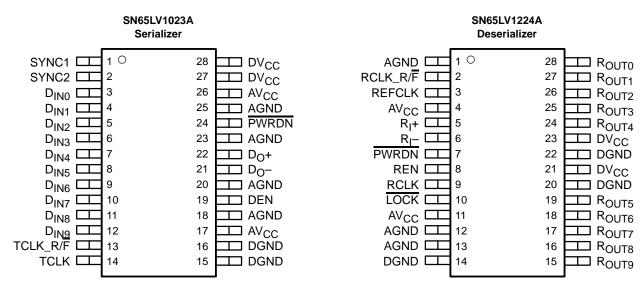
- 100-Mbps to 660-Mbps Serial LVDS Data Payload Bandwidth at 10-MHz to 66-MHz System Clock
- Pin-Compatible Superset of NSM DS92LV1023/DS92LV1224
- Chipset (Serializer/Deserializer) Power Consumption <450 mW (Typ) at 66 MHz</li>
- Synchronization Mode for Faster Lock

- Lock Indicator
- No External Components Required for PLL

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- Low-Cost 28-Pin SSOP Package
- Industrial Temperature Qualified, T<sub>A</sub> = -40°C to 85°C
- Programmable Edge Trigger on Clock
- Flow-Through Pinout for Easy PCB Layout



•

## description

The SN65LV1023A serializer and SN65LV1224A deserializer comprise a 10-bit serdes chipset designed to transmit and receive serial data over LVDS differential backplanes at equivalent parallel word rates from 10 MHz to 66 MHz. Including overhead, this translates into a serial data rate between 120-Mbps and 792-Mbps payload encoded throughput.

Upon power up, the chipset link can be initialized via a synchronization mode with internally generated SYNC patterns, or the deserializer can be allowed to synchronize to random data. By using the synchronization mode, the deserializer establishes lock within specified, shorter time parameters.

The device can be entered into a power-down state when no data transfer is required. Alternatively, a mode is available to place the output pins in the high-impedance state without losing PLL lock.

The SN65LV1023A and SN65LV1224A are characterized for operation over ambient air temperature of  $-40^{\circ}$ C to 85°C.

ORDERING INFORMATION						
DEVICE PART NUMBER						
Serializer	SN65LV1023ADB					
Deserializer	SN65LV1224ADB					



Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet.

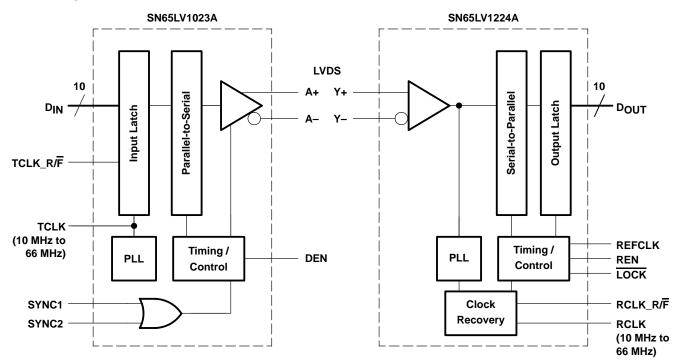
PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.



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## block diagrams



## functional description

The SN65LV1023A and SN65LV1224A are a 10-bit serializer/deserializer chipset designed to transmit data over differential backplanes or unshielded twisted pair (UTP) at clock speeds from 10 MHz to 66 MHz. The chipset has five states of operation: initialization mode, synchronization mode, data transmission mode, power-down mode, and high-impedance mode. The following sections describe each state of operation.

### initialization mode

Initialization of both devices must occur before data transmission can commence. Initialization refers to synchronization of the serializer and deserializer PLLs to local clocks.

When  $V_{CC}$  is applied to the serializer and/or deserializer, the respective outputs enter the high-impedance state, while on-chip power-on circuitry disables internal circuitry. When  $V_{CC}$  reaches 2.45 V, the PLL in each device begins locking to a local clock. For the serializer, the local clock is the transmit clock (TCLK) provided by an external source. For the deserializer, a local clock must be applied to the REFCLK pin. The serializer outputs remain in the high-impedance state, while the PLL locks to the TCLK.



## functional description (continued)

### synchronization mode

The deserializer PLL must synchronize to the serializer in order to receive valid data. Synchronization can be accomplished in one of two ways:

Rapid Synchronization: The serializer has the capability to send specific SYNC patterns consisting of six
ones and six zeros switching at the input clock rate. The transmission of SYNC patterns enables the
deserializer to lock to the serializer signal within a deterministic time frame. This transmission of SYNC
patterns is selected via the SYNC1 and SYNC2 inputs on the serializer. Upon receiving valid SYNC1 or
SYNC2 pulse (wider than 6 clock cycles), 1026 cycles of SYNC pattern are sent.

When the deserializer detects edge transitions at the LVDS input, it attempts to lock to the embedded clock information. The deserializer LOCK output remains high while its PLL locks to the incoming data or SYNC patterns present on the serial input. When the deserializer locks to the LVDS data, the LOCK output goes low. When LOCK is low, the deserializer outputs represent incoming LVDS data. One approach is to tie the deserializer LOCK output directly to SYNC1 or SYNC2.

