

General Description

The MAXQ3120 microcontroller is a high-performance, 16-bit microcontroller that incorporates dual, true-differential, 16-bit sigma-delta analog-to-digital converters (ADCs), a liquid-crystal display (LCD) interface that can drive up to 112 segments, and a real-time clock (RTC) module with a dedicated battery-backup supply. The MAXQ3120 is uniquely suited for the single-phase electricity metering market, but can be used in any application that requires high-performance operation. The device can operate at a maximum of 8MHz (DV_{DD} = 3.3V). The MAXQ3120 has 16kWords of flash memory, 256 words of RAM, three 16-bit timers, and two universal synchronous/asynchronous receiver/transmitters (USARTs). The microcontroller core and I/O are powered by a single 3.3V supply, and an additional battery supply keeps the RTC running during power outages.

Applications

Single-Phase Home Appliances **Electricity Metering** Consumer Electronics Battery-Powered and Thermostats/Humidity Portable Devices Sensors Electrochemical and Security Sensors **Optical Sensors** Gas and Chemical Industrial Control Sensors Data-Acquisition **HVAC** Systems and Data

Ordering Information

Smart Transmitters

PART	TEMP RANGE	PIN-PACKAGE
MAXQ3120-FFN	-40°C to +85°C	80 MQFP
MAXQ3120-FFN+	-40°C to +85°C	80 MQFP

⁺Denotes a Pb-free/RoHS-compliant device.

Loggers

Selector Guide, Typical Operating Circuit, and Pin Configuration appear at end of data sheet.

MAXQ is a trademark of Maxim Integrated Products, Inc.

Note: Some revisions of this device may incorporate deviations from published specifications known as errata. Multiple revisions of any device may be simultaneously available through various sales channels. For information about device errata, go to: www.maxim-ic.com/errata.

Features

♦ High-Performance, Low-Power, 16-Bit RISC Core

DC to 8MHz Operation, Approaching 1MIPS per MHz

3.3V Core and I/O

33 Instructions, Most Single-Cycle

Three Independent Data Pointers Accelerate Data Movement with Automatic Increment/

16-Level Hardware Stack

16-Bit Instruction Word, 16-Bit Data Bus

16 x 16-Bit, General-Purpose Working Registers Optimized for C-Compiler (High-Speed/Density Code)

♦ Program and Data Memory

16kWords Flash Memory 1,000,000 Flash Write/Erase Cycles 256 Words of Internal Data RAM JTAG Bootloader for Programming

◆ Dual, 16-Bit Sigma-Delta ADCs

Differential Analog Input Channels Programmable Gain of 1x or 16x Integrated Sinc³ Filters Digital Phase Compensation and Trimmable Bandgap Reference

♦ Peripheral Features

Up to 32 General-Purpose I/O Pins 112-Segment LCD Driver Up to 4 COM and 28 Segments Static, 1/2, and 1/3 LCD Bias Supported No External Resistors Required Two Serial USARTs, One with Infrared PWM

Support One-Cycle, 16 x 16 Hardware Multiply/

Accumulate with 40-Bit Accumulator Three 16-Bit Programmable Timers/Counters. One with Infrared PWM Support 8-Bit, Subsecond, System Timer/Alarm

Battery-Backed, 32-Bit RTC with Time-of-Day Alarm and Digital Trim

Programmable Watchdog Timer

♦ Flexible Programming Interface

Bootloader Simplifies Programming In-System Programming Through JTAG Supports In-Application Programming of Flash Memory

♦ Power Consumption

< 28mA at 8MHz, 3.3V Flash Operation 320µA Standby Current in Sleep Mode Low-Power Divide-by-256 Mode



ABSOLUTE MAXIMUM RATINGS

Voltage Range on DV _{DD} Relative to DGND0.3V to +4.0V
Voltage Range on AVDD Relative to AGND0.3V to +4.0V
Voltage Range on AGND Relative to DGND0.3V to +0.3V
Voltage Range on AVDD Relative to DVDD0.3V to +0.3V
Voltage Range on Any Pin Relative to DGND
Except ANO+ ANO- ANO+ ANO- $= 0.3V \text{ to } (DVDD + 0.5V)$

Voltage Range on ANO+, ANO-, AN1-	⊦,
AN1- Relative to AGND	4.0V to +4.0V
Operating Temperature Range	40°C to +85°C
Junction Temperature	+150°C
Storage Temperature Range	65°C to +150°C
Soldering Temperature	See IPC/JEDEC
-	J-STD-020 Specification

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

ELECTRICAL CHARACTERISTICS

(DV_{DD}, AV_{DD} = V_{RST} to 3.6V, V_{REF} = 1.25V (external), f_{HFXIN} = 8MHz, T_A = -40°C to +85°C, unless otherwise noted.) (Note 1)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS	
Digital Supply Voltage	DV _{DD}		V _{RST}	3.3	3.6	V	
Digital Supply Ramp Rate		Can be controlled by placing a 1µF or higher capacitor between DV _{DD} and ground	-16		+16	V/ms	
Digital Power-Fail Reset	V _{RST}		2.8	2.9	3.03	V	
	I _{DD1}	/1 mode		21	28		
	I _{DD2}	/2 mode		11			
Active Current (Note 2)	I _{DD3}	/4 mode		5.7		mA	
(14010 2)	I _{DD4}	/8 mode		3.1			
	I _{DD5}	PMM mode		1.0			
Stop-Mode Current (DV _{DD} plus AV _{DD})	ISTOP			320	760	μΑ	
Battery Supply Voltage	V _{BAT}		1.8		3.8	V	
Battery Current	I _{BAT}	RTCE = 1, DV _{DD} = 0V, V _{BAT} = 3.6V		5.1	10	μΑ	
Input High Voltage	V _{IH}		0.7 x DV _{DD}		DV _{DD} + 0.3	V	
Input Low Voltage	VIL		-0.3		0.3 x DV _{DD}	V	
Input Hysteresis (Schmitt)	VIHYS			0.6		V	
Output High Voltage	\/o	$I_{OH} = +1.5mA$	DV _{DD} - 0	.4		V	
(All Ports)	V _{OH}	$I_{OH} = +2.5$ mA	DV _{DD} - 0	.5		V	
Output Low Voltage (All Ports,	VOL	I _{OL} = 3mA sink current			0.4	V	
RESET)	VOL	I _{OL} = 3.65mA sink current				V	
Input Low Current (All Ports)	IJĽ	V _{IL} = 0.4V; weak pullup enabled	-50		·	μΑ	
RESET Pullup Resistance	R _{RST}		50	100	200	kΩ	
Input Leakage (All Ports)	ΙL	Weak pullup disabled	-1		+1	μΑ	



ELECTRICAL CHARACTERISTICS (continued)

(DV_{DD}, AV_{DD} = V_{RST} to 3.6V, V_{REF} = 1.25V (external), f_{HFXIN} = 8MHz, T_{A} = -40°C to +85°C, unless otherwise noted.) (Note 1)

PARAMETER	SYMBOL		CONDITIONS	MIN	TYP	MAX	UNITS
Analog Supply Voltage	AV _{DD}	$AV_{DD} = DV_{DD}$		2.8	3.3	3.6	V
Active Analog Supply Current	lavdd	Normal operation	on		2.65	3.5	mA
Power-Down Analog Supply Current		APD2:0 = 111b			250	635	μΑ
ANALOG-TO-DIGITAL CONVER	TER DC ACC	URACY					
ADC Resolution		No missing coofilter (see Appe	les, with software lowpass ndix A)		16		bits
Offset Error		Gain = 1			±5.0		mV
Gain Error		Gain = 1			±5.0		%
Gain-Error Drift					±10		ppm
Gain-Error Match					±0.5		%
DC Power-Supply Rejection	PSRR	AN0+, AN0- = A	AGND; $AV_{DD} = 3.0V$ to $3.6V$		80		dB
ANALOG-TO-DIGITAL CONVER	TER DYNAMI	C SPECIFICATION	ONS				
			N _{DD} = 3.3V, AN0 = 25mV, ine wave at 65Hz, gain = 16	48	57		
Signal-to-Noise Ratio	SNR	With software lo	owpass filter, cutoff at 21st Appendix A)		71		dB
		With software lo	owpass filter, cutoff at 7th Appendix A)		74		
Total Harmonic Distortion	THD		NVDD = 3.3V, ANO = 25mV, ine wave at 65Hz, gain = 16 nonic)		-79	-55	dB
ANALOG-TO-DIGITAL CONVER	TER INPUTS						ı
Input-Voltage Range		ANO+, ANO-; AI	N1+, AN1- to AGND	-1		+1	V
Input Sampling Capacitance			Gain = 1		1		_
(Note 3)	CIN	Channel 0	Gain = 16		16		pF
Input Sampling Rate	fs	(Note 4)			1.33		MHz
Sample Output Rate					f _{HFXIN} / 384		sample / sec
Input Impedance to AGND		Gain = 1			750		
(Note 5)		Gain = 16			46		kΩ
Differential Input Impedance		Gain = 1			1500		
(Note 6)		Gain = 16			93		kΩ
Input Bandwidth (-3dB)					5.5		kHz
Reference Input Voltage	V _{REF}			1.2	1.25	1.3	V
Reference Input Sampling Capacitance					2		pF
Reference Input Sampling Rate	fs				1.33		MHz

ELECTRICAL CHARACTERISTICS (continued)

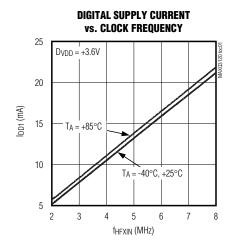
(DV_{DD}, AV_{DD} = V_{RST} to 3.6V, V_{REF} = 1.25V (external), f_{HFXIN} = 8MHz, T_A = -40°C to +85°C, unless otherwise noted.) (Note 1)

