## 100MHz 100mA VCoM Amplifiers

The EL9211, EL9212, and EL9214 feature 1, 2, and 4 channel high power output amplifiers. They are designed primarily for generation of $\mathrm{V}_{\text {COM }}$ voltages in TFT-LCD applications. Each amplifier features a -3dB bandwidth of 130 MHz with slew rates of $115 \mathrm{~V} / \mu \mathrm{s}$. Each device comes in a thermal package and can drive 300 mA peak per output.
All units are available in Pb -free packaging only and are specified for operation over the $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ temperature range.

## Ordering Information

| PART NUMBER <br> (See Note) | PACKAGE <br> (Pb-Free) |  <br> REEL | PKG. <br> DWG. \# |
| :--- | :---: | :---: | :---: |
| EL9211IWZ-T7 | 5-Pin SOT-23 | 7" (3K pcs) | MDP0038 |
| EL9211IWZ-T7A | 5-Pin SOT-23 | 7" (250 pcs) | MDP0038 |
| EL9211IYEZ | 8-Pin HMSOP | - | MDP0050 |
| EL9211IYEZ-T7 | 8-Pin HMSOP | $7 "$ | MDP0050 |
| EL9211IYEZ-T13 | 8-Pin HMSOP | $13 "$ | MDP0050 |
| EL9212IYEZ | 8-Pin HMSOP | - | MDP0050 |
| EL9212IYEZ-T7 | 8-Pin HMSOP | $7 "$ | MDP0050 |
| EL9212IYEZ-T13 | 8-Pin HMSOP | $13 "$ | MDP0050 |
| EL9214IREZ | 14-Pin HTSSOP | - | MDP0048 |
| EL9214IREZ-T7 | 14-Pin HTSSOP | $7 "$ | MDP0048 |
| EL9214IREZ-T13 | 14-Pin HTSSOP | $13 " ~$ | MDP0048 |

NOTE: Intersil Pb -free products employ special Pb -free material sets; molding compounds/die attach materials and $100 \%$ matte tin plate termination finish, which are RoHS compliant and compatible with both SnPb and Pb -free soldering operations. Intersil Pb -free products are MSL classified at Pb -free peak reflow temperatures that meet or exceed the Pb-free requirements of IPC/JEDEC J STD-020C.

## Features

- 1,2 , and 4 channel versions
- $130 \mathrm{MHz}-3 \mathrm{~dB}$ bandwidth
- $115 \mathrm{~V} / \mu \mathrm{s}$ slew rate
- 300mA peak output current
- Supply voltage from 5 V to 13.5 V
- Low supply current - <2.4mA per channel
- Pb-free available (RoHS compliant)


## Applications

- TFT-LCD $V_{C O M}$ supply
- Electronics notebooks
- Computer monitors
- Electronics games
- Touch-screen displays
- Portable instrumentation


## Pinouts



EL9211
(8-PIN HMSOP)
TOP VIEW


EL9214
(14-PIN HTSSOP) TOP VIEW


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Absolute Maximum Ratings \(\left(\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}\right)\)
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Supply Voltage between $\mathrm{V}_{\mathrm{S}^{+}}$and $\mathrm{V}_{\mathrm{S}^{-}} \ldots . .$. Input Voltage . . . . . . . . . . . . . . . . . . . . . . . . . . . V $\mathrm{V}_{\mathrm{S}}-0.5 \mathrm{~V}, \mathrm{~V}_{\mathrm{S}}+0.5 \mathrm{~V}$ Maximum Continuous Output Current . . . . . . . . . . . . . . . . . . 100 mA Ambient Operating Temperature . . . . . . . . . . . . . . . . $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$

Power Dissipation . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . See Curves
Maximum Die Temperature . . . . . . . . . . . . . . . . . . . . . . . . . . $+125^{\circ} \mathrm{C}$
Storage Temperature . . . . . . . . . . . . . . . . . . . . . . . . $-65^{\circ} \mathrm{C}$ to $+150^{\circ} \mathrm{C}$

CAUTION: Stresses above those listed in "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress only rating and operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied.

