

#### LT1785/LT1785A LT1791/LT1791A

## 60V Fault Protected RS485/RS422 Transceivers

#### FEATURES

- Protected from Overvoltage Line Faults to ±60V
- Pin Compatible with LTC485 and LTC491
- High Input Impedance Supports Up to 128 Nodes
- No Damage or Latchup to ESD IEC-1000-4-2 Level 4: ±15kV Air Discharge IEC-1000-4-2 Level 2: ±4kV Contact Discharge
- Controlled Slew Rates for EMI Emissions Control
- Guaranteed High Receiver Output State for Floating, Shorted or Inactive Inputs
- Outputs Assume a High Impedance When Off or Powered Down
- Drives Low Cost, Low Impedance Cables
- Short-Circuit Protection on All Outputs
- Thermal Shutdown Protection

### **APPLICATIONS**

Industrial Control Data Networks

TYPICAL APPLICATION

- CAN Bus Applications
- HVAC Controls

#### DESCRIPTION

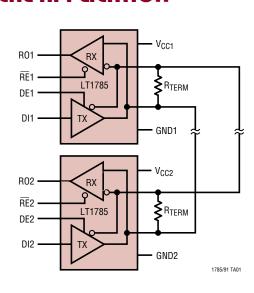
The LT<sup>®</sup>1785/LT1791 are half-duplex and full-duplex differential bus transceivers for RS485 and RS422 applications which feature on-chip protection from overvoltage faults on the data transmission lines. Receiver input and driver output pins can withstand voltage faults up to  $\pm$ 60V with respect to ground with no damage to the device. Faults may occur while the transceiver is active, shut down or powered off.

Data rates to 250kbaud on networks of up to 128 nodes are supported. Controlled slew rates on the driver outputs control EMI emissions and improve data transmission integrity on improperly terminated lines. Drivers are specified to operate with inexpensive cables as low as  $72\Omega$  characteristic impedance.

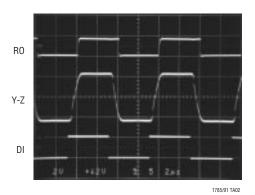
The LT1785A/LT1791A devices have "fail-safe" receiver inputs to guarantee a receiver output high for shorted, open or inactive data lines. On-chip ESD protection eliminates need for external protection devices.

The LT1785/LT1785A are available in 8-lead DIP and SO packages and the LT1791/LT1791A in 14-lead DIP and SO packages.

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#### Normal Operation Waveforms at 250kBaud



**LINEAR** TECHNOLOGY

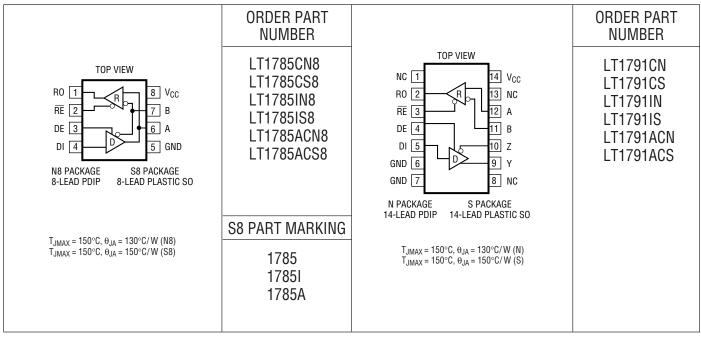
## 1

#### ABSOLUTE MAXIMUM RATINGS (Note 1)

Supply Voltage (V <sub>CC</sub> )	18V
Receiver Enable Input Voltage	0.3V to 6V
Driver Enable Input Voltage	0.3V to 6V
Driver Input Voltage	0.3V to 18V
Receiver Input Voltage	60V to 60V
Driver Output Voltage	– 60V to 60V
Receiver Output Voltage	-0.3V to (V <sub>CC</sub> + 6V)

Operating Temperature Range	
LT1785C/LT1791C/	
LT1785AC/LT1791AC	0°C to 70°C
LT1785I/LT1791I	40°C to 85°C
Storage Temperature Range	-65°C to 150°C
Lead Temperature (Soldering, 10 sec).	300°C

# PACKAGE/ORDER INFORMATION



Consult factory for Military grade parts.