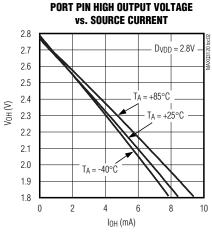
PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
INTERNAL REFERENCE			•			
Reference Output Voltage				1.25		V
Defended to the state of the st				±120		
Reference-Output Temperature Coefficient		With V _{REF} bandgap trimming (ATRM[4:0] = 01111b)		±35		ppm/°C
Load Regulation		$I_{REF} = \pm 2\mu A$, $C_L = 12pF$		±50	±500	μV/μΑ
LCD INTERFACE						
LCD Reference Voltage	V _{LCD}				DV _{DD}	V
LCD Bias Voltage 1	V _{LCD1}	Guaranteed by design	V _{ADJ} + 2/ (V _{LCD} - V _/			V
LCD Bias Voltage 2	V _{LCD2}	Guaranteed by design	V _{ADJ} + 1/ (V _{LCD} - V _/			V
LCD Adjustment Voltage (Note 7)	V _{ADJ}	Guaranteed by design	0		0.4 x V _{LCD}	V
LCD Bias Resistor	RLCD			20		kΩ
LCD Adjust Resistor	RLADJ	LRA4:LRA0 = 0		40		kΩ
		Segment is driven at V _{LCD} ; V _{LCD} = 3V, I _{SEGxx} = -3µA, guaranteed by design	V _{LCD} - 0.02		V _{LCD}	V
LOD Commonwall (allowers	Maria de	Segment is driven at V _{LCD1} ; V _{LCD1} = 2V, I _{SEGxx} = -3µA, guaranteed by design	V _{LCD1} - 0.02		V _{LCD1}	V
LCD Segment Voltage	V _{SEGxx}	Segment is driven at V _{LCD2} ; V _{LCD2} = 1V, I _{SEGxx} = -3µA, guaranteed by design	V _{LCD2} - 0.02		V _{LCD2}	V
		Segment is driven at V _{ADJ} ; V _{ADJ} = 0V, I _{SEGxx} = +3µA, guaranteed by design	V _{ADJ}		0.1	V
CLOCK SOURCE						
External Crystal Frequency	fHFXIN		1		8	MHz
REAL-TIME CLOCK						
RTC Input Frequency	f32KIN	32kHz watch crystal		32.768		kHz
JTAG/FLASH PROGRAMMING						
JTAG Clock Rate	fTCK			5	sysclk / 8	
Flack France Times		Mass erase		904		
Flash Erase Time		Page erase	313			ms
Flash Programming Time			17			μs
Write/Erase Cycles			1,000,000			Cycles
Data Retention			20			Years

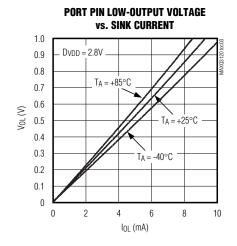
- **Note 1:** Specifications to -40°C are guaranteed by design and not production tested. All typical values are guaranteed by design characterization and are not production tested.
- **Note 2:** Tested with $T_A = +25$ °C, $DV_{DD} = 3.3V$, and all peripherals inactive except for port pins.
- Note 3: These numbers are guaranteed by design and are not tested.
- Note 4: Can be calculated as (fHFXIN / 6).
- **Note 5:** Can be calculated as 6 / (f_{HFXIN} x C_{IN}).
- Note 6: Can be calculated as 12 / (f_{HFXIN} x C_{IN}).
- Note 7: Assumes that no external components are connected to V_{LCD}, V_{LCD1}, V_{LCD2}, or V_{ADJ}.

Typical Operating Characteristics

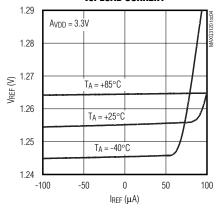
 $(T_A = +25^{\circ}C, \text{ unless otherwise noted.})$

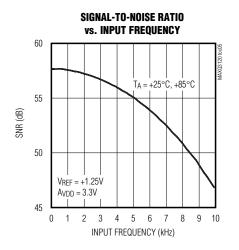






REFERENCE VOLTAGE OUTPUT vs. LOAD CURRENT





Pin Description

PIN	NAME	FUNCTION
22, 38, 60, 74	DV _{DD}	Digital Supply Voltage (+3.3V)
19, 37, 43, 59, 75	DGND	Digital Ground
44	AV_{DD}	Analog Supply Voltage
52	AGND	Analog Ground
12	V _{LCD}	LCD Bias-Control Voltage. Highest LCD drive voltage used in all bias modes. This pin must be connected to an external supply when using the LCD display controller.
13	V _{LCD1}	LCD Bias, Voltage 1. Next highest LCD drive voltage, used in 1/2 and 1/3 LCD bias modes. An internal resistor-divider sets the voltage at this pin. External resistors and capacitors can be used to change LCD voltage or drive capability at this pin. This pin must be shunted externally to V _{LCD2} when using 1/2 bias mode.
14	V _{LCD2}	LCD Bias, Voltage 2. Third highest LCD drive voltage, used in 1/3 LCD bias mode only. An internal resistor-divider sets the voltage at this pin. External resistors and capacitors can be used to change LCD voltage or drive capability at this pin. This pin must be shunted externally to V _{LCD1} when using 1/2 bias mode.
15	V _{ADJ}	LCD Adjustment Voltage. Lowest LCD drive voltage, used in all bias modes. Connect to DGND through an external resistor to provide external control of the LCD contrast. Leave disconnected for internal contrast adjustment.
63	RESET	Digital, Active-Low, Reset Input/Output. The CPU is held in reset when this is low and begins executing from the reset vector when released. The pin includes a pullup current source and should be driven by an open-drain, external source capable of sinking in excess of 2mA. This pin is driven low as an output when an internal reset condition occurs.
20	HFXIN	High-Frequency Crystal Input. Connect an external crystal between HFXIN and HFXOUT to generate the high-frequency system clock. HFXIN and HFXOUT contain integral 16pF load capacitors, so no external capacitor is required.
21	HFXOUT	High-Frequency Crystal Output. Connect an external crystal between HFXIN and HFXOUT to generate the high-frequency system clock. HFXIN and HFXOUT contain integral 16pF load capacitors, so no external capacitor is required.
53	V _{BAT}	Digital Battery-Backup Supply. This supply provides an optional battery backup for the RTC when DV _{DD} power is removed. If this supply is not provided, all functions of the device operate as normal, but the RTC is cleared upon power-on reset (POR).
61	32KIN	32kHz Crystal Input. Connect an external, 32kHz watch crystal between 32KIN and 32KOUT to generate the 32kHz system clock. This clock is required for the RTC to operate.
62	32KOUT	32kHz Crystal Output. Connect an external, 32kHz watch crystal between 32KIN and 32KOUT to generate the 32kHz system clock. This clock is required for the RTC to operate.
51	V _{REF}	Voltage Reference Input/Output. Bias voltage (+1.25V) for the ADCs. An external reference voltage can be connected to this pin when extremely high accuracy is required.
45	ANO-	Negative Input for Sigma-Delta ADC Channel 0
46	AN0+	Positive Input for Sigma-Delta ADC Channel 0
47	AN1-	Negative Input for Sigma-Delta ADC Channel 1
48	AN1+	Positive Input for Sigma-Delta ADC Channel 1



_Pin Description (continued)

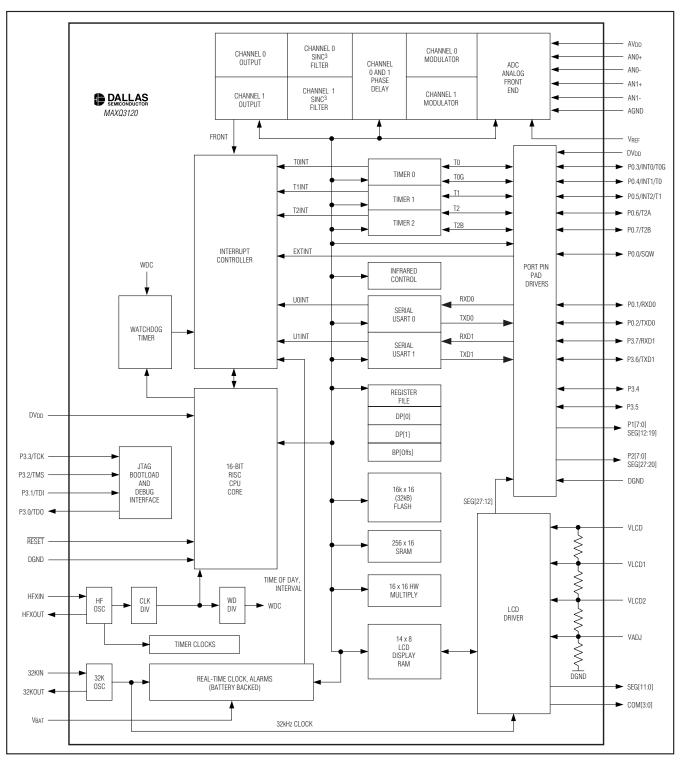
PIN	NAME		FUNC	CTION					
		function as bidirectional I/	O pins only. All port pins only. All port pins only. 3, P0.4, and P0.5 can be only.	nal Edge-Selectable Interru default to input mode with we configured as external interru	ak pullups enabled				
	P0.0-P0.7/	PIN	ALTERNATE F	FUNCTIONS					
	SQW, RXD0,	64	P0.0	RTC Square-Wave Outp	ut —				
64–71	TXD0,	65	P0.1	Serial Port 0 Receive	_				
0171	INTO-INT2/	66	P0.2	Serial Port 0 Transmit	_				
	T2A, T2B/ T0G, T0, T1	67	P0.3	Timer 0 Gate Input	INT0				
	100, 10, 11	68	P0.4	Timer 0 Input	INT1				
		69	P0.5	Timer 1 Input/Output	INT2				
		70	P0.6	Timer 2 Input/Output A (T	2P) —				
		71	P0.7	Timer 2 Input/Output B (T2	2PB) —				
		with weak pullups enabled enables the LCD function	function as both bidirectional I/O pins and LCD segment-drive outputs. All port pins defawith weak pullups enabled after a reset. Setting the LCD enable (PCFx) bit for a group of enables the LCD function and disables the general-purpose I/O function on all pins in the						
		PIN	PORT	LCD SEGMENT	LCD ENABLE				
70.00.1	P1.0-P1.7/ SEG19- SEG12	76	P1.0	SEG19					
76–80, 1, 2, 3		77	P1.1	SEG18	PCF1				
2, 0		78	P1.2	SEG17	-				
		79	P1.3	SEG16					
		80	P1.4	SEG15					
		1	P1.5	SEG14	PCF0				
		2	P1.6	SEG13					
		3	P1.7	SEG12					
		function as both bidirection with weak pullups enabled	nal I/O pins and LCD segm I after a reset. Setting the L	LCD Segment-Driver Output nent-drive outputs. All port pin CD enable (PCFx) bit for a gr nurpose I/O function on all pin	s default to input mode oup of four port pins				
		PIN	PORT	LCD SEGMENT	LCD ENABLE				
	P2.0-P2.7/	28	P2.0	SEG20					
28–34, 39	SEG20-	29	P2.1	SEG21	PCF2				
	SEG27	30	P2.2	SEG22	1 01 2				
		31	P2.3	SEG23					
		32	P2.4	SEG24					
		33	P2.5	SEG25	PCF3				
		34	P2.6	SEG26	1 01 0				
		39	P2.7	SEG27					

