

Low Distortion 1.0 GHz Differential Amplifier

AD8350

FEATURES

High Dynamic Range Output IP3: +22 dBm: Re 50 Ω @ 250 MHz Low Noise Figure: 5.9 dB @ 250 MHz Two Gain Versions: AD8350-15 15 dB AD8350-20 20 dB -3 dB Bandwidth: 1.0 GHz Single Supply Operation: +5 V to +10 V Supply Current: 28 mA Input/Output Impedance: 200 Ω Single-Ended or Differential Input Drive 8-Lead SOIC Package

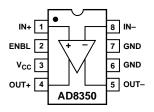
APPLICATIONS Cellular Base Stations Communications Receivers RF/IF Gain Block Differential A-to-D Driver SAW Filter Interface Single-Ended to Differential Conversion High Performance Video High Speed Data Transmission

PRODUCT DESCRIPTION

The AD8350 series are high performance fully-differential amplifiers useful in RF and IF circuits up to 1000 MHz. The amplifier has excellent noise figure of 5.9 dB at 250 MHz. It offers a high output third order intercept (OIP3) of +22 dBm at 250 MHz. Gain versions of 15 dB and 20 dB are offered.

The AD8350 is designed to meet the demanding performance requirements of communications transceiver applications. It enables a high dynamic range differential signal chain, with exceptional linearity and increased common-mode rejection. The device can be used as a general purpose gain block, an A-to-D driver, and high speed data interface driver, among other functions. The AD8350 input can also be used as a singleended-to-differential converter.

FUNCTIONAL BLOCK DIAGRAMS 8-Lead SOIC Package (with Enable)



The amplifier can be operated down to +5 V with an OIP3 of +22 dBm at 250 MHz and slightly reduced distortion performance. The wide bandwidth, high dynamic range and temperature stability make this product ideal for the various RF and IF frequencies required in cellular, CATV, broadband, instrumentation and other applications.

The AD8350 is offered in an 8-lead single SOIC package. It operates from +5 V and +10 V power supplies, drawing 28 mA typical. The AD8350 offers a power enable function for powersensitive applications. The AD8350 is fabricated using Analog Devices' proprietary high speed complementary bipolar process. The device is available in the industrial (-40° C to $+85^{\circ}$ C) temperature range.

REV.0

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AD0220-12-	-3PEUIFIUALIUNS	to differential inputs and differential outputs unless noted.)

Parameter	Conditions	Min	Тур	Max	Units
DYNAMIC PERFORMANCE					
–3 dB Bandwidth	$V_{S} = +5 V, V_{OUT} = 1 V p - p$		0.9		GHz
	$V_{\rm S} = +10$ V, $V_{\rm OUT} = 1$ V p-p		1.1		GHz
Bandwidth for 0.1 dB Flatness	$V_{\rm S} = +5$ V, $V_{\rm OUT} = 1$ V p-p		270		MHz
	$V_{\rm S} = +10 \text{ V}, V_{\rm OUT} = 1 \text{ V} \text{ p-p}$		270		MHz
Slew Rate	$V_{OUT} = 1 V p p$		2000		V/µs
Settling Time	$0.1\%, V_{OUT} = 1 V p p$		10		ns
Gain (S21) ¹	$V_{\rm S} = +5$ V, f = 50 MHz	14	15	16	dB
Gain Supply Sensitivity	$V_{\rm S} = +5$ V to +10 V, f = 50 MHz		0.003	10	dB/V
Gain Temperature Sensitivity	T_{MIN} to T_{MAX}		-0.002		dB/°C
Isolation (S12) ¹	f = 50 MHz		-18		dB/C
	1 – 50 WH 12		-10		uD
NOISE/HARMONIC PERFORMANCE					
50