#### DC ELECTRICAL CHARACTERISTICS

The  $\bullet$  denotes specifications which apply over the full operating temperature range, otherwise specifications are T<sub>A</sub> = 25°C, V<sub>CC</sub> = 5V.

SYMBOL	PARAMETER	CONDITIONS		MIN	ТҮР	MAX	UNITS
V <sub>OD1</sub>	Differential Driver Output Voltage (Unloaded)	I <sub>0</sub> = 0	•		4.1	5	V
V <sub>OD2</sub>	Differential Driver Output Voltage (With Load)	$ \begin{array}{l} R=50\Omega \;(RS422), \mbox{ Figure 1} \\ R=27\Omega \;(RS485), \mbox{ Figure 1} \\ R=18\Omega \end{array} $	•	2.0 1.5 1.2	2.70 2.45 2.2		V V V
$\Delta V_{OD}$	Change in Magnitude of Driver Differential Output Voltage for Complementary Output States	$R=27\Omega$ or $R=50\Omega,$ Figure 1	•			0.2	V
V <sub>OC</sub>	Driver Common Mode Output Voltage	$R = 27\Omega$ or $R = 50\Omega$ , Figure 1		2	2.5	3	V
$\Delta  V_{0C} $	Change in Magnitude of Driver Common Mode Output Voltage for Complementary Output States	$R=27\Omega$ or $R=50\Omega,$ Figure 1	•			0.2	V
V <sub>IH</sub>	Input High Voltage	DI, DE, RE	•	2			V
V <sub>IL</sub>	Input Low Voltage	DI, DE, RE	•			0.8	V
I <sub>IN1</sub>	Input Current	DI, DE, RE	•			5	μA
I <sub>IN2</sub>	Input Current (A, B); (LT1791 or LT1785 with DE = 0V)	$V_{IN} = 12V$ $V_{IN} = -7V$ $-60V \le V_{IN} \le 60V$	•	-0.15 -6	0.15 -0.08	0.3 6	mA mA mA
V <sub>TH</sub>	Differential Input Threshold Voltage for Receiver	$\begin{array}{c} LT1785/LT1791: -7V \leq V_{CM} \leq 12V \\ LT1785A/LT1791A: -7V \leq V_{CM} \leq 12V \end{array}$	•	-0.2 -0.2		0.2 0	V V
$\Delta V_{TH}$	Receiver Input Hysteresis	-7V < V <sub>CM</sub> < 12V			20		mV
V <sub>OH</sub>	Receiver Output High Voltage	$I_0 = -400 \mu A, V_{1D} = 200 m V$	•	3.5	4		V
V <sub>OL</sub>	Receiver Output Low Voltage	$I_0 = 1.6 \text{mA}, V_{\text{ID}} = -200 \text{mV}$	•		0.3	0.5	V
	Three-State (High Impedance) Output Current at Receiver 0V < V <sub>OUT</sub> < 6V	$\overline{\text{RE}}$ > 2V or Power Off	•	-1		1	μA
R <sub>IN</sub>	Receiver Input Resistance (LT1791)	$\begin{array}{l} -7V \leq V_{CM} \leq 12V \\ -60V \leq V_{CM} \leq 60V \end{array}$	•	85	125 125		kΩ kΩ
	LT1785	$-7V \le V_{CM} \le 12V$	•	50	90		kΩ
	RS485 Unit Load					0.25	
I <sub>SC</sub>	Driver Short-Circuit Current	$V_{OUT} = HIGH$ , Force $V_0 = -7V$ $V_{OUT} = LOW$ , Force $V_0 = 12V$	•	35 35		250 250	mA mA
	Driver Output Fault Current	$V_0 = 60V$ $V_0 = -60V$	•	-6		6	mA mA
	Receiver Short-Circuit Current	$0V \le V_0 \le V_{CC}$	•			±30	mA
	Driver Three-State Output Current	$-7V \le V_0 \le 12V$ $-60V \le V_0 \le 60V$	•	-0.2 -6		0.3 6	mA mA
I <sub>CC</sub>	Supply Current	No Load, $\overline{RE} = 0V$ , $DE = 5V$ No Load, $\overline{RE} = 5V$ , $DE = 5V$ No Load, $\overline{RE} = 0V$ , $DE = 0V$ No Load, $\overline{RE} = 5V$ , $DE = 0V$	• • •		5.5 5.5 4.5 0.2	9 9 8 0.3	mA mA mA mA