• Random-Lock Synchronization: The deserializer can attain lock to a data stream without requiring the serializer to send special SYNC patterns. This allows the SN65LV1224A to operate in open-loop applications. Equally important is the deserializer's ability to support hot insertion into a running backplane. In the open-loop or hot-insertion case, it is assumed the data stream is essentially random. Therefore, because lock time varies due to data stream characteristics, the exact lock time cannot be predicted. The primary constraint on the random lock time is the initial phase relation between the incoming data and the REFCLK when the deserializer powers up.

The data contained in the data stream can also affect lock time. If a specific pattern is repetitive, the deserializer could enter false lock—falsely recognizing the data pattern as the start/stop bits. This is referred to as repetitive multitransition (RMT); see Figure 1 for RMT examples. This occurs when more than one low-high transition takes place per clock cycle over multiple cycles. In the worst case, the deserializer could become locked to the data pattern rather than the clock. Circuitry within the deserializer can detect that the possibility of false lock exists. Upon detection, the circuitry prevents the LOCK output from becoming active until the potential false lock pattern changes. Notice that the RMT pattern only affects the deserializer lock time, and once the deserializer is in lock, the RMT pattern does not affect the deserializer state as long as the same data boundary happens each cycle. The deserializer does not go into lock until it finds a unique four consecutive cycles of data boundary (stop/start bits) at the same position.

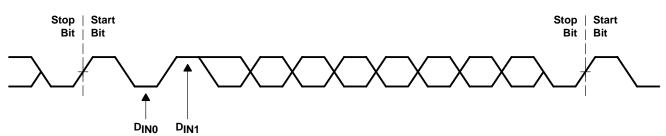
The deserializer stays in lock until it cannot detect the same data boundary (stop/start bits) for four consecutive cycles. Then the deserializer goes out of lock and hunts for the new data boundary (stop/start bits). In the event of loss of synchronization, the LOCK pin output goes high and the outputs (including RCLK) enter a high-impedance state. The user's system should monitor the LOCK pin in order to detect a loss of synchronization. Upon detection of locs of lock, sending sync patterns for resynchronization is desirable if reestablishing lock within a specific time is critical. However, the deserializer can lock to random data as previously noted.



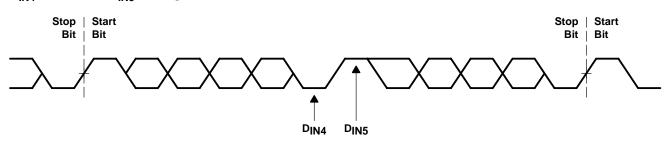
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### synchronization mode (continued)

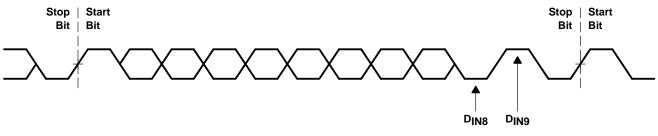
DIN0 Held Low and DIN1 Held High



DIN4 Held Low and DIN5 Held High



DIN8 Held Low and DIN9 Held High





### data transmission mode

After initialization and synchronization, the serializer accepts parallel data from inputs  $D_{IN0}-D_{IN9}$ . The serializer uses the TCLK input to latch the incoming data. The TCLK\_R/ $\overline{F}$  pin selects which edge the serializer uses to strobe incoming data. If either of the SYNC inputs is high for six TCLK cycles, the data at  $D_{IN0}-D_{IN9}$  is ignored regardless of the clock edge selected and 1026 cycles of SYNC pattern are sent.

After determining which clock edge to use, a start and stop bit, appended internally, frames the data bits in the register. The start bit is always high and the stop bit is always low. The start and stop bits function as the embedded clock bits in the serial stream.

The serializer transmits serialized data and appended clock bits (10+2 bits) from the serial data output (DO±) at 12 times the TCLK frequency. For example, if TCLK is 66 MHz, the serial rate is  $66 \times 12 = 792$  Mbps. Because only 10 bits are input data, the useful data rate is 10 times the TCLK frequency. For instance, if TCLK = 66 MHz, the useful data rate is  $66 \times 10 = 660$  Mbps. The data source, which provides TCLK, must be in the range of 10 MHz to 66 MHz.



## functional description (continued)

The serializer outputs (DO $\pm$ ) can drive point-to-point connections or limited multipoint or multidrop backplanes. The outputs transmit data when the enable pin (DEN) is high, PWRDN = high, and SYNC1 and SYNC2 are low. When DEN is driven low, the serializer output pins enter the high-impedance state.

Once the deserializer has synchronized to the serializer, the  $\overline{LOCK}$  pin transitions low. The deserializer locks to the embedded clock and uses it to recover the serialized data.  $R_{OUT}$  data is valid when  $\overline{LOCK}$  is low, otherwise  $R_{OUT0}-R_{OUT9}$  is invalid. The  $R_{OUT0}-R_{OUT9}$  data is strobed out by RCLK. The specific RCLK edge polarity to be used is selected by the RCLK\_R/F input. The  $R_{OUT0}-R_{OUT9}$ ,  $\overline{LOCK}$  and RCLK outputs can drive a maximum of three CMOS input gates (15-pF load. total for all three) with a 66-MHz clock.

## power down

When no data transfer is required, the power-down mode can be used. The serializer and deserializer use the power-down state, a low-power sleep mode, to reduce power consumption. The deserializer enters power down when you drive PWRDN and REN low. The serializer enters power down when you drive PWRDN low. In power down, the PLL stops and the outputs enter a high-impedance state, which disables load current and reduces supply current to the milliampere range. To exit power down, you must drive the PWRDN pin high.