Pin Description (continued)

PIN	NAME		FUNCTION	
		General-Purpose, 8-Bit, Digital, I/O only. All port pins default to input mo enabled by default following reset; a	de with weak pullups enabled af	ter a reset. JTAG functions are
		PIN	PORT	ALTERNATE FUNCTION
	P3.0-P3.7/	40	P3.0	TDO-JTAG Data Out
40, 41, 42,	TDO, TDI,	41	P3.1	TDI-JTAG Data In
54–58	TMS, TCK,	42	P3.2	TMS-JTAG Mode Select
	TXDI, RXDI	54	P3.3	TCK-JTAG Clock
		55	P3.4	_
		56	P3.5	_
		57	P3.6	Serial Port 1 Transmit
		58	P3.7	Serial Port 1 Receive
23	SEG0	LCD Segment 0 Driver. Dedicated L	CD drive output.	
18	SEG1	LCD Segment 1 Driver. Dedicated L	_CD drive output.	
17	SEG2	LCD Segment 2 Driver. Dedicated L	CD drive output.	
16	SEG3	LCD Segment 3 Driver. Dedicated L	CD drive output.	
11	SEG4	LCD Segment 4 Driver. Dedicated L	CD drive output.	
10	SEG5	LCD Segment 5 Driver. Dedicated L	CD drive output.	
9	SEG6	LCD Segment 6 Driver. Dedicated I	CD drive output.	
8	SEG7	LCD Segment 7 Driver. Dedicated L	CD drive output.	
7	SEG8	LCD Segment 8 Driver. Dedicated L	CD drive output.	
6	SEG9	LCD Segment 9 Driver. Dedicated L	CD drive output.	
5	SEG10	LCD Segment 10 Driver. Dedicated	LCD drive output.	
4	SEG11	LCD Segment 11 Driver. Dedicated	LCD drive output.	
27	COM0	LCD Common 0 Driver. Dedicated	_CD common-voltage output.	
26	COM1	LCD Common 1 Driver. Dedicated	_CD common-voltage output.	
25	COM2	LCD Common 2 Driver. Dedicated	_CD common-voltage output.	
24	COM3	LCD Common 3 Driver. Dedicated	_CD common-voltage output.	
35, 36, 49, 50, 72, 73	N.C.	No Connection		



Functional Diagram



Detailed Description

The following is an introduction to the primary features of the microcontroller. More detailed descriptions of the device features can be found in the data sheets, errata sheets, and user's guides described later in the *Additional Documentation* section.

MAXQ Core Architecture

The MAXQ3120 is a low-cost, high-performance, CMOS, 16-bit RISC microcontroller with flash memory and an integrated 112-segment LCD controller. It is structured on a highly advanced, accumulator-based, 16-bit RISC architecture. Fetch and execution operations are completed in one cycle without pipelining, because the instruction contains both the op code and data. The result is a streamlined 8 million instructions-per-second (MIPS) microcontroller.

The highly efficient core is supported by a 16-level hardware stack, enabling fast subroutine calling and task switching. Data can be quickly and efficiently manipulated with three internal data pointers. Multiple data pointers allow more than one function to access data memory without having to save and restore data pointers each time. The data pointers can automatically increment or decrement following an operation, eliminating the need for software intervention. As a result, the application speed is greatly increased.

Instruction Set

The instruction set is composed of fixed-length, 16-bit instructions that operate on registers and memory locations. The instruction set is highly orthogonal, allowing arithmetic and logical operations to use any register along with the accumulator. System registers control functionality common to all MAXQ microcontrollers, while peripheral registers control peripherals and functions specific to the MAXQ3120. All registers are subdivided into register modules. The family architecture is modular, so that new devices and modules can reuse code developed for existing products.

The architecture is transport-triggered. This means that writes or reads from certain register locations can also cause side effects to occur. These side effects form the basis for the higher-level op codes defined by the assembler, such as ADDC, OR, JUMP, etc. The op codes are actually implemented as MOVE instructions between certain system register locations, while the assembler handles the encoding, which need not be a concern to the programmer.

The 16-bit instruction word is designed for efficient execution. Bit 15 indicates the format for the source field of

the instruction. Bits 0 to 7 of the instruction represent the source for the transfer. Depending on the value of the format field, this can either be an 8-bit immediate value or a source register. If this field represents a register, the lower four bits contain the module specifier and the upper four bits contain the register index in that module.

Bits 8 to 14 represent the destination for the transfer. This value always represents a destination register, with the lower four bits containing the module specifier and the upper three bits containing the register subindex within that module.

The following types of instructions require the use of the prefix register, PFX, to supply additional data.

- Loading a 16-bit immediate value (with a nonzero high byte) into any register
- Branching to a 16-bit absolute destination address (LJMP or LCALL)
- Selecting one of the upper 8 registers in a system register module as a destination
- Selecting one of the upper 16 registers in a peripheral register module as a source
- Selecting one of the upper 24 registers in a peripheral register module as a destination

For any of these instruction types, the prefix register is used to supply the additional immediate value bits, source bits, and destination bits as needed. This prefix register write is inserted automatically by the assembler and requires only one additional execution cycle for any or all of these conditions.

Memory Organization

The device incorporates several memory areas:

- 2kWords utility ROM
- 16kWords of flash memory for program storage
- 256 words of SRAM for storage of temporary variables
- 16-level, 16-bit-wide stack memory for storage of program return addresses and general-purpose use

The memory is arranged by default in a Harvard architecture, with separate address spaces for program and data memory. The configuration of program and data space depends on the current execution location.

- When executing code from flash memory, the SRAM and utility ROM are accessible in data space.
- When executing code from SRAM, the flash memory and utility ROM are accessible in data space.
- When executing code from the utility ROM, the flash memory and SRAM are accessible in data space.



Refer to the user's guide supplement for this device for more details.

In all cases, whichever memory segment is currently being executed from cannot be accessed in data space. To allow the use of lookup tables and similar constructs in the flash memory, the utility ROM contains a set of lookup and block copy routines (refer to the user's guide supplement for this device for more details).

The incorporation of flash memory allows the device to be reprogrammed, eliminating the expense of throwing away one-time programmable devices during development and field upgrades. Flash memory can be password protected with a 16-word key, denying access to program memory by unauthorized individuals.

Stack Memory

A 16-bit-wide internal stack provides storage for program return addresses and general-purpose use. The stack is used automatically by the processor when the CALL, RET, and RETI instructions are executed and interrupts serviced. The stack can also be used explic-

itly to store and retrieve data by using the PUSH, POP, and POPI instructions.

On reset, the stack pointer, SP, initializes to the top of the stack (0Fh). The CALL, PUSH, and interrupt-vectoring operations increment SP, then store a value at the stack location pointed to by SP. The RET, RETI, POP, and POPI operations retrieve the value at the stack location pointed to by SP, and then decrement SP.

Utility ROM

The utility ROM is a 2kWord block of internal ROM memory that defaults to a starting address of 8000h. The utility ROM consists of subroutines that can be called from application software. These include:

- In-system programming (bootloader) over the JTAG interface
- In-circuit debug routines
- Test routines (internal memory tests, memory loader, etc.)
- User-callable routines for in-application flash programming and code space table lookup

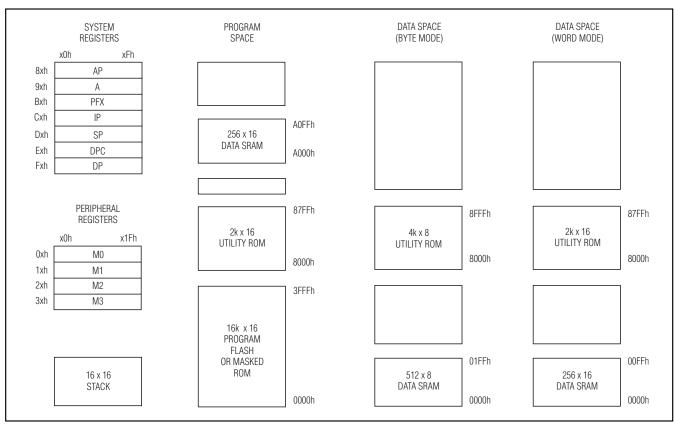


Figure 1. Memory Map



Following any reset, execution begins in the utility ROM. The ROM software determines whether the program execution should immediately jump to the start of user-application code (located at address 0000h), or to one of the special routines mentioned. Routines within the utility ROM are user-accessible and can be called as subroutines by the application software. More information on the utility ROM contents is contained in the user's guide supplement for this device.