IMPORTANT NOTE: All parameters having Min/Max specifications are guaranteed. Typ values are for information purposes only. Unless otherwise noted, all tests are at the specified temperature and are pulsed tests, therefore: $T_{J}=T_{C}=T_{A}$

Electrical Specifications $\quad V_{S^{+}}=+6 \mathrm{~V}, \mathrm{~V}_{\mathrm{S}^{-}}=-6 \mathrm{~V}, \mathrm{R}_{\mathrm{L}}=10 \mathrm{k} \Omega, \mathrm{R}_{\mathrm{F}}=0 \Omega, \mathrm{C}_{\mathrm{L}}=10 \mathrm{pF}$ to 0 V , Gain $=-1, \mathrm{~T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$, unless otherwise specified.

| PARAMETER | DESCRIPTION | CONDITIONS | MIN | TYP | MAX | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| INPUT CHARACTERISTICS |  |  |  |  |  |  |
| V OS | Input Offset Voltage | $\mathrm{V}_{\mathrm{CM}}=6 \mathrm{~V}$ | -6 | -1 | +2 | mV |
| TCV ${ }_{\text {OS }}$ | Average Offset Voltage Drift | (Note) |  | 10 |  | $\mu \mathrm{V} /{ }^{\circ} \mathrm{C}$ |
| $\mathrm{I}_{\mathrm{B}}$ | Input Bias Current | $\mathrm{V}_{\mathrm{CM}}=6 \mathrm{~V}$ | -1.4 |  | -0.4 | $\mu \mathrm{A}$ |
| $\mathrm{R}_{\mathrm{IN}}$ | Input Impedance |  |  | 1 |  | G $\Omega$ |
| $\mathrm{C}_{\mathrm{IN}}$ | Input Capacitance |  |  | 1.35 |  | pF |
| $\mathrm{V}_{\text {REG }}$ | Load Regulation | $\mathrm{V}_{\mathrm{COM}}=6 \mathrm{~V},-100 \mathrm{~mA}<\mathrm{I}_{\mathrm{L}}<100 \mathrm{~mA}$ | -20 |  | +20 | mV |
| CMIR | Common Mode Input Range |  | -0.5 |  | +12.5 | V |
| CMRR | Common Mode Rejection Ratio | For $\mathrm{V}_{\mathrm{IN}}$ from -0.5 to +12.5 V | 75 | 100 |  | dB |
| AVOL | Open Loop Gain |  | 55 | 70 |  | dB |
| OUTPUT CHARACTERISTICS |  |  |  |  |  |  |
| V ${ }_{\text {OL }}$ | Output Swing Low | $\mathrm{I}_{\mathrm{L}}=-5 \mathrm{~mA}$ |  | 0.9 | 1.1 | V |
| $\mathrm{V}_{\mathrm{OH}}$ | Output Swing High | $\mathrm{I}_{\mathrm{L}}=+5 \mathrm{~mA}$ | 10.7 | 10.94 |  | V |
| ISC | Short Circuit Current |  |  | 300 |  | mA |
| POWER SUPPLY PERFORMANCE |  |  |  |  |  |  |
| PSRR | Power Supply Rejection Ratio | $\mathrm{V}_{\mathrm{S}}$ from 4.5 V to 10.5 V | 50 | 75 |  | dB |
| Is | Total Supply Current | EL9211 (no load) |  | 2.3 | 2.9 | mA |
|  |  | EL9212 (no load) |  | 4.5 | 5 | mA |
|  |  | EL9214 (no load) |  | 8.8 | 9.6 | mA |
| DYNAMIC PERFORMANCE |  |  |  |  |  |  |
| SR | Slew Rate (Note) | 2V step, 20\% to 80\% | 90 | 115 |  | $\mathrm{V} / \mu \mathrm{s}$ |
| $\mathrm{t}_{5}$ | Settling to $+0.1 \%\left(A_{V}=-1\right)$ | $\left(A_{V}=-1\right), V_{O}=2 \mathrm{~V}$ step |  | 30 |  | ns |
| BW | -3dB Bandwidth | $\mathrm{R}_{\mathrm{L}}=10 \mathrm{k} \Omega, \mathrm{C}_{\mathrm{L}}=10 \mathrm{pF}, A_{V}=+1$ |  | 130 |  | MHz |
|  |  | $R_{L}=10 \mathrm{k} \Omega, C_{L}=10 \mathrm{pF}, A_{V}=-1$ |  | 52 |  | MHz |
| GBWP | Gain-Bandwidth Product | $\mathrm{R}_{\mathrm{L}}=10 \mathrm{k} \Omega, \mathrm{C}_{\mathrm{L}}=10 \mathrm{pF}$ |  | 63 |  | MHz |
| PM | Phase Margin | $\mathrm{R}_{\mathrm{L}}=10 \mathrm{k} \Omega, \mathrm{C}_{\mathrm{L}}=10 \mathrm{pF}$ |  | 43 |  | - |