Before valid data exchanges between the serializer and deserializer can resume, you must reinitialize and resynchronize the devices to each other. Initialization of the serializer takes 1026 TCLK cycles. The deserializer initialize and drives LOCK high until lock to the LVDS clock occurs.

## high-impedance mode

The serializer enters the high-impedance mode when the DEN pin is driven low. This puts both driver output pins (DO+ and DO-) into a high-impedance state. When you drive DEN high, the serializer returns to the previous state, as long as all other control pins remain static (SYNC1, SYNC2, PWRDN, TCLK\_R/ $\overline{F}$ ). When the REN pin is driven low, the deserializer enters high-impedance mode. Consequently, the receiver output pins (R<sub>OUT0</sub>-R<sub>OUT9</sub>) and RCLK are placed into the high-impedance state. The LOCK output remains active, reflecting the state of the PLL.

INPU	TS	OUTPUTS			
PWRDN	REN	ROUT[0:9]	LOCK	RCLK	
Н	Н	Z	Н	Z	
н	Н	Active	L	Active	
L	Х	Z	Z	Z	
Н	L	Z	Active	Z	

### **Deserializer Truth Table**

NOTES: 1. LOCK output reflects the state of the deserializer with regard to the selected data stream.

- RCLK active indicates the RCLK is running if the deserializer is locked. The timing of RCLK with respect to ROUT is determined by RCLK\_R/F.
- 3. ROUT and RCLK are 3-stated when  $\overline{LOCK}$  is asserted high.

### failsafe biasing for the SN65LV1224A

The SN65LV1224A has an input threshold sensitivity of  $\pm$ 50 mV. This allows for greater differential noise margin in the SN65LV1224A. However, in cases where the receiver input is not being actively driven, the increased sensitivity of the SN65LV1224A can pickup noise as a signal and cause unintentional locking. This may occur when the input cable is disconnected. SN65LV1224A has an on-chip fail-safe circuit that drives the serial input and LOCK Signal high. The response time of the fail-safe circuit depends on interconnect characteristics.



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## **Terminal Functions**

## serializer

PIN	NAME	DESCRIPTION
18, 20, 23, 25	AGND	Analog circuit ground (PLL and analog circuits)
17, 26	AVCC	Analog circuit power supply (PLL and analog circuits)
19	DEN	LVTTL logic input. Low puts the LVDS serial output into the high-impedance state. High enables serial data output.
15, 16	DGND	Digital circuit ground
3–12	D <sub>IN0</sub> – D <sub>IN9</sub>	Parallel LVTTL data inputs
21	D <sub>O</sub> -	Inverting LVDS differential output
22	D <sub>O</sub> +	Noninverting LVDS differential output
27, 28	DVCC	Digital circuit power supply
24	PWRDN	LVTTL logic input. Asserting this pin low turns off the PLL and places the outputs into the high-impedance state, putting the device into a low-power mode.
1, 2	SYNC1, SYNC2	LVTTL logic inputs SYNC1 and SYNC2 are ORed together. When at least one of the two pins is asserted high for 6 cycles of TCLK, the serializer initiates transmission of a minimum 1026 SYNC patterns. If after completion of the transmission of 1026 patterns SYNC continues to be asserted, then the transmission continues until SYNC is driven low and if the time SYNC holds > 6 cycles, another 1026 SYNC pattern transmission initiates.
13	TCLK_R/F	LVTTL logic input. Low selects a TCLK falling-edge data strobe; high selects a TCLK rising-edge data strobe.
14	TCLK	LVTTL-level reference clock input. The SN65LV1023A accepts a 10-MHz to 66-MHz clock. TCLK strobes parallel data into the input latch and provides a reference frequency to the PLL.

## deserializer

PIN	NAME	DESCRIPTION
1, 12, 13	AGND	Analog circuit ground (PLL and analog circuits)
4, 11	AVCC	Analog circuit power supply (PLL and analog circuits)
14, 20, 22	DGND	Digital circuit ground
21, 23	DVCC	Digital circuit power supply
10	LOCK	LVTTL level output. LOCK goes low when the deserializer PLL locks onto the embedded clock edge.
7	PWRDN	LVTTL logic input. Asserting this pin low turns off the PLL and places outputs into a high-impedance state, putting the device into a low-power mode.
2	RCLK_R/F	LVTTL logic input. Low selects an RCLK falling-edge data strobe; high selects an RCLK rising-edge data strobe.
9	RCLK	LVTTL level output recovered clock. Use RCLK to strobe ROUTx.
3	REFCLK	LVTTL logic input. Use this pin to supply a REFCLK signal for the internal PLL frequency.
8	REN	LVTTL logic input. Low places ROUT0-ROUT9 and RCLK in the high-impedance state.
5	R <sub>I</sub> +	Serial data input. Noninverting LVDS differential input
6	RI–	Serial data input. Inverting LVDS differential input
15–19, 24–28	ROUT0-ROUT9	Parallel LVTTL data outputs



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## absolute maximum ratings (unless otherwise noted)<sup>†</sup>