Some applications require protection against unauthorized viewing of program code memory. For these applications, access to in-system programming, inapplication programming, or in-circuit debugging functions is prohibited until a password has been supplied.

A single password-lock (PWL) bit is implemented in the SC register. When the PWL is set to one (power-on reset default), the password is required to access the utility ROM, including in-circuit debug and in-system programming routines that allow reading or writing of internal memory. When PWL is cleared to zero, these utilities are fully accessible without the password. The password is automatically set to all ones following a mass erase.

Programming

The flash memory of the microcontroller can be programmed by two different methods: in-system programming and in-application programming. Both methods afford great flexibility in system design as well as reduce the life-cycle cost of the embedded system. These features can be password protected to prevent unauthorized access to code memory.

In-System Programming

An internal bootstrap loader allows the device to be reloaded over a simple JTAG interface. As a result, system software can be upgraded in-system, eliminating the need for a costly hardware retrofit when software updates are required. Remote software uploads are possible that enable physically inaccessible applications to be frequently updated. The interface hardware can be a JTAG connection to another microcontroller, or a connection to a PC serial port using a serial-to-JTAG converter such as the MAXQJTAG-001, available from Maxim Integrated Products/Dallas Semiconductor. If insystem programmability is not required, a commercial gang programmer can be used for mass programming.

Activating the JTAG interface and loading the test access port (TAP) with the system programming instruction invokes the bootloader. Setting the SPE bit to 1 during reset through the JTAG interface executes the bootloader-mode program that resides in the utility ROM. When programming is complete, the bootloader can clear the SPE bit and reset the device, allowing the device to bypass the utility ROM and begin execution of the application software.

The following bootloader functions are supported:

- Load
- Dump
- CRC
- Verify
- Erase

In-Application Programming

The in-application programming feature allows the microcontroller to modify its own flash program memory while simultaneously executing its application software. This allows on-the-fly software updates in mission-critical applications that cannot afford downtime. Alternatively, it allows the application to develop custom loader software that can operate under the control of the application software. The utility ROM contains user-accessible flash programming functions that erase and program flash memory. These functions are described in detail in the user's guide supplement for this device.

Register Set

Most functions of the device are controlled by sets of registers. These registers provide a working space for memory operations as well as configuring and addressing peripheral registers on the device. Registers are divided into two major types: system registers and peripheral registers. The common register set, also known as the system registers, includes the ALU, accumulator registers, data pointers, interrupt vectors and control, and stack pointer. The peripheral registers define additional functionality that may be included by different products based on the MAXQ architecture. This functionality is broken up into discrete modules so that only the features required for a given product need to be included. Tables 1 and 4 show the MAXQ3120 register set.



Table 1. System Register Map

REGISTER INDEX	AP (8h)	A (9h)	PFX (Bh)	IP (Ch)	SP (Dh)	DPC (Eh)	DP (Fh)
0xh	AP	A[0]	PFX	IP	_		_
1xh	APC	A[1]	_	_	SP	_	_
2xh	_	A[2]	_	_	IV	_	_
3xh	_	A[3]	_	_	_	Offs	DP[0]
4xh	PSF	A[4]	_	_	_	DPC	_
5xh	IC	A[5]	_	_	_	GR	_
6xh	IMR	A[6]	_	_	LC[0]	GRL	_
7xh	_	A[7]	_	_	LC[1]	ВР	DP[1]
8xh	SC	A[8]	_	_	_	GRS	_
9xh	_	A[9]	_	_	_	GRH	_
Axh	_	A[10]	_	_	_	GRXL	_
Bxh	IIR	A[11]	_	_	_	BP[offs]	_
Cxh		A[12]				_	
Dxh		A[13]	_	_	_		_
Exh	CKCN	A[14]	_	_	_		_
Fxh	WDCN	A[15]	_	_	_	_	_

Note: Names that appear in italics indicate that all bits of a register are read-only. Names that appear in bold indicate that a register is 16 bits wide. Registers in module AP are bit addressable.

Table 2. System Register Bit Functions

REGISTER								RE	GISTER	BIT									
REGISTER	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0			
AP										_			AP (4 bits)						
APC									CLR	IDS		_	_	MOD2	MOD1	MOD0			
PSF									Z	S		GPF1	GPF0	OV	С	E			
IC									_	_	CGDS	_	_	_	INS	IGE			
IMR									IMS		—		IM3	IM2	IM1	IM0			
SC									TAP	_			_	ROD	PWL				
IIR									IIS		—		II3	II2	II1	IIO			
CKCN									—		—	STOP	SWB	PMME	CD1	CD0			
WDCN									POR	EWDI	WD1	WD0	WDIF	WTRF	EWT	RWT			
A[015]	A[n] (16 bits)																		
PFX	PFX (16 bits)																		
IP								- 1	P (16 bits	s)									
SP	_	_	_	_	_	_	_	_		_		_		SP (4	l bits)				
IV								ľ	V (16 bits	s)									
LC[0]									[0] (16 b										
LC[1]								LC	[1] (16 b	its)									
Offs										1			8 bits)		1				
DPC		_		_	_	_		_	_	_	_	WBS2	WBS1	WBS0	SDPS1	SDPS0			
GR								G	R (16 bit	s)		1	1						
GRL									GR.7	GR.6	GR.5	GR.4	GR.3	GR.2	GR.1	GR.0			
BP				ı	T.	ı	1		P (16 bit			1	1		1				
GRS	GR.7	GR.6	GR.5	GR.4	GR.3	GR.2	GR.1	GR.0	GR.15	GR.14	GR.13	GR.12	GR.11	GR.10	GR.9	GR.8			
GRH									GR.15	GR.14	GR.13	GR.12	GR.11	GR.10	GR.9	GR.8			
GRXL	GR.7	GR.7	GR.7	GR.7	GR.7	GR.7	GR.7	GR.7	GR.7	GR.6	GR.5	GR.4	GR.3	GR.2	GR.1	GR.0			
BP[offs]									offs] (16										
DP[0]								DF	P[0] (16 b	oits)									
DP[1]								DF	P[1] (16 b	oits)									

Table 3. System Register Reset Values

DECICTED								REGIS1	ER BIT	Ī						
REGISTER	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
AP									0	0	0	0	0	0	0	0
APC									0	0	0	0	0	0	0	0
PSF									i	i	0	0	0	0	0	0
IC									0	0	0	0	0	0	0	0
IMR									0	0	0	0	0	0	0	0
SC									1	0	i	i	i	0	0	0
IIR									0	0	0	0	0	0	0	0
CKCN									1	0	0	0	0	0	0	0
WDCN									S	s	0	0	0	s	s	0
A[015]	i	i	i	i	i	i	i	i	i	i	i	i	i	i	i	i
PFX	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0
IP	1	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0
SP	0	0	0	0	0	0	0	0	0	0	0	0	1	1	1	1
IV	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0
LC[0]	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0
LC[1]	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0
Offs									0	0	0	0	0	0	0	0
DPC	0	0	0	0	0	0	0	0	0	0	0	1	1	1	0	0
GR	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0
GRL									0	0	0	0	0	0	0	0
BP	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0
GRS	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0
GRH									0	0	0	0	0	0	0	0
GRXL	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0
BP[offs]	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0
DP[0]	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0
DP[1]	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0

Note: Bits marked with an "i" have an indeterminate value upon reset. Bits marked with an "s" have special behavior upon reset. Refer to the user's guide supplement for this device for more details.

Table 4. Peripheral Register Map

REGISTER INDEX	M0 (0h)	M1 (1h)	M2 (2h)	M3 (3h)	M4 (4h)	M5 (5h)
00h	PO0	T0CN	T1CN	MCNT	_	_
01h	PO1	TOL	T1L	MA	_	_
02h	PO2	TOH	T1H	MB	_	_
03h	PO3	SCON0	T2CNA	MC2	_	_
04h	_	SBUF0	T2H	MC1	_	_
05h	_	SCON1	T2RH	MC0	_	_
06h	EIF0	SBUF1	T2CH	MC1R	_	_
07h	EIE0	_	IRCN	MC0R	_	_
08h	PI0	_	T1CL	ADCN	_	_
09h	PI1	SMD0	T1CH	PHC	_	_
0Ah	PI2	PR0	T1MD	AD0	_	_
0Bh	PI3	SMD1	T2CNB	AD1	_	_
0Ch	EIES0	PR1	T2V	ATRM	_	_
0Dh	_	_	T2R	LCRA	_	_
0Eh	_	_	T2C	LCFG	_	_
0Fh	_	_	T2CFG	_	_	_
10h	PD0	_	_	LCD0	_	_
11h	PD1	_	_	LCD1	_	_
12h	PD2	_	_	LCD2	_	_
13h	PD3	_	_	LCD3	_	_
14h	_	_	_	LCD4	_	_
15h	_	_	_	LCD5	_	_
16h	_	_	_	LCD6	_	_
17h	_	_	_	LCD7	_	_
18h	RTRM	_	_	LCD8	_	_
19h	RCNT	_	_	LCD9	_	_
1Ah	RTSS	_	_	LCD10	_	_
1Bh	RTSH	ICDF	_	LCD11	_	_
1Ch	RTSL	_	_	LCD12	_	_
1Dh	RSSA	_	_	LCD13	_	_
1Eh	RASH	_	_	_	_	_
1Fh	RASL	_	_	_	_	_

Note: Names that appear in italics indicate that all bits of a register are read-only. Names that appear in bold indicate that a register is 16 bits wide. Registers in module AP are bit addressable.