NOTE: Slew rate is measured on rising and falling edges.

## Typical Performance Curves



FIGURE 1. FREQUENCY RESPONSE FOR VARIOUS $R_{L}$


FIGURE 3. OPEN LOOP GAIN AND PHASE vs FREQUENCY


FIGURE 5. PSRR


FIGURE 2. FREQUENCY RESPONSE FOR VARIOUS $C_{L}$


FIGURE 4. CLOSED LOOP OUTPUT IMPEDANCE vs FREQUENCY


FIGURE 6. CMRR

## Typical Performance Curves (Continued)



FIGURE 7. CHANNEL SEPARATION FOR EL9212/EL9214


FIGURE 9. THD + NOISE vs FREQUENCY


FIGURE 11. SMALL SIGNAL OVERSHOOT vs LOAD CAPACITANCE


FIGURE 8. VOLTAGE NOISE vs FREQUENCY


FIGURE 10. MAXIMUM OUTPUT SWING vs FREQUENCY


FIGURE 12. VOUT $-\mathrm{V}_{\mathrm{S}}$ - vs $\mathrm{I}_{\text {SINK }}$

## Typical Performance Curves (Continued)



FIGURE 13. $\mathbf{V}_{\mathbf{S}^{+}}-\mathrm{V}_{\text {OUT }}$ vs Isource


FIGURE 15. SMALL SIGNAL TRANSIENT RESPONSE


FIGURE 17. GOING INTO SATURATION NEGATIVE EDGE


FIGURE 14. LARGE SIGNAL TRANSIENT RESPONSE


FIGURE 16. GOING INTO SATURATION POSITIVE EDGE


FIGURE 18. DELAY TIME

Typical Performance Curves (Continued)