V <sub>CC</sub> to GND	–0.3 V to 4 V
LVTTL input voltage	
LVTTL output voltage	$-0.3 \text{ V to } (\text{V}_{\text{CC}} + 0.3 \text{ V})$
LVDS receiver input voltage	−0.3 V to 3.9 V
LVDS driver output voltage	0.3 V to 3.9 V
LVDS output short circuit duration	
Electrostatic discharge: HBM	up to 6 kV
MM	
Junction temperature	150°C
Storage temperature	– 65°C to 150°C
Lead temperature (soldering, 4 seconds)	
Maximum package power dissipation, T <sub>A</sub> = 25°C	1.27 W
Package derating	10.3 mW/°C above 25°C

<sup>†</sup> Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

## recommended operating conditions

	MIN	NOM	MAX	UNIT
Supply voltage, V <sub>CC</sub> <sup>‡</sup>	3	3.3	3.6	V
Receiver input voltage range	0		2.4	V
Receiver input common mode range, $V_{CM}$	$\frac{V_{\text{ID}}}{2}$	2	$2.4 - \left(\frac{V_{\text{ID}}}{2}\right)$	V
Supply noise voltage			100	mV <sub>P-P</sub>
Operating free-air temperature, T <sub>A</sub>	-40	25	85	°C

<sup>‡</sup> By design, DVCC and AVCC are separated internally and does not matter what the difference is for |DVCC-AVCC|, as long as both are within 3 V to 3.6 V.



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# electrical characteristics over recommended operating supply and temperature ranges (unless otherwise noted)

	PARAMETER	TEST CONDITION	IS	MIN	TYP	MAX	UNIT
SERIAL	IZER LVCMOS/LVTTL DC SPECIFICATIONS	6 (see Note 4)					
VIH	High-level input voltage			2		VCC	V
VIL	Low-level input voltage			GND		0.8	V
V <sub>CL</sub>	Input clamp voltage	I <sub>CL</sub> = -18 mA			-0.86	-1.5	V
I <sub>IN</sub>	Input current (see Note 5)	V <sub>IN</sub> = 0 V or 3.6 V		-200	±100	200	μΑ
DESER	IALIZER LVCMOS/LVTTL DC SPECIFICATIO	NS (see Note 6)					
VIH	High-level input voltage			2		VCC	V
VIL	Low-level input voltage			GND		0.8	V
VCL	Input clamp voltage	$I_{CL} = -18 \text{ mA}$			-0.62	-1.5	V
IIN	Input current (pull-up and pull-down resistors on inputs)	V <sub>IN</sub> = 0 V or 3.6 V		-200		200	μA
Vон	High-level output voltage	I <sub>OH</sub> = -5 mA		2.2	3	VCC	V
VOL	Low-level output voltage	I <sub>OL</sub> = 5 mA		GND	0.25	0.5	V
los	Output short-circuit current	V <sub>OUT</sub> = 0 V		-15	-47	-85	mA
IOZ	High-impedance output current	PWRDN or REN = 0.8 V, VOUT	= 0 V or V <sub>CC</sub>	-10	±1	10	μΑ
SERIAL	IZER LVDS DC SPECIFICATIONS (apply to	pins DO+ and DO-)					
VOD	Output differential voltage (DO+)-(DO-)	R <sub>L</sub> = 27 Ω, See Figure 19		350	450		mV
$\Delta V_{OD}$	Output differential voltage unbalance					35	mV
VOS	Offset voltage			1.1	1.2	1.3	V
ΔVOS	Offset voltage unbalance				4.8	35	mV
IOS	Output short circuit current	$\frac{D0 = 0 \text{ V}, D_{INX} = \text{high},}{PWRDN}$ and DEN = 2.4 V			-10	-90	mA
IOZ	High-impedance output current	$\overline{PWRDN}$ or $DEN = 0.8$ V, $DO = 0.00$	OV or V <sub>CC</sub>	-10	±1	10	μΑ
IOX	Power-off output current	$V_{CC}$ = 0 V, DO = 0 V or 3.6 V		-20	±1	25	μΑ
DESER	IALIZER LVDS DC SPECIFICATIONS (apply	to pins RI+ and RI–)					
VTH	Differential threshold high voltage	V <sub>CM</sub> = 1.1 V				50	mV
VTL	Differential threshold low voltage			-50			mV
I	land a summant	$V_{IN}$ = 2.4 V, $V_{CC}$ = 3.6 V or 0 V		-10	±1	15	A
IIN	Input current	$V_{IN}$ = 0 V, $V_{CC}$ = 3.6 V or 0 V				10	μA
SERIAL	IZER SUPPLY CURRENT (applies to pins D	VCC and AVCC)					
1			f = 10 MHz		20	25	~ ^
ICCD	Serializer supply current worst case	$R_L = 27 \Omega$ , See Figure 4	f = 66 MHz		55	70	mA
ICCXD	Serializer supply current	PWRDN = 0.8 V			200	500	μA
DESER	IALIZER SUPPLY CURRENT (applies to pin	s DVCC and AVCC)	<u>.</u>				
1007		Cu – 15 pE Soo Eiguro 4	f = 10 MHz		15	35	m ^
ICCR	Deserializer supply current, worst case	C <sub>L</sub> = 15 pF, See Figure 4	f = 66 MHz		80	95	mA
ICCXR	Deserializer supply current, power down	PWRDN = 0.8 V, REN = 0.8 V			0.36	1	mA

NOTES: 4. Apply to DINO-DIN9, TCLK, PWRDN, TCLK\_R/F, SYNC1, SYNC2, DEN

High I<sub>IN</sub> values are due to pullup and pulldown resistors on the inputs.