Table 5. Peripheral Register Bit Functions

REGISTER								R	EGISTER	BIT						
nedio i en	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
PO0												PO0 (8	B bits)			
PO1												PO1 (8	B bits)			
PO2												PO2 (8	B bits)			
PO3												PO3 (8	bits)			
EIF0									_	_	_	_	_	IE2	IE1	IE0
EIE0									_	1	_	_	_	EX2	EX1	EX0
PI0												PI0 (8	bits)			
PI1												PI1 (8	bits)			
PI2												PI2 (8	bits)			
PI3												PI3 (8	bits)			
EIES0									_		_	_	_	IT2	IT1	ITO
PD0												PD0 (8	bits)			
PD1												PD1 (8	bits)			
PD2									PD2 (8 bits)							
PD3									PD3 (8 bits)							
RTRM									TSGN TRM (5 bits)							
RCNT	WE	_	_	_	_	_	FT	SQE	ALSF	ALDF	RDYE	RDY	BUSY	ASE	ADE	RTCE
RTSH								R	TSH (16 b	its)						
RTSL								F	TSL (16 b	its)						
RSSA	_	_	_	_	_					RS	SSA (11 bit	s)				
RASH									_	_	_	_		RASH	(4 bits)	
RASL								R	ASL (16 b	its)						
TOCN									ET0	TOM	TF0	TR0	GATE	С/Т	M1	MO
TOL												T0L (8	bits)			
TOH												T0H (8	bits)			
SCON0									SM0/FE	SM1	SM2	REN	TB8	RB8	TI	RI
SBUF0												SBUF0	(8 bits)			
SCON1									SM0/FE	SM1	SM2	REN	TB8	RB8	TI	RI
SBUF1												SBUF1	(8 bits)	•	•	
SMD0									EPWM	OFS	_	_	_	ESI	SMOD	FEDE
PR0								F	PR0 (16 bi	ts)						
SMD1									_	_	_	_	_	ESI	SMOD	FEDE
PR1								·	PR1 (16 bit	ts)		•				

Table 5. Peripheral Register Bit Functions (continued)

	REGISTER BIT															
REGISTER	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
ICDF									_	_	_	_	PSS1	PSS0	SPE	_
T1CN									TF1	EXF1	T10E	DCEN	EXEN1	TR1	C/T1	CPRL1
T1L												T1L (8	bits)			
T1H												T1H (8	bits)			
T2CNA									ET2	T20E0	T2POL0	TR2L	TR2	CPRL2	SS2	G2EN
T2H											Т	2V[15:8]	(8 bits)			
T2RH											Т	2R[15:8] (8 bits)			
T2CH											T	2C[15:8] (8 bits)			
IRCN									_	_	_	_	_	IREN	IRTX	IRBB
T1CL												T1CL (3 bits)			
T1CH												T1CH (8 bits)			
T1MD									_	_	_	_	_	_	ET1	T1M
T2CNB									ET2L T2OE1 T2POL1 — TF2 TCC2 TF2L							TC2L
T2V								-	T2V (16 bits)							
T2R								-	T2R (16 bi	ts)						
T2C								-	Г2С (16 bi	ts)						
T2CFG									T2CI	DIV2	DIV1	DIV0	T2MD	CCF1	CCF0	C/T2
MCNT									OF	MCW	CLD	SQU	OPCS	MSUB	MMAC	SUS
MA									MA (16 bit	ts)						
MB									MB (16 bit	ts)						
MC2	_	_	_	_	_	_	_					MC2 (8	3 bits)			
MC1								N	ИС1 (16 b	its)						
MC0								N	/C0 (16 b	its)						
MC1R								M	IC1R (16 b	oits)						
MC0R								M	ICOR (16 b	oits)						
ADCN	G3	G2	G1	G0	_	APD2	APD1	APD0	UFF	EDBI	FLU1	FLU0	FOV1	FOV0	ABF1	ABF0
PHC	ZPS	_	_	_	_	_	_				PH	H (9 bits))			
AD0								,	AD0 (16 bi	its)						
AD1								/	AD1 (16 bits)							
ATRM										_	_		AE	3GT (5 bi	ts)	
LCRA	_	_	_	DUTY1	DUTY0	FRM3	FRM2	FRM1	FRM0	LCCS	LRIG	_	LRA3	LRA2	LRA1	LRA0
LCFG									PCF3	PCF2	PCF1	PCF0	_	_	OPM	DPE
LCD[013]												LCD[n]	(8 bits)			

Table 6. Peripheral Register Bit Reset Values

REGISTER								REGIS1	ER BIT	•						
REGISTER	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
PO0									1	1	1	1	1	1	1	1
PO1									1	1	1	1	1	1	1	1
PO2									1	1	1	1	1	1	1	1
PO3									1	1	1	1	1	1	1	1
EIF0									0	0	0	0	0	0	0	0
EIE0									0	0	0	0	0	0	0	0
PI0									S	S	S	S	S	S	S	S
PI1									S	S	S	S	S	S	S	S
PI2									S	S	S	S	S	S	S	S
PI3									S	S	S	S	S	S	S	S
EIES0									0	0	0	0	0	0	0	0
PD0									0	0	0	0	0	0	0	0
PD1									0	0	0	0	0	0	0	0
PD2									0	0	0	0	0	0	0	0
PD3									0	0	0	0	0	0	0	0
RTRM									0	0	S	S	S	S	S	S
RCNT	0	0	0	0	0	0	0	0	0	0	0	0	1	0	0	S
RTSS									S	S	S	S	S	S	S	S
RTSH	S	S	S	S	S	S	S	S	S	S	S	S	S	S	S	S
RTSL	S	S	S	S	S	S	S	S	S	S	S	S	S	S	S	S
RSSA	0	0	0	0	0	S	S	S	S	S	S	S	S	S	S	S
RASH									0	0	0	0	S	S	S	S
RASL	S	S	S	S	S	S	S	S	S	S	S	S	S	S	S	S
T0CN									0	0	0	0	0	0	0	0
TOL									0	0	0	0	0	0	0	0
T0H									0	0	0	0	0	0	0	0
SCON0									0	0	0	0	0	0	0	0
SBUF0									0	0	0	0	0	0	0	0
SCON1									0	0	0	0	0	0	0	0
SBUF1									0	0	0	0	0	0	0	0
SMD0									0	0	0	0	0	0	0	0
PR0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0
SMD1									0	0	0	0	0	0	0	0
PR1	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0
ICDF									0	0	0	0	0	0	0	0
T1CN									0	0	0	0	0	0	0	0
T1L									0	0	0	0	0	0	0	0

Table 6. Peripheral Register Bit Reset Values (continued)

								REGIST	ER BIT							
REGISTER	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
T1H									0	0	0	0	0	0	0	0
T2CNA									0	0	0	0	0	0	0	0
T2H									0	0	0	0	0	0	0	0
T2RH									0	0	0	0	0	0	0	0
T2CH									0	0	0	0	0	0	0	0
IRCN									0	0	0	0	0	0	0	0
T1CL									0	0	0	0	0	0	0	0
T1CH									0	0	0	0	0	0	0	0
T1MD									0	0	0	0	0	0	0	0
T2CNB									0	0	0	0	0	0	0	0
T2V	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0
T2R	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0
T2C	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0
T2CFG									0	0	0	0	0	0	0	0
MCNT									0	0	0	0	0	0	0	0
MA	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0
MB	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0
MC2	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0
MC1	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0
MC0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0
MC1R	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0
MC0R	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0
ADCN	S	S	S	S	0	S	S	S	0	0	0	0	0	0	0	0
PHC	S	0	0	0	0	0	0	S	S	S	S	S	S	S	S	S
AD0	i	i	i	i	i	i	i	i	i	i	i	i	i	i	i	i
AD1	i	i	i	i	i	i	i	i	i	i	i	i	i	i	i	i
ATRM									0	0	0	S	S	S	S	S
LCRA	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0
LCFG									0	0	0	0	0	0	0	0
LCD[013]									0	0	0	0	0	0	0	0

Note: Bits marked with an "i" have an indeterminate value upon reset. Bits marked with an "s" have special behavior upon reset. Refer to the user's guide supplement for this device for more details.

System Timing

The MAXQ3120 generates its internal system clock from an external high-frequency crystal. Because the MAXQ3120 includes internal capacitors for this purpose, no external capacitors are required to use the high-frequency crystal. The MAXQ3120 should not be driven directly by an external clock source.

A crystal warmup counter enhances operational reliability. Each time the external crystal oscillation must restart, such as after exiting stop mode, the device initiates a crystal warmup period of 65,536 oscillations. This allows time for the crystal amplitude and frequency to stabilize before using it as a clock source.

Power Management

Advanced power-management features minimize power consumption by dynamically matching the processing speed of the device to the required performance level. This means device operation can be slowed and power consumption minimized during periods of reduced

activity. When more processing power is required, the microcontroller can increase its operating frequency. Software-selectable clock-divide operations allow flexibility, selecting whether a system clock cycle is 1, 2, 4, or 8 oscillator cycles. By performing this function in software, a lower power state can be entered without the cost of additional hardware.

For extremely power-sensitive applications, two additional low-power modes are available.

- Divide-by-256 power-management mode (PMM) (PMME = 1, CD1:0 = 00b)
- Stop mode (STOP = 1)

In PMM, one system clock is 256 oscillator cycles, significantly reducing power consumption while the microcontroller functions at reduced speed. The optional switchback feature allows enabled interrupt sources, such as the external interrupts and USARTs, to cause the processor to quickly exit PMM mode and return to a faster internal clock rate.

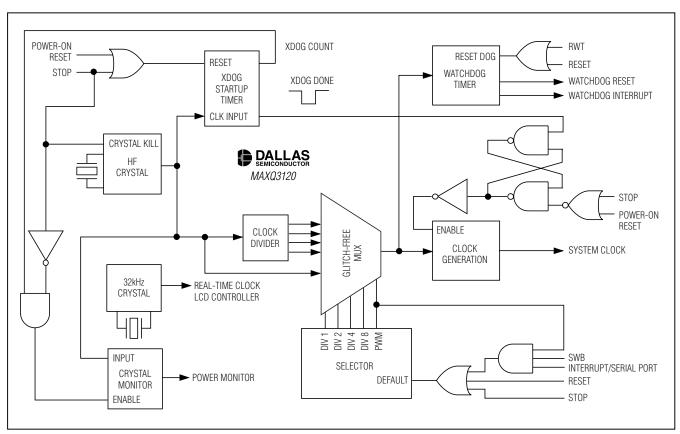


Figure 2. Clock Sources

Power consumption reaches its minimum in stop mode. In this mode, the external high-frequency oscillator, system clock, and all code execution is halted. Stop mode is exited when an enabled external interrupt pin is triggered, an external reset signal is applied to the RESET pin, or the RTC time-of-day alarm is activated.