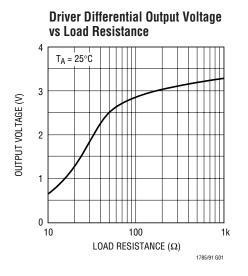
## **SWITCHING CHARACTERISTICS**

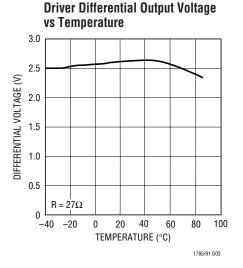
The  $\bullet$  denotes specifications which apply over the full operating temperature range, otherwise specifications are T<sub>A</sub> = 25°C, V<sub>CC</sub> = 5V.

SYMBOL	PARAMETER	CONDITIONS		MIN	ТҮР	MAX	UNITS
t <sub>PLH</sub>	Driver Input to Output	Figures 3, 5	•		700	2000	ns
t <sub>PHL</sub>	Driver Input to Output	Figures 3, 5	•		700	2000	ns
t <sub>SKEW</sub>	Driver Output to Output	Figures 3, 5			100		ns
t <sub>r</sub> , t <sub>f</sub>	Driver Rise or Fall Time	Figures 3, 5	•	200	800	2000	ns
t <sub>ZH</sub>	Driver Enable to Output High	Figures 4, 6	•		500	3000	ns
t <sub>ZL</sub>	Driver Enable to Output Low	Figures 4, 6	•		800	3000	ns
t <sub>LZ</sub>	Driver Disable Time from Low	Figures 4, 6	•		200	5000	ns
t <sub>HZ</sub>	Driver Disable Time from High	Figures 4, 6	•		800	5000	ns
t <sub>PLH</sub>	Receiver Input to Output	Figures 3, 7	•		400	900	ns
t <sub>PHL</sub>	Receiver Input to Output	Figures 3, 7	•		400	900	ns
t <sub>SKD</sub>	Differential Receiver Skew				200		ns
t <sub>ZL</sub>	Receiver Enable to Output Low	Figures 2, 8	•		300	1000	ns
t <sub>ZH</sub>	Receiver Enable to Output High	Figures 2, 8	•		300	1000	ns
t <sub>LZ</sub>	Receiver Disable from Low	Figures 2, 8	•		400	1000	ns
t <sub>HZ</sub>	Receiver Disable from High	Figures 2, 8	•		400	1000	ns
f <sub>MAX</sub>	Maximum Data Rate		•	250			kbps
t <sub>SHDN</sub>	Time to Shut Down	Figures 2, 6, 8			3		μs
t <sub>ZH(SHDN)</sub>	Driver Enable from Shutdown to Output High	Figures 2, 6; $\overline{\text{RE}}$ = 5V			12		μs
t <sub>ZL(SHDN)</sub>	Driver Enable from Shutdown to Output Low	Figures 2, 6; $\overline{\text{RE}}$ = 5V			12		μs
t <sub>ZH(SHDN)</sub>	Receiver Enable from Shutdown to Output High	Figures 2, 8; DE = 0V			4		μs
t <sub>ZL(SHDN)</sub>	Receiver Enable from Shutdown to Output Low	Figures 2, 8; DE = 0V			4		μs

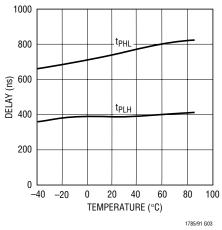
**Note 1:** Absolute Maximum Ratings are those values beyond which the life of a device may be impaired.