6. Apply to pins PWRDN, RCLK\_R/F, REN, REFCLK = inputs; apply to pins ROUTX, RCLK, LOCK = outputs



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# serializer timing requirements for TCLK over recommended operating supply and temperature ranges (unless otherwise specified)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
<sup>t</sup> TCP	Transmit clock period		15.15	Т	100	ns
<sup>t</sup> TCIH	Transmit clock high time		0.4T	0.5T	0.6T	ns
<sup>t</sup> TCIL	Transmit clock low time		0.4T	0.5T	0.6T	ns
<sup>t</sup> t(CLK)	TCLK input transition time			3	6	ns
<sup>t</sup> JIT	TCLK input jitter	See Figure 18			150	ps (RMS)

# serializer switching characteristics over recommended operating supply and temperature ranges (unless otherwise specified)

	PARAMETER		TEST CONDITIONS	MIN	ТҮР	MAX	UNIT
<sup>t</sup> TLH(L)	LVDS low-to-high transition time		$R_L = 27 \Omega$ , $C_L = 10 \text{ pF}$ to GND,		0.2	0.4	ns
<sup>t</sup> LTHL(L)	LVDS high-to-low transition time		See Figure 5		0.25	0.4	ns
<sup>t</sup> su(DI)	D <sub>IN0</sub> –D <sub>IN9</sub> setup to TCLK		$R_L = 27 \Omega$ , $C_L = 10 \text{ pF to GND}$ ,	0.5			ns
<sup>t</sup> su(DI)	DIN0-DIN9 hold from TCLK		See Figure 8	4			ns
<sup>t</sup> d(HZ)	DO± high-to-high-impedance-sta	te delay			2.5	5	
<sup>t</sup> d(LZ)	DO± low-to-high-impedance-stat	e delay			2.5	5	
<sup>t</sup> d(ZH)	DO± high-to-high-impedance-sta high delay	te-to-	$R_L = 27 \Omega$ , $C_L = 10 pF$ to GND, See Figure 9		5	10	ns
<sup>t</sup> d(ZL)	DO± high-to-high-impedance-sta delay	te-to-low			6.5	10	
<sup>t</sup> w(SPW)	SYNC pulse duration			6×tTCP			ns
<sup>t</sup> (PLD)	Serializer PLL lock time		R <sub>L</sub> = 27 Ω, See Figure 11	1026×tTCP			ns
<sup>t</sup> d(S)	Serializer delay		$R_L$ = 27 Ω, See Figure 12	tTCP+1	tTCP+2	tTCP+3	ns
		10 MHz				230	
<sup>t</sup> DJIT	Deterministic jitter	66 MHz	R <sub>L</sub> = 27 Ω, C <sub>L</sub> = 10 pF to GND			150	ps
<sup>t</sup> RJIT	Random jitter		$R_L$ = 2.7 Ω, $C_L$ = 10 pF to GND		10	19	ps (RMS)



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# deserializer timing requirements for REFCLK over recommended operating supply and temperature ranges (unless otherwise specified)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
<sup>t</sup> RFCP	REFCLK period		15.15	Т	100	ns
<sup>t</sup> RFDC	REFCLK duty cycle		30%	50%	70%	
<sup>t</sup> t(RF)	REFCLK transition time			3	6	ns

# deserializer switching characteristics over recommended operating supply and temperature ranges (unless otherwise specified)

	PARAMETER	TEST CONDITIONS	PIN/FREQ	MIN	ТҮР	MAX	UNIT
<sup>t</sup> (RCP)	Receiver out clock period	<sup>t</sup> (RCP) = <sup>t</sup> (TCP), See Figure 12	RCLK	15.15		100	ns
<sup>t</sup> TLH(C)	CMOS/TTL low-to-high transition time	C <sub>I</sub> = 15 pF,	ROUTO-ROUT9,		1.2	2.5	
<sup>t</sup> THL(C)	CMOS/TTL high-to-low transition time	See Figure 6	LOCK, RCLK		1.1	2.5	ns
	Deserializer delay,	Room temperature,	10 MHz	1.75×t <sub>RCP</sub> +4.2		1.75×t <sub>RCP</sub> +12.6	
<sup>t</sup> d(D) <sup>†</sup>	See Figure 13	3.3 V	66 MHz	1.75×t <sub>RCP</sub> +7.4		1.75×t <sub>RCP</sub> +9.7	ns
	ROUTx data valid		RCLK 10 MHz	0.4×tRCP	0.5×tRCP		
<sup>t</sup> (ROS)	before RCLK		RCLK 66 MHz	0.4×tRCP	0.5×tRCP		
	R <sub>OUTx</sub> data valid	See Figure 14	10 MHz	-0.4×tRCP	-0.5×tRCP		ns
<sup>t</sup> (ROH)	after RCLK		66 MHz	-0.4×tRCP	-0.5×tRCP		
<sup>t</sup> (RDC)	RCLK duty cycle			40%	50%	60%	
<sup>t</sup> d(HZ)	High-to-high-impedan ce state delay				6.5	8	ns
<sup>t</sup> d(LZ)	Low-to-high-impedan ce state delay	0.5.45	Davie Davie		4.7	8	ns
<sup>t</sup> d(HR)	High-impedance state-to-high delay	See Figure 15	ROUT0-ROUT9		5.3	8	ns
<sup>t</sup> d(ZL)	High-impedance-stat e-to-low delay				4.7	8	ns
	Deserializ <u>er PLL lo</u> ck		10 MHz			815 x tRFCP	
<sup>t</sup> (DSR1)	time from PWRDN (with SYNCPAT)		66 MHz			815 x tRFCP	
	Deserializer PLL lock	See Figure 16,	10 MHz	1		0.7	μs
<sup>t</sup> (DSR2)	time from SYNCPAT	Figure 17, and 66 MHz 66 MHz				0.2	1
<sup>t</sup> d(ZHLK)	High-impedance-stat e to-high delay (power up)		LOCK			3	ns

<sup>†</sup> The deserializer delay time for all frequencies does not exceed 2 serial bit times.

