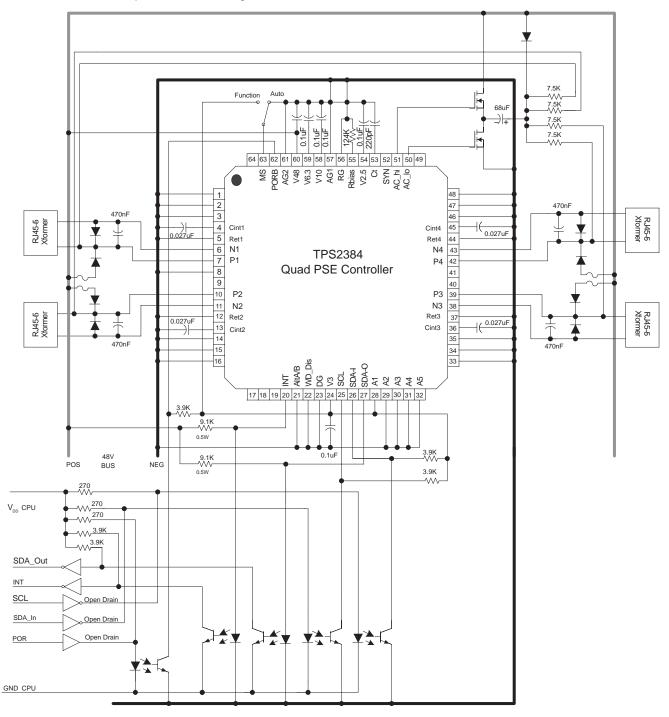


Figure 23.



## PAP (S-PQFP-G64)

## PowerPAD™ PLASTIC QUAD FLATPACK



NOTES:

- A. All linear dimensions are in millimeters.
- B. This drawing is subject to change without notice.
- C. Body dimensions do not include mold flash or protrusion
- D. This package is designed to be soldered to a thermal pad on the board. Refer to Technical Brief, PowerPad Thermally Enhanced Package, Texas Instruments Literature No. SLMA002 for information regarding recommended board layout. This document is available at www.ti.com <a href="http://www.ti.com">http://www.ti.com</a>.
- E. Falls within JEDEC MS-026

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