# **TYPICAL PERFORMANCE CHARACTERISTICS**



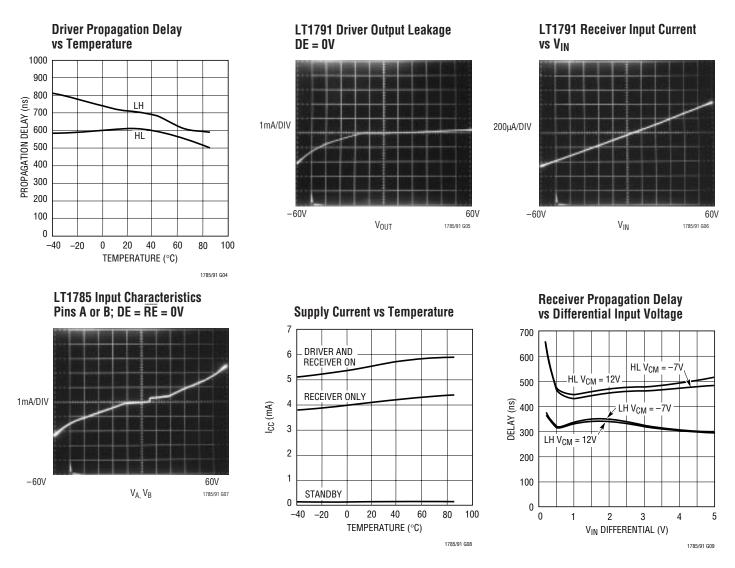


#### Receiver Propagation Delay vs Temperature





# TYPICAL PERFORMANCE CHARACTERISTICS



### PIN FUNCTIONS

**R0:** Receiver Output. TTL level logic output. If the receiver is active ( $\overline{RE}$  pin low), RO is high if receiver input A  $\geq$  B by 200mV. If A  $\leq$  B by 200mV, then RO will be low. RO assumes a high impedance output state when  $\overline{RE}$  is high or the part is powered off. RO is protected from output shorts from ground to 6V.

**RE:** Receiver Output Enable. TTL level logic input. A logic low on RE enables normal operation of the receiver output RO. A logic high level at RE places the receiver output pin RO into a high impedance state. If receiver enable RE and driver enable DE are both in the disable state, the circuit

goes to a low power shutdown state. Placing either  $\overline{\text{RE}}$  or DE into its active state brings the circuit out of shutdown. Shutdown state is not entered until a  $3\mu$ s delay after both  $\overline{\text{RE}}$  and DE are disabled, allowing for logic skews in toggling between transmit and receive modes of operation. For CAN bus applications,  $\overline{\text{RE}}$  should be tied low to prevent the circuit from entering shutdown.

**DE:** Driver Output Enable. TTL level logic input. A logic high on DE enables normal operation of the driver outputs (Y and Z on LT1791, A and B on LT1785). A logic low level at DE places the driver output pins into a high impedance



### PIN FUNCTIONS

state. If receiver enable  $\overline{\text{RE}}$  and driver enable DE are both in the disable state, the circuit goes to a low power shutdown state. Placing either  $\overline{\text{RE}}$  or DE into its active state brings the circuit out of shutdown. Shutdown state is not entered until a 3µs delay after both  $\overline{\text{RE}}$  and DE are disabled, allowing for logic skews in toggling between transmit and receive modes of operation. For CAN bus operation the DE pin is used for signal input to place the data bus in dominant or recessive states.

**DI:** Driver Input. TTL level logic input. A logic high at DI causes driver output A or Y to a high state, and output B or Z to a low state. Complementary output states occur for DI low. For CAN bus applications DI should be tied low.