NOTE 7: t<sub>(DSR1)</sub> represents the time required for the deserializer to register that a lock has occurred upon powerup or when leaving the powerdown mode. t<sub>(DSR2)</sub> represents the time required to register that a lock has occurred for the powered up and enabled deserializer when the input (RI±) conditions change from not receiving data to receiving synchronization patterns (SYNCPATs). In order to specify deserializer PLL performance t<sub>DSR1</sub> and t<sub>DSR2</sub> are specified with REFCLK active and stable and specific conditions of SYNCPATs.



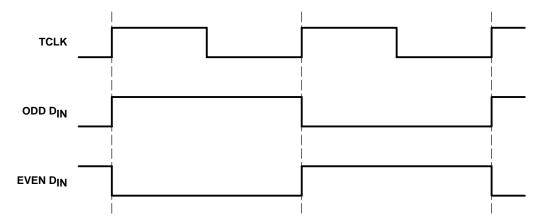
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# deserializer switching characteristics over recommended operating supply and temperature ranges (unless otherwise specified) (continued)

	PARAMETER	TEST CONDITIONS	PIN/FREQ	MIN	TYP	MAX	UNIT
	Deserializer noise	See Figure 18 and	10 MHz		3680		
<sup>t</sup> RNN	<sup>A</sup> margin	Note 8	66 MHz		540		ps

NOTE 8: tRNM represents the phase noise or jitter that the deserializer can withstand in the incoming data stream before bit errors occur.

## timing diagrams and test circuits





SUPPLY CURRENT

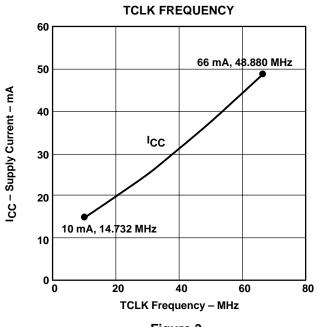
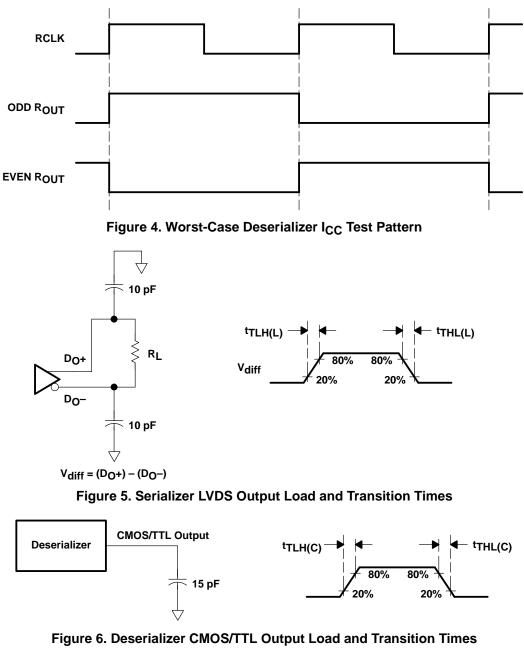


Figure 3.



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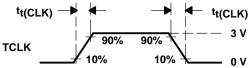
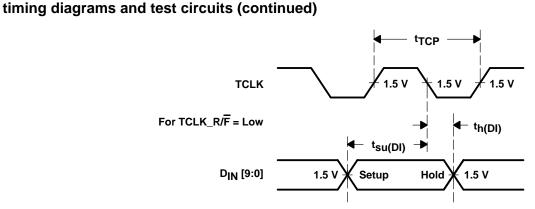


Figure 7. Serializer Input Clock Transition Time



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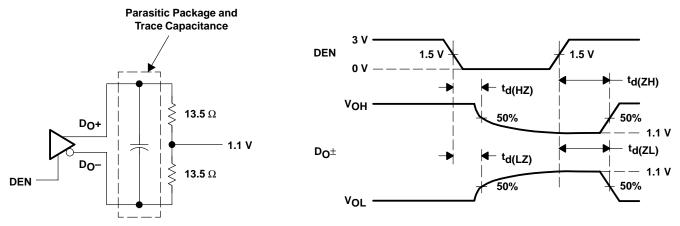