The 32kHz clock continues running during stop mode, enabling the following peripherals to keep running during stop mode.

- The RTC always continues running during stop mode.
- The LCD controller continues running during stop mode if it is running from the 32kHz clock (LCCS = 0).

Interrupts

Multiple reset sources are available for quick response to internal and external events. The MAXQ architecture uses a single interrupt vector (IV), single interrupt-service routine (ISR) design. For maximum flexibility, interrupts can be enabled globally, individually, or by module. When an interrupt condition occurs, its individual flag is set, even if the interrupt source is disabled at the local, module, or global level. Interrupt flags must be cleared within the user-interrupt routine to avoid repeated interrupts from the same source. Application software must ensure a delay between the write to the flag and the RETI instruction to allow time for the interrupt hardware to remove the internal interrupt condition. Asynchronous interrupt flags require a one-instruction delay and synchronous interrupt flags require a twoinstruction delay.

When an enabled interrupt is detected, software jumps to a user-programmable interrupt vector location. The IV register defaults to 0000h on reset or power-up, so if it is not changed to a different address, the user program must determine whether a jump to 0000h came from a reset or interrupt source.

Once software control has been transferred to the ISR, the interrupt identification register (IIR) can determine if a system register or peripheral register was the source of the interrupt. The specified module can then be interrogated for the specific interrupt source and software can take appropriate action. Because the interrupts are evaluated by user software, the user can define a unique interrupt priority scheme for each application. The following interrupt sources are available.

- Watchdog Interrupt
- External Interrupts 0 to 2
- RTC Time-of-Day and Subsecond Alarms

- Serial Port 0 Receive and Transmit Interrupts
- Serial Port 1 Receive and Transmit Interrupts
- Timer 0 Overflow Interrupt
- Timer 1 Overflow and External Trigger Interrupts
- Timer 2 Low Compare, Low Overflow, Capture/ Compare, and Overflow Interrupts

Reset Sources

Several reset sources are provided for microcontroller control. Although code execution is halted in the reset state, the high-frequency oscillator and the 32kHz oscillator continue to oscillate. Internal resets such as the power-on and watchdog resets assert the RESET pin low.

Power-On Reset

An internal power-on reset circuit enhances system reliability. This circuit forces the device to perform a power-on reset whenever a rising voltage on DV_{DD} climbs above the V_{RST} level. At this point, the following events occur:

- All registers and circuits enter their reset state (except for the RTC, if it is battery-backed)
- The POR flag (WDCN.7) is set to indicate the source of the reset
- Code execution begins at location 8000h

Watchdog Timer Reset

The watchdog timer functions are described in the *MAXQ Family User's Guide*. Execution resumes at location 8000h following a watchdog timer reset.

External System Reset

Asserting the external $\overline{\text{RESET}}$ pin low causes the device to enter the reset state. The external reset functions as described in the *MAXQ Family User's Guide*. Execution resumes at location 8000h after the $\overline{\text{RESET}}$ pin is released.

I/O Ports

The microcontroller uses the Type C and Type D bidirectional I/O ports described in the *MAXQ Family User's Guide*. The use of two port types allows for maximum flexibility when interfacing to external peripherals. Each port has eight independent, general-purpose I/O pins and three configure/control registers. Many pins support alternate functions such as timers or interrupts, which are enabled, controlled, and monitored by dedicated peripheral registers. Using the alternate function automatically converts the pin to that function.



Type C port pins have Schmitt Trigger receivers and full CMOS output drivers, and can support alternate functions. The pin is either tri-stated or a weak pullup when defined as an input, dependent on the state of the corresponding bit in the output register.

Type D port pins have Schmitt Trigger receivers and full CMOS output drivers, and can support alternate functions. The pin is either tri-stated or a weak pullup when defined as an input, dependent on the state of the corresponding bit in the output register. All Type D pins also have interrupt capability.

_High-Speed Hardware Multiplier

The hardware multiplier module performs high-speed multiply, square, and accumulate operations, and can complete a 16-bit x 16-bit multiply-and-accumulate operation in a single cycle. The hardware multiplier consists of two 16-bit parallel-load operand registers (MA, MB), a 40-bit accumulator that is formed by three

16-bit parallel registers (MC2, MC1, and MC0), and a status/control register (MCNT). Loading the registers can automatically initiate the operation, saving time on repetitive calculations. The accumulate function of the hardware multiplier is an essential element of digital filtering, signal processing, and proportional/integral/derivative (PID) algorithm-based control systems.

The hardware multiplier module supports the following operations:

- Multiply unsigned (16 bit x 16 bit)
- Multiply signed (16 bit x 16 bit)
- Multiply-accumulate unsigned (16 bit x 16 bit)
- Multiply-accumulate signed (16 bit x 16 bit)
- Square unsigned (16 bit)
- Square signed (16 bit)
- Square-accumulate unsigned (16 bit)
- Square-accumulate signed (16 bit)

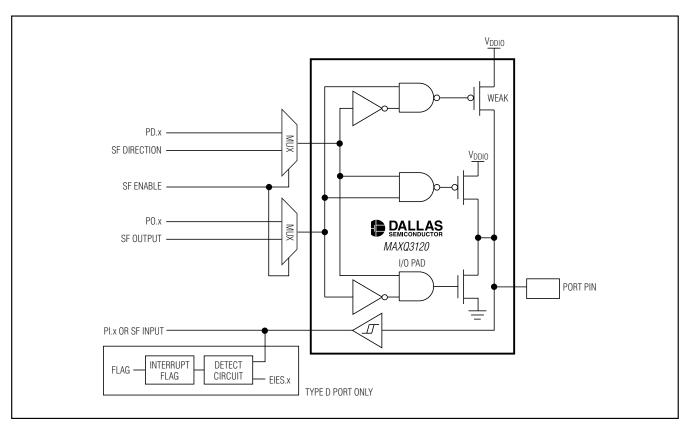


Figure 3. Type C/D Port Pin Schematic



Real-Time Clock

A binary real-time clock keeps the time of day in absolute seconds with 1/256-second resolution. The 32-bit second counter can count up to approximately 136 years and be translated to calendar format by the application software. A time-of-day alarm and independent subsecond alarm can cause an interrupt, or wake the device from stop mode.

The independent subsecond alarm runs from the same RTC, and allows the application to perform periodic interrupts up to 8 seconds with a granularity of approximately 3.9ms. This creates an additional timer that can be used to measure long periods without performance degradations. Traditionally, long time periods have been measured using multiple interrupts from shorter programmable timers. Each timer interrupt required servicing, with each accompanying interruption slowing system operation. By using the RTC subsecond timer as a long-period timer, only one interrupt is needed, eliminating the performance hit associated with using a shorter timer.

An internal crystal oscillator clocks the RTC using integrated 6pF load capacitors, and gives the best performance when mated with a 32.768kHz crystal rated for a 6pF load. No external load capacitors are required. Higher accuracy can be obtained by using the digital RTC trim function. The frequency accuracy of a crystal-based oscillator circuit is dependent upon crystal accuracy, the match between the crystal and the oscillator capacitor load, ambient temperature, etc.

Programmable Timers

The MAXQ3120 incorporates one instance each of the timer 0, timer 1, and timer 2 peripherals. These timers can be used in counter/timer/capture/compare/PWM functions, allowing precise control of internal and external events. Timer 2 supports optional single-shot, external gating, and polarity control options as well as carrier generation support for infrared transmit/receive functions using serial port 0.

Timer 0

The timer 0 peripheral includes the following:

- 8-bit autoreload timer/counter
- 13-bit or 16-bit timer/counter
- Dual 8-bit timer/counter
- External pulse counter

Timer 1

The timer 1 peripheral includes the following:

- 16-bit autoreload timer/counter
- 16-bit capture
- 16-bit counter
- Clock generation output

Timer 2

The timer 2 peripheral includes the following:

- 16-bit autoreload timer/counter
- 16-bit capture
- 16-bit counter
- 8-bit capture and 8-bit timer
- 8-bit counter and 8-bit timer
- Infrared carrier generation support

Watchdog Timer

An internal watchdog timer greatly increases system reliability. The timer resets the processor if software execution is disturbed. The watchdog timer is a free-running counter designed to be periodically reset by the application software. If software is operating correctly, the counter is periodically reset and never reaches its maximum count. However, if software operation is interrupted, the timer does not reset, triggering a system reset and optionally a watchdog timer interrupt. This protects the system against electrical noise or electrostatic discharge (ESD) upsets that could cause uncontrolled processor operation. The internal watchdog timer is an upgrade to older designs with external watchdog devices, reducing system cost and simultaneously increasing reliability.

The watchdog timer is controlled through bits in the WDCN register. Its timeout period can be set to one of four programmable intervals ranging from 2^{12} to 2^{21} system clocks in its default mode, allowing flexibility to support different types of applications. The interrupt occurs 512 system clocks before the reset, allowing the system to execute an interrupt and place the system in a known, safe state before the device performs a total system reset. At 8MHz, watchdog timeout periods can be programmed from 512µs to 61.7s, depending on the system clock mode.