#### GND: Ground.

**Y:** Driver Output. The Y driver output is in phase with the driver input DI. In the LT1785 driver output Y is internally connected to receiver input A. The driver output assumes a high impedance state when DE is low, power is off or thermal shutdown is activated. The driver output is protected from shorts between  $\pm$ 60V in both active and high impedance modes. For CAN applications, output Y is the CANL output node.

**Z:** Driver Output. The Z driver output is opposite in phase to the driver input DI. In the LT1785 driver output Z is internally connected to receiver input B. The driver output assumes a high impedance state when DE is low, power is off or thermal shutdown is activated. The driver output is

protected from shorts between  $\pm 60V$  in both active and high impedance modes. For CAN applications, output Z is the CANH output node.

A: Receiver Input. The A receiver input forces a high receiver output when  $V(A) \ge [V(B) + 200mV]$ .  $V(A) \le [V(B) - 200mV]$  forces a receiver output low. Receiver inputs A and B are protected against voltage faults between  $\pm 60V$ . The high input impedance allows up to 128 LT1785 or LT1791 transceivers on one RS485 data bus.

The LT1785A/LT1791A have guaranteed receiver input thresholds  $-200mV < V_{TH} < 0$ . Receiver outputs are guaranteed to be in a high state for 0V inputs.

**B**: Receiver Input. The B receiver input forces a high receiver output when  $V(A) \ge [V(B) + 200mV]$ . When  $V(A) \le [V(B) - 200mV]$ , the B receiver forces a receiver output low. Receiver inputs A and B are protected against voltage faults between  $\pm 60V$ . The high input impedance allows up to 128 LT1785 or LT1791 transceivers on one RS485 data bus.

The LT1785A/LT1791A have guaranteed receiver input thresholds  $-200mV < V_{TH} < 0$ . Receiver outputs are guaranteed to be in a high state for 0V inputs.

 $V_{CC}$ : Positive Supply Input. For RS422 or RS485 operation, 4.75V  $\leq V_{CC} \leq 5.25$ V. Higher  $V_{CC}$  input voltages increase output drive swing.  $V_{CC}$  should be decoupled with a 0.1  $\mu$ F low ESR capacitor directly at Pin 8 (V\_{CC}).

# TEST CIRCUITS

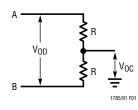


Figure 1. Driver DC Test Load

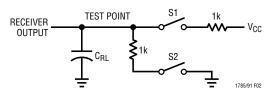


Figure 2. Receiver Timing Test Load



### **TEST CIRCUITS**

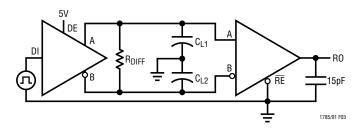


Figure 3. Driver/Receiver Timing Test Circuit

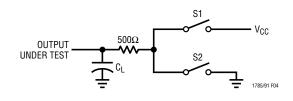


Figure 4. Driver Timing Test Load

### FUNCTION TABLES

LT1785 Transmitting						
INPUTS			OUTPUTS			
RE	DE	DI	Α	В	RO	
0	1	0	0	1	0	
0	1	1	1	0	1	
1	0	Х	Hi-Z	Hi-Z	Hi-Z	
1	1	0	0	1	Hi-Z	
1	1	1	1	0	Hi-Z	

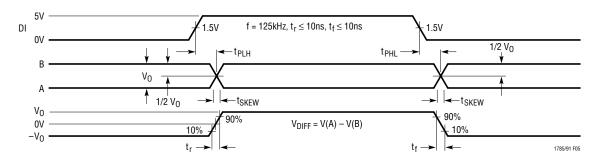
#### LT1785 Receiving

INPUTS					
DE	DI	A-B	RO		
0	Х	≤-200mV	0		
0	Х	≥200mV*	1		
0	Х	Open	1		
0	Х	Х	Hi-Z		
	INP	INPUTS	INPUTSDEDIA-B0X $\leq -200$ mV0X $\geq 200$ mV*		