Figure 9. Serializer High-Impedance-State Test Circuit and Timing

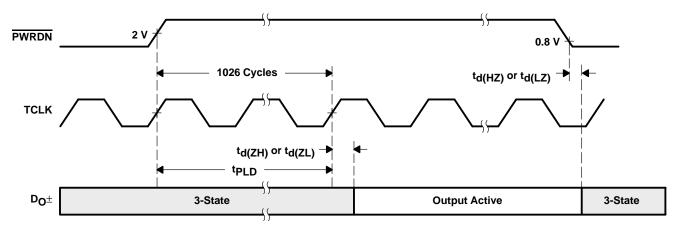


Figure 10. Serializer PLL Lock Time and PWRDN High-Impedance-State Delays

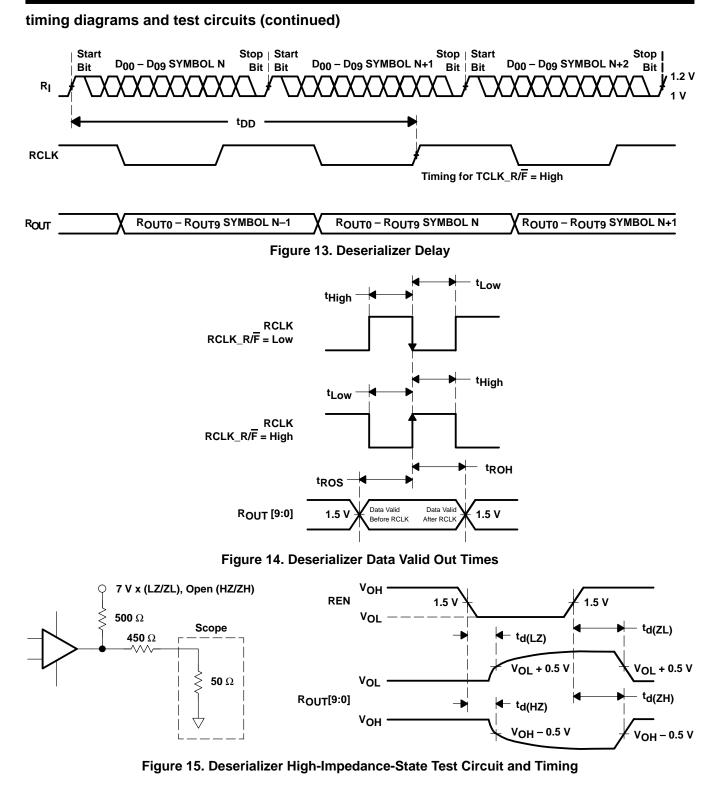


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# timing diagrams and test circuits (continued) REN PWRDN TCLK tw(SP) SYNC1 or SYNC2 $\mathbf{D}_{\mathbf{O}^{\pm}}$ DATA SYNC Pattern TCLK SYNC1 or tw(SP) Min. Timing Met SYNC2 DO± SYNC Pattern DATA Figure 11. SYNC Timing Delays DIN DINO - DIN9 SYMBOL N DIN0 - DIN9 SYMBOL N+1 <sup>t</sup>d(S) TCLK Timing for TCLK\_R/ $\overline{F}$ = High Stop Start Start D<sub>00</sub> - D<sub>09</sub> SYMBOL N-1 Stop | D00 - D09 SYMBOL N Bit Bit Bit | Figure 12. Serializer Delay



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## timing diagrams and test circuits (continued)

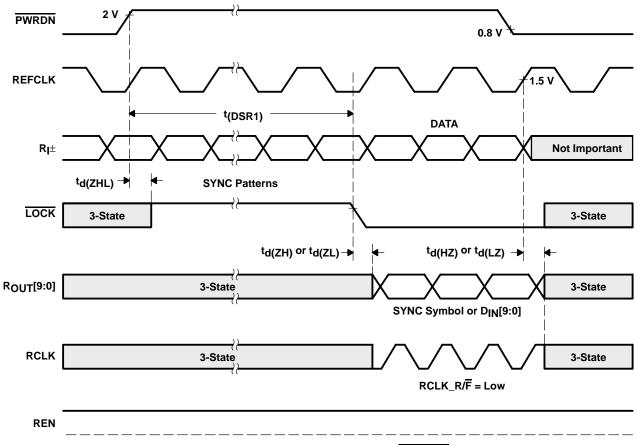


Figure 16. Deserializer PLL Lock Times and PWRDN 3-State Delays



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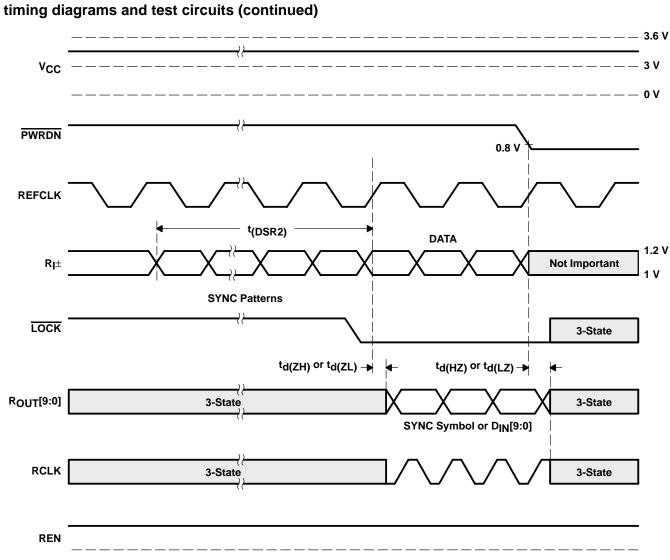
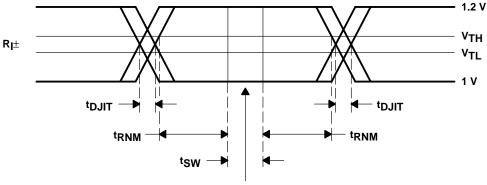