FIGURE 19. SUPPLY CURRENT(PER AMPLIFIER) vs SUPPLY VOLTAGE


FIGURE 21. PACKAGE POWER DISSIPATION vs AMBIENT TEMPERATURE


FIGURE 23. PACKAGE POWER DISSIPATION vs AMBIENT TEMPERATURE


FIGURE 20. PACKAGE POWER DISSIPATION vs AMBIENT TEMPERATURE


FIGURE 22. PACKAGE POWER DISSIPATION vs AMBIENT TEMPERATURE

## Pin Descriptions

| $\begin{gathered} \text { EL9211 } \\ \text { (5-PIN } \\ \text { SOT-23) } \end{gathered}$ | $\begin{gathered} \text { EL9211 } \\ \text { (8-PIN } \\ \text { HMSOP) } \end{gathered}$ | $\begin{gathered} \text { EL9212 } \\ \text { (8-PIN } \\ \text { HMSOP) } \end{gathered}$ | $\begin{gathered} \text { EL9214 } \\ \text { (14-PIN } \\ \text { HTSSOP) } \end{gathered}$ | PIN NAME | FUNCTION | EQUIVALENT CIRCUIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 1 | 6 | 1 | 1 | VOUTA | Amplifier A output | CIRCUIT 1 |
| 4 | 2 | 2 | 2 | VINA- | Amplifier A inverting input |  |
| 3 | 3 | 3 | 3 | VINA+ | Amplifier A non-inverting input | (Reference Circuit 2) |
| 5 | 7 | 8 | 4 | VS+ | Positive power supply |  |
|  |  | 5 | 5 | VINB+ | Amplifier B non-inverting input | (Reference Circuit 2) |
|  |  | 6 | 6 | VINB- | Amplifier B inverting input | (Reference Circuit 2) |
|  |  | 7 | 7 | VOUTB | Amplifier B output | (Reference Circuit 1) |
|  |  |  | 8 | VOUTC | Amplifier C output | (Reference Circuit 1) |
|  |  |  | 9 | VINC- | Amplifier C inverting input | (Reference Circuit 2) |
|  |  |  | 10 | VINC+ | Amplifier C non-inverting input | (Reference Circuit 2) |
| 2 | 4 | 4 | 11 | VS- | Negative power supply |  |
|  |  |  | 12 | VIND+ | Amplifier D non-inverting input | (Reference Circuit 2) |
|  |  |  | 13 | VIND- | Amplifier D inverting input | (Reference Circuit 2) |
|  |  |  | 14 | VOUTD | Amplifier D output | (Reference Circuit 1) |
|  | 1, 5, 8 |  |  | NC | Not connected |  |

## Application Information

## Product Description

The EL9211, EL9212, and EL9214 voltage feedback amplifiers are fabricated using a high voltage CMOS process. They exhibit rail-to-rail input and output capability, are unity gain stable and have low power consumption ( 2.4 mA per amplifier). These features make the EL9211, EL9212, and EL9214 ideal for a wide range of generalpurpose applications. Connected in voltage follower mode and driving a load of 10K, the EL9211, EL9212, and EL9214 have a -3 dB bandwidth of 130 MHz while maintaining a $115 \mathrm{~V} / \mu \mathrm{s}$ slew rate. The EL9211 is a single amplifier, EL9212 is a dual amplifier, and EL9214 is a quad amplifier.

## Operating Voltage, Input, and Output

The EL9211, EL9212, and EL9214 are specified with a single nominal supply voltage from 5 V to 13.5 V or a split supply with its total range from 5 V to 13.5 V . Most EL9211, EL9212, and EL9214 specifications are stable over both the full supply range and operating temperatures of $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$. Parameter variations with operating voltage and/or temperature are shown in the typical performance curves.

## Short Circuit Current Limit

The EL9211, EL9212, and EL9214 will limit the short circuit current to 300 mA if the output is directly shorted to the positive or negative supply. If an output is shorted indefinitely, the power dissipation could easily increase such that the device may be damaged. Maximum reliability is maintained if the output continuous current never exceeds $\pm 65 \mathrm{~mA}$. This limit is set by the design of the internal metal interconnects.

## Output Phase Reversal

The EL9211, EL9212, and EL9214 are immune to phase reversal as long as the input voltage is limited from $-\mathrm{VS}-0.5 \mathrm{~V}$ to $+\mathrm{VS}+0.5 \mathrm{~V}$. Although the device's output will not change phase, the input's over-voltage should be avoided. If an input voltage exceeds supply voltage by more than 0.6 V , electrostatic protection diodes placed in the input stage of the device begin to conduct and over-voltage damage could occur.

## Unused Amplifiers

It is recommended that any unused amplifiers in a dual and quad package be configured as a unity gain follower. The inverting input should be directly connected to the output and the non-inverting input tied to the ground plane.

## Power Supply Bypassing and Printed Circuit Board Layout

The EL9211, EL9212, and EL9214 can provide gain at high frequency. As with any high-frequency device, good printed circuit board layout is necessary for optimum performance. Ground plane construction is highly recommended, lead lengths should be as short as possible and the power supply pins must be well bypassed to reduce the risk of oscillation. For normal single supply operation, where the -VS pin is connected to ground, a $0.1 \mu \mathrm{~F}$ ceramic capacitor should be placed from +VS to pin and -VS to pin. A $4.7 \mu \mathrm{~F}$ tantalum capacitor should then be connected in parallel, placed in the region of the amplifier. One $4.7 \mu \mathrm{~F}$ capacitor may be used for multiple devices. This same capacitor combination should be placed at each supply pin to ground if split supplies are to be used.

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