	I	NPUTS	C	UTPUTS		
RE	DE	DI	A-B	Y	Z	RO
0	0	Х	≤-200mV	Hi-Z	Hi-Z	0
0	0	Х	≥200mV*	Hi-Z	Hi-Z	1
0	0	Х	Open	Hi-Z	Hi-Z	1
0	1	0	$\leq$ -200mV	0	1	0
0	1	0	≥200mV*	0	1	1
0	1	0	Open	0	1	1
0	1	1	≤-200mV	1	0	0
0	1	1	≥200mV*	1	0	1
0	1	1	Open	1	0	1
1	0	Х	Х	Hi-Z	Hi-Z	Hi-Z
1	1	0	Х	0	1	Hi-Z
1	1	1	Х	1	0	Hi-Z
* ≥ 0m\	for LT17	91A				•

 $^{\star} \geq 0mV$  for LT1785A

# SWITCHING TIME WAVEFORMS

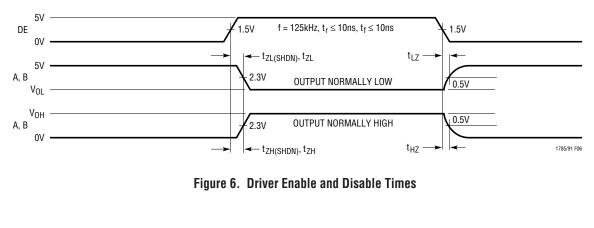


LT1791





## SWITCHING TIME WAVEFORMS



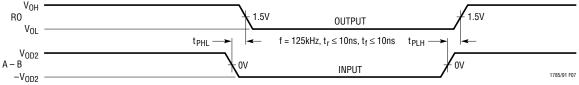


Figure 7. Receiver Propagation Delays

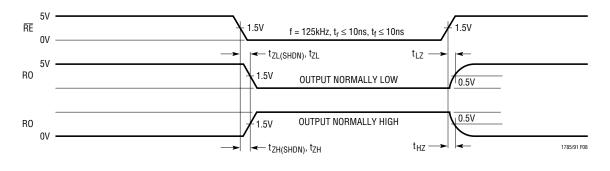


Figure 8. Receiver Enable and Disable Times

# **APPLICATIONS INFORMATION**

#### **Overvoltage Protection**

The LT1785/LT1791 RS485/RS422 transceivers answer an applications need for overvoltage fault tolerance on data networks. Industrial installations may encounter common mode voltages between nodes far greater than the –7V to 12V range specified for compliance to RS485 standards. CMOS RS485 transceivers can be damaged by voltages above their absolute maximum ratings of typically -8V to 12.5V. Replacement of standard RS485 transceiver components with the LT1785 or LT1791 devices eliminates field failures due to overvoltage faults or the use of costly external protection devices. The limited overvoltage tolerance of CMOS RS485 transceivers makes implementation of effective external protection networks difficult without interfering with proper data network performance within the -7V to 12V region of RS485 operation.



## **APPLICATIONS INFORMATION**

The high overvoltage rating of the LT1785/LT1791 facilitates easy extension to almost any level. Simple discrete component networks that limit the receiver input and driver output voltages to less than  $\pm$ 60V can be added to the device to extend protection to any desired level. Figure 11 shows a protection network against faults to the 120VAC line voltage.

The LT1785/LT1791 protection is achieved by using a high voltage bipolar integrated circuit process for the transceivers. The naturally high breakdown voltages of the bipolar process provides protection in powered-off and high impedance conditions. The driver outputs use a foldback current limit design to protect against overvoltage faults while still allowing high current output drive.

#### **ESD** Protection

The LT1785/LT1791 I/O pins have on-chip ESD protection circuitry to eliminate field failures caused by discharges to exposed ports and cables in application environments. The LT1785 pins A and B and the LT1791 driver output pins Y and Z are protected to IEC-1000-4-2 level 2. These pins will survive multiple ESD strikes of  $\pm$ 15kV air discharge or  $\pm$ 4kV contact discharge. Due to their very high input impedance, the LT1791 receiver pins are protected to IEC-1000-4-2 level 2, or  $\pm$ 15kV air and  $\pm$ 4kV contact discharges. This level of ESD protection will guarantee immunity from field failures in all but the most severe ESD environments. The LT1791 receiver input ESD tolerance may be increased to IEC level 4 compliance by adding 2.2k resistors in series with these pins.