In-Circuit Debug

Embedded debugging capability is available through the JTAG-compatible TAP. Embedded debug hardware and embedded ROM firmware provide in-circuit debugging capability to the user application, eliminating the need for an expensive in-circuit emulator. Figure 4 shows a block diagram of the in-circuit debugger. The in-circuit debug features include:

- Hardware debug engine
- Set of registers able to set breakpoints on register, code, or data accesses
- Set of debug service routines stored in the utility ROM

The embedded hardware debug engine is an independent hardware block in the microcontroller. The debug engine can monitor internal activities and interact with selected internal registers while the CPU is executing user code. Collectively, the hardware and software features allow two basic modes of in-circuit debugging:

- Background mode allows the host to configure and set up the in-circuit debugger while the CPU continues to execute the application software at full speed. Debug mode can be invoked from background mode.
- Debug mode allows the debug engine to take control of the CPU, providing read/write access to internal registers and memory, and single-step trace operation.

Serial Peripherals

The MAXQ3120 incorporates two 8051-style universal synchronous/asynchronous receiver/transmitters. The USARTs allow the device to conveniently communicate with other RS-232 interface-enabled devices, as well as PCs and serial modems when paired with an external RS-232 line driver/receiver. The dual independent USARTs can communicate simultaneously at different baud rates with two separate peripherals. The USART can detect framing errors and indicate the condition through a user-accessible software bit.

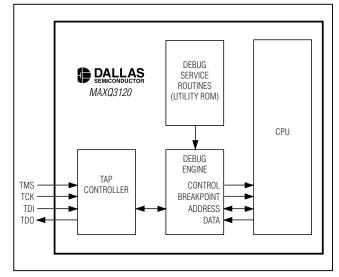


Figure 4. In-Circuit Debugger

The time base of the serial ports is derived from either a division of the system clock or the dedicated baud clock generator. The following table summarizes the operating characteristics as well as the maximum baud rate of each mode.

Serial port 0 contains additional functionality to support low-speed infrared transmission in combination with the PWM function of timer 2. When enabled in this mode, the serial port automatically outputs a waveform generated by combining the normal serial port output waveform with the PWM carrier waveform output by timer 2, using a logical OR or logical NOR function. The output of serial port 0 in this mode can be used to drive an infrared LED to communicate using a fixed-frequency carrier modulated signal. Depending on the drive strength required, the output may require a buffer when used for this purpose.

MODE	TYPE	START BITS	DATA BITS	STOP BIT	MAX BAUD RATE AT 8MHz
Mode 0	Synchronous	_	8	_	2Mbps
Mode 1	Asynchronous	1	8	1	250kbps
Mode 2	Asynchronous	1	8 + 1	1	250kbps
Mode 3	Asynchronous	1	8 + 1	1	250kbps

LCD Driver

The MAXQ3120 microcontroller incorporates an LCD driver that interfaces to common low-voltage displays. By incorporating the LCD driver into the microcontroller, the design requires only an LCD glass rather than a considerably more expensive LCD module. Every character in an LCD glass is composed of one or more segments, each of which is activated by selecting the appropriate segment and common signal. The microcontroller can drive up to 112 LCD glass segments by multiplexing combinations of 28 segment (SEG0–SEG27) outputs and four common-signal outputs (COM0–COM3). Eight of the segment outputs can also be used as general-purpose port pins, if they are not needed to drive the LCD.

The segments are easily addressed by writing to dedicated display memory. Once the LCD driver settings and display memory have been initialized, the 14-byte display memory is periodically scanned, and the segment and common signals are generated automatically at the selected display frequency. No additional processor overhead is required while the LCD driver is running. Unused display memory can be used for general-purpose storage.

The design is further simplified and cost-reduced by the inclusion of software-adjustable internal voltagedividers to control display contrast. If desired, contrast can also be controlled by an external resistance. The features of the LCD driver include the following:

Automatic LCD segment and common-drive signal generation

- Four display modes supported:
 - Static (COM0)
 - 1/2 duty multiplexed with 1/2 bias voltages (COM0, COM1)
 - 1/3 duty multiplexed with 1/3 bias voltages (COM0, COM1, COM2)
 - 1/4 duty multiplexed with 1/3 bias voltages (COM0, COM1, COM2, COM3)
- Up to 28 segment outputs and four common-signal outputs
- 14 bytes (112 bits) of display memory
- Adjustable frame frequency
- Internal voltage-divider resistors eliminate requirement for external components
- Internal adjustable resistor allows contrast adjustment without external components
- Flexibility to use external resistors to adjust drive voltages and current capacity

A simple LCD-segmented glass interface example demonstrates the minimal hardware required to interface to a MAXQ3120 microcontroller. A two-character LCD is controlled, with each character containing seven segments plus decimal point. The LCD driver is configured for 1/2 duty cycle operation, meaning the active segment is controlled using a combination of segment signals, and COM0 or COM1 signals are used to select the active display.

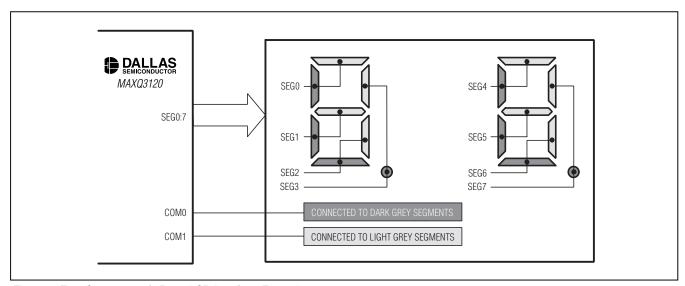


Figure 5. Two-Character, 1/2 Duty, LCD Interface Example



Analog Front-End

The MAXQ3120 microcontroller incorporates an analog front-end for dedicated analog-to-digital conversion. This peripheral converts and digitally filters two differential signal channels with no CPU overhead.

The two conversion channels operate completely in parallel, running at the same sample rate whether one channel or both channels are enabled. If one or both channels are not in use, they may be powered down to conserve supply current.

The two input signals for each channel form a true differential pair. Each of the two signals (AN0+ and AN0- for channel 0, AN1+ and AN1- for channel 1) can vary across the entire +1V to -1V analog input range, without regard to the level of the other signal in the pair. The ini-

tial stage for each channel is a programmable gain function (1x, 16x) that can be set by software independently for each channel. Next, a second-order sigmadelta modulator samples each input signal.

When using both channels to measure the same signal (as is the case when measuring voltage and current for power calculations), a phase-correction buffer is provided to compensate for any phase shift between the two channels caused by external circuitry. The phase-correction buffer operates digitally on the output bit stream of one of the two channels and can delay either channel's bit stream with respect to the other channel's bit stream by up to 140 bits.

Next, the bit streams for the two channels travel through two digital sinc³ lowpass filters, which convert the bit streams to 16-bit PCM values for additional processing.

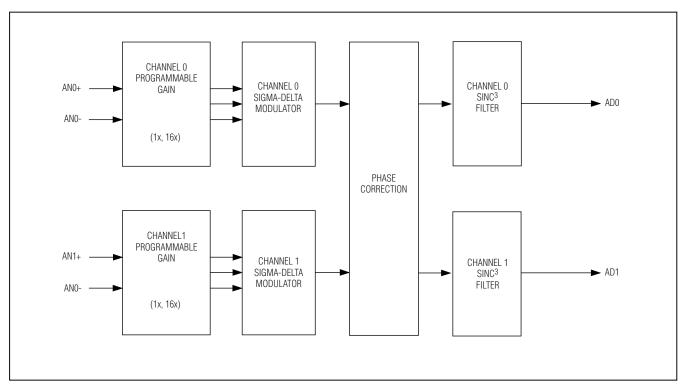


Figure 6. Analog Front-End Block Diagram

Applications Information

The low-power, high-performance RISC architecture of the MAXQ3120 makes it an excellent fit for many portable or battery-powered applications that require cost-effective computing. The high-throughput core is complemented by a dual-differential channel, 16-bit sigma-delta ADC, and 16-bit hardware multiplier-accumulator, allowing the implementation of sophisticated computational algorithms. Applications benefit from a wide range of peripheral interfaces, allowing the microcontroller to communicate with many external devices. With integrated LCD support of up to 112 segments (4 x 28), applications can support complex user interfaces. Displays are driven directly with no additional external hardware required. Contrast can be adjusted using a built-in, adjustable resistor. The simplified architecture reduces component count and board space, critical factors in the design of portable systems.

The MAXQ3120 is ideally suited for single-phase electricity metering applications as well as other applications that require high-precision analog-to-digital conversion and signal processing.

Additional Documentation

Designers must have four documents to fully use all the features of this device. This data sheet contains pin descriptions, feature overviews, and electrical specifications. Errata sheets contain deviations from published specifications. The user's guides offer detailed information about programming, device features, and operation. The following documents can be downloaded from www.maxim-ic.com/microcontrollers.

- The MAXQ3120 data sheet, which contains electrical/timing specifications and pin descriptions, available at www.maxim-ic.com/MAXQ3120.
- The MAXQ3120 errata sheet, available at www.maxim-ic.com/errata.
- The MAXQ Family User's Guide, which contains detailed information on core features and operation, including programming, available at www.maxim-ic.com/MAXQUG.

 The MAXQ Family User's Guide: MAXQ3120 Supplement, which contains detailed information on features specific to the MAXQ3120, available at www.maxim-ic.com/MAXQ3120UG.

Development and Technical Support

A variety of highly versatile, affordably priced development tools for this microcontroller are available from Maxim/Dallas Semiconductor and third-party suppliers, including:

- Compilers
- In-circuit emulators
- Integrated development environments (IDEs)
- JTAG-to-serial converters for programming and debugging

A partial list of development tool vendors can be found on our website at www.maxim-ic.com/microcontrollers. Technical support is available through email at maxq.support@dalsemi.com.

Definitions

Offset Error

For an ideal converter, the first transition occurs at 0.5 LSB above zero. Offset error is the amount of deviation between the measured first transition point and the ideal point.

Gain Error

With a full-scale analog voltage applied to the ADC (resulting in all ones in the digital code), gain error is defined as the amount of deviation between the ideal transfer function and the measured transfer function (with the offset error removed). Gain error is usually expressed in LSB or as a percentage of full-scale range (%FSR).