Figure 17. Deserializer PLL Lock Time From SyncPAT



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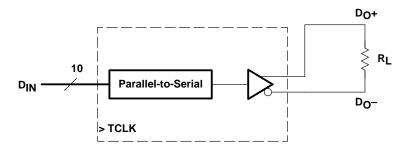
## timing diagrams and test circuits (continued)



**Ideal Sampling Position** 

t<sub>SW</sub>: Setup and Hold Time (Internal Data Sampling Window) t<sub>DJIT</sub>: Serializer Output Bit Position Jitter That Results From Jitter on TCLK t<sub>RNM</sub>: Receiver Noise Margin Time





 $V_{OD} = (D_O+) - (D_O-)$ Differential Output Signal Is Shown as  $(D_O+) - (D_O-)$ 

Figure 19. V<sub>OD</sub> Diagram



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## APPLICATION INFORMATION

## differential traces and termination

The performance of the SN65LV1023A/SN65LV1224A is affected by the characteristics of the transmission medium. Use controlled-impedance media and termination at the receiving end of the transmission line with the media's characteristics impedance.

Use balanced cables such as twisted pair or differential traces that are ran close together. A balanced cable picks up noise together and appears to the receiver as common mode. Differential receivers reject common-mode noise. Keep cables or traces matched in length to help reduce skew.

Running the differential traces close together helps cancel the external magnetic field, as well as maintain a constant impedance. Avoiding sharp turns and reducing the number of vias also helps.

## topologies

There are several topologies that the serializers can operate. Three common examples are shown below.

Figure 20 shows an example of a single-terminated point-to-point connection. Here a single termination resistor is located at the deserializer end. The resistor value should match that of the characteristic impedance of the cable or PC board traces. The total load seen by the serializer is 100  $\Omega$ . Double termination can be used and typically reduces reflections compared with single termination. However, it also reduces the differential output voltage swing.

AC-coupling is only recommended if the parallel TX data stream is encoded to achieve a dc-balanced data stream. Otherwise the AC-caps can induce common mode voltage drift due to the dc-unbalanced data stream.

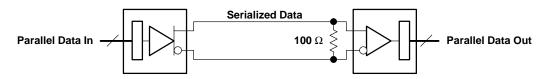


Figure 20. Single-Terminated Point-to-Point Connection

Figure 21 shows an example of a multidrop configuration. Here there is one transmitter broadcasting data to multiple receivers. A  $50-k\Omega$  resistor at the far end terminates the bus.

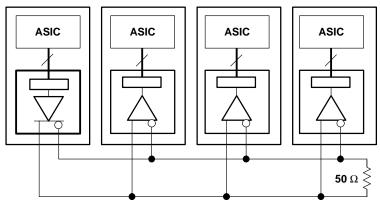


Figure 21. Multidrop Configuration



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Figure 22 shows an example of multiple serializers and deserializers on the same differential bus, such as in a backplane. This is a multipoint configuration. In this situation, the characteristic impedance of the bus can be significantly less due to loading. Termination resistors that match the loaded characteristic impedance are required at each end of the bus. The total load seen by the serializer in this example is 27  $\Omega$ .

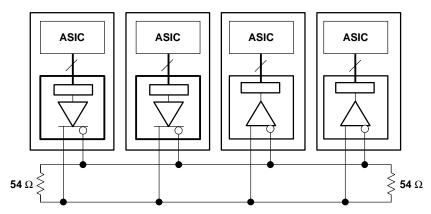


Figure 22. Multiple Serializers and Deserializers on the Same Differential Bus



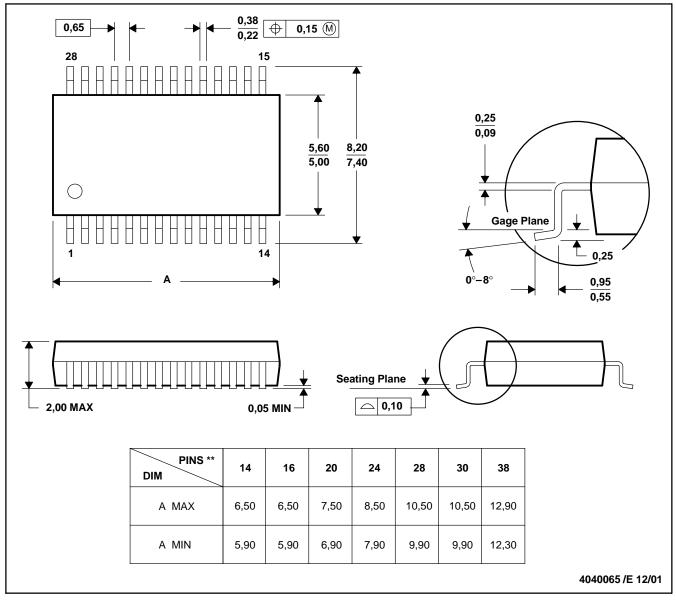
# **MECHANICAL DATA**

MSSO002E - JANUARY 1995 - REVISED DECEMBER 2001

## DB (R-PDSO-G\*\*)

PLASTIC SMALL-OUTLINE

28 PINS SHOWN



NOTES: A. All linear dimensions are in millimeters.

- B. This drawing is subject to change without notice.
- C. Body dimensions do not include mold flash or protrusion not to exceed 0,15.
- D. Falls within JEDEC MO-150



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