#### Low Power Shutdown

The LT1785/LT1791 have  $\overline{RE}$  and DE logic inputs to control the receive and transmit modes of the transceivers. The  $\overline{RE}$  input allows normal data reception when in the low state. The receiver output goes to a high impedance state when  $\overline{RE}$  is high, allowing multiplexing the RO data

line. The DE logic input performs a similar function on the driver outputs. A high state on DE activates the differential driver outputs, a low state places both driver outputs into high impedance. Tying the RE and DE logic inputs together may be done to allow one logic signal to toggle the transceiver from receive to transmit modes. The DE input is used as the data input in CAN bus applications.

Disabling both the driver and receiver places the device into a low supply current shutdown mode. An internal time delay of  $3\mu$ s minimum prevents entering shutdown due to small logic skews when a toggle between receive and transmit is desired. The recovery time from shutdown mode is typically  $12\mu$ s. The user must be careful to allow for this wake-up delay from shutdown mode. To allow full 250kbaud data rate transmission in CAN applications, the RE pin should be tied low to prevent entering shutdown mode.

#### Slew Limiting for EMI Emissions Control

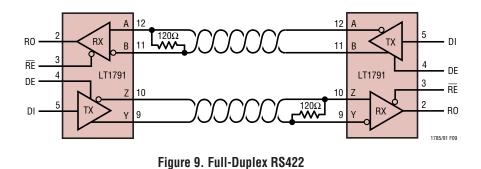
The LT1785/LT1791 feature controlled driver output slew rates to control high frequency EMI emissions from equipment and data cables. The slew limiting limits data rate operation to 250kbaud. Slew limiting also mitigates the adverse affects of imperfect transmission line termination caused by stubs or mismatched cable. In some low speed, short distance networks, cable termination may be eliminated completely with no adverse effect on data transmission.

#### Data Network Cable Selection and Termination

Long distance data networks operating at high data transmission rates should use high quality, low attenuation cable with well-matched cable terminations. Short distance networks at low data rates may use much less expensive PVC cable. These cables have characteristic impedances as low as  $72\Omega$ . The LT1785/LT1791 output drivers are guaranteed to drive cables as low as  $72\Omega$ .



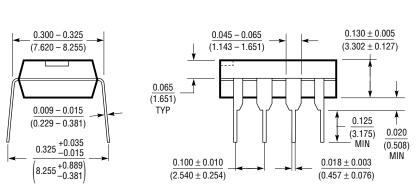
## **APPLICATIONS INFORMATION**

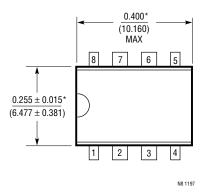


N8 Package 8-Lead PDIP (Narrow 0.300) (LTC DWG # 05-08-1510)

#### PACKAGE DESCRIPTION

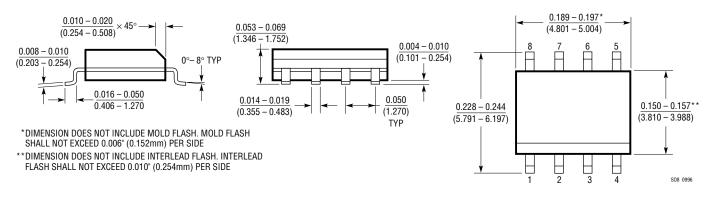
Dimensions in inches (millimeters) unless otherwise noted.