Power-Supply Rejection Ratio

Power-supply rejection ratio (PSRR) is the ratio of changes in the power supply (V) to changes in the converter output (V). It is typically measured in decibels.



Appendix A: Applying a Lowpass Filter (LPF) for ApplicationOptimized ADC Performance

The MAXQ3120 gives a user application program direct access to the ADC data stream right after its sinc3 filters. This unique feature permits the MAXQ3120 to be optimized for its target applications. With the device's 8MIPS processing power and the 1-cycle MAC, a linear FIR filter can be easily computed in the user application program. This section provides a simple LP FIR filter to improve the SNR and THD performance of the MAXQ3120 for the power-metering application.

Filter Specifications and Coefficients

Input signal frequency (fin) = 60Hz

Sampling frequency (fs) = 20833Hz

Window = hamming

Cutoff frequency (attenuates after 7th harmonic) = 0.06π = 625Hz

Cutoff frequency (for 21st harmonic) = $0.18\pi = 1875Hz$

Transition width = $0.35\pi = 3646$ Hz

Filter length = 23

1) LPF coefficients (up to 21st harmonic). Note that these coefficients have been converted to 16-bit fixed-point numbers. A shift of 17 places is required after the multiply-accumulate operation.

FILTER_COEFFICIENT_0	-19	FILTER_COEFFICIENT_12	21893
FILTER_COEFFICIENT_1	-288	FILTER_COEFFICIENT_13	17547
FILTER_COEFFICIENT_2	-786	FILTER_COEFFICIENT_14	11739
FILTER_COEFFICIENT_3	-1402	FILTER_COEFFICIENT_15	6018
FILTER_COEFFICIENT_4	-1670	FILTER_COEFFICIENT_16	1627
FILTER_COEFFICIENT_5	-876	FILTER_COEFFICIENT_17	-876
FILTER_COEFFICIENT_6	1627	FILTER_COEFFICIENT_18	-1670
FILTER_COEFFICIENT_7	6018	FILTER_COEFFICIENT_19	-1402
FILTER_COEFFICIENT_8	11739	FILTER_COEFFICIENT_20	-786
FILTER_COEFFICIENT_9	17547	FILTER_COEFFICIENT_21	-288
FILTER_COEFFICIENT_10	21893	FILTER_COEFFICIENT_22	-19
FILTER_COEFFICIENT_11	23506	_	_

2) LPF coefficients (up to 7th harmonic). Note that these coefficients have been converted to 16-bit fixed-point numbers. A shift of 18 places is required after the multiply-accumulate operation.

FILTER_COEFFICIENT_0	849	FILTER_COEFFICIENT_12	23295
FILTER_COEFFICIENT_1	1420	FILTER_COEFFICIENT_13	21720
FILTER_COEFFICIENT_2	2553	FILTER_COEFFICIENT_14	19291
FILTER_COEFFICIENT_3	4334	FILTER_COEFFICIENT_15	16269
FILTER_COEFFICIENT_4	6754	FILTER_COEFFICIENT_16	12966
FILTER_COEFFICIENT_5	9700	FILTER_COEFFICIENT_17	9700
FILTER_COEFFICIENT_6	12966	FILTER_COEFFICIENT_18	6754
FILTER_COEFFICIENT_7	16269	FILTER_COEFFICIENT_19	4334
FILTER_COEFFICIENT_8	19291	FILTER_COEFFICIENT_20	2553
FILTER_COEFFICIENT_9	21720	FILTER_COEFFICIENT_21	1420
FILTER_COEFFICIENT_10	23295	FILTER_COEFFICIENT_22	849
FILTER_COEFFICIENT_11	23840	_	_



Filter Results

Below is the summary of the SNR and THD values before and after applying the above LPFs, for 60 MAXQ3120 units.

An application engineer can easily implement his own favorite lowpass filters to optimize MAXQ3120 for his target applications. However, he does need to consider

the filter complexity and its processor resource requirement (CPU cycles and storage space) to strike an optimal balance. The above 23-tap LPF takes 23 x 2 bytes of RAM and 107 clock cycles of the MAXQ3120 to complete. Note that the number of cycles varies from filter to filter because the number of shifts required to normalize the multiply-accumulate result will vary.

CONDITION		SNR		THD				
CONDITION	MIN	AVG	MAX	MIN	AVG	MAX		
Before LPF	56.5	56.8	57.1	-84.5	-81.1	-77.3		
After LPF, 21st Harmonic	68.2	70.7	71.4	-85.8	-83.1	-79.4		
After LPF, 7th Harmonic	71.5	73.8	74.7	-88.9	-85.7	-82.0		

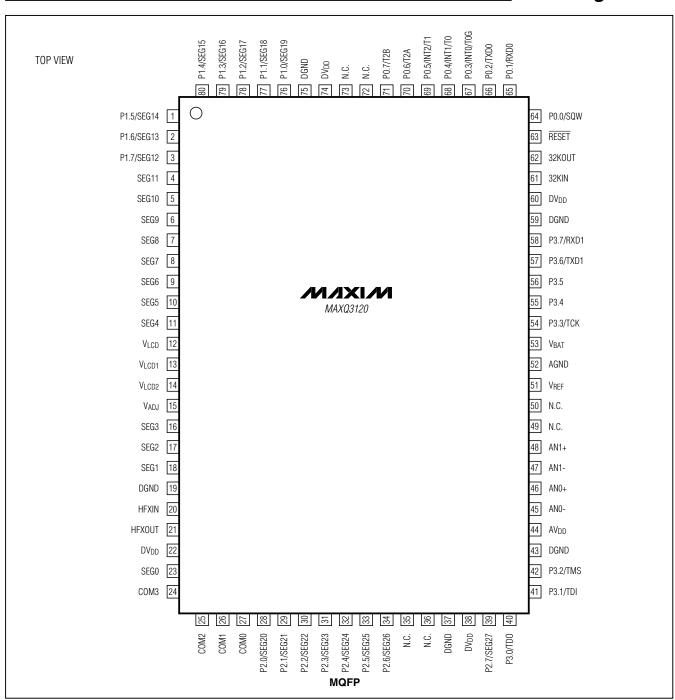
Selector Guide

PART	TEMP RANGE	PROGRAM MEMORY	DATA MEMORY	LCD SEGMENTS	EXTERNAL INTERRUPTS	USARTS	PIN- PACKAGE
MAXQ3120-FFN	-40°C to +85°C	16kWord Flash	256 Word SRAM	112	3	2	80 MQFP
MAXQ3120-FFN+	-40°C to +85°C	16kWord Flash	256 Word SRAM	112	3	2	80 MQFP

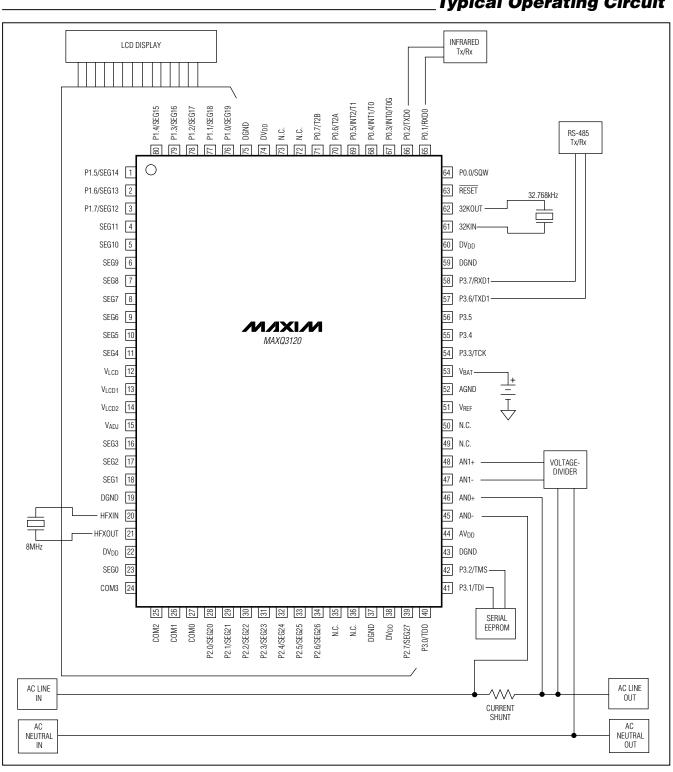
⁺Denotes a Pb-free/RoHS-compliant device.



Pin Configuration

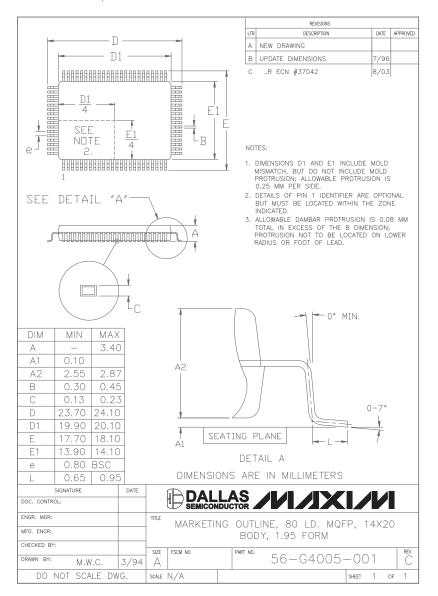


Typical Operating Circuit



Package Information

(The package drawing(s) in this data sheet may not reflect the most current specifications. For the latest package outline information, go to www.maxim-ic.com/DallasPackInfo).



Revision History

33

Rev 0; 7/05: Original release.

Rev 1; 8/05: Added clarification to ADC Resolution condition (No missing codes, with software lowpass filter, see

Appendix A).

Deleted paragraph on Integral Nonlinearity.

Maxim cannot assume responsibility for use of any circuitry other than circuitry entirely embodied in a Maxim product. No circuit patent licenses are implied. Maxim reserves the right to change the circuitry and specifications without notice at any time.

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