\*THESE DIMENSIONS DO NOT INCLUDE MOLD FLASH OR PROTRUSIONS. MOLD FLASH OR PROTRUSIONS SHALL NOT EXCEED 0.010 INCH (0.254mm)

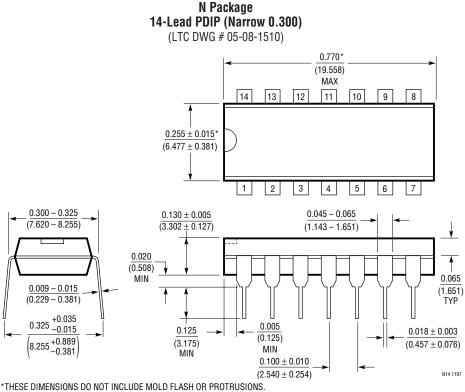






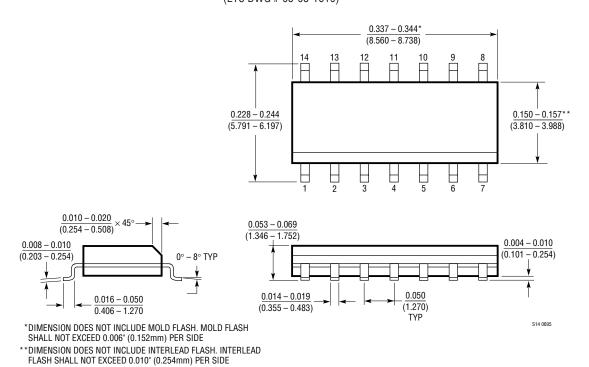
#### PACKAGE DESCRIPTION

Dimensions in inches (millimeters) unless otherwise noted.



MOLD FLASH OR PROTRUSIONS SHALL NOT EXCEED 0.010 INCH (0.254mm)

S Package 14-Lead Plastic Small Outline (Narrow 0.150) (LTC DWG # 05-08-1610)





Information furnished by Linear Technology Corporation is believed to be accurate and reliable. However, no responsibility is assumed for its use. Linear Technology Corporation makes no representation that the interconnection of circuits as described herein will not infringe on existing patent rights.

### TYPICAL APPLICATIONS

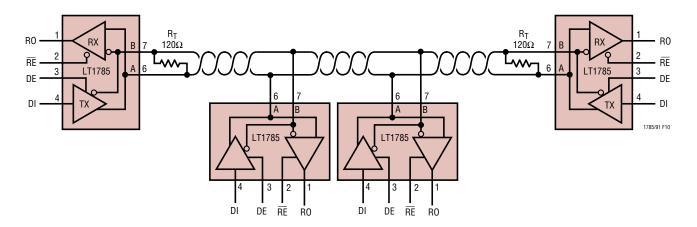
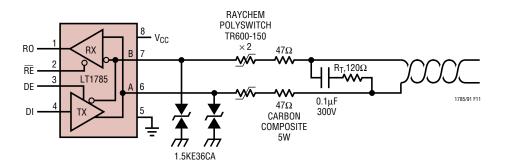
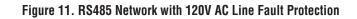


Figure 10. Half-Duplex RS485 Network Operation





### **RELATED PARTS**

PART NUMBER	DESCRIPTION	COMMENTS
LTC485	Low Power RS485 Interface Transceiver	I <sub>CC</sub> = 300μA (Typ)
LTC491	Differential Driver and Receiver Pair	I <sub>CC</sub> = 300μA
LTC1483	Ultralow Power RS485 Low EMI Transceiver	Controlled Driver Slew Rate
LTC1485	Differential Bus Transceiver	10Mbaud Operation
LTC1487	Ultralow Power RS485 with Low EMI, Shutdown and High Input Impedance	Up to 256 Transceivers on the Bus
LTC1520	50Mbps Precision Quad Line Receiver	Channel-to-Channel Skew 400ps (Typ)
LTC1535	Isolated RS485 Full-Duplex Transceiver	2500V <sub>RMS</sub> Isolation in Surface Mount Package
LTC1685	52Mbps RS485 Half-Duplex Transceiver	Propagation Delay Skew 500ps (Typ)
LTC1687	52Mbps RS485 Full-Duplex Transceiver Propagation Delay Skew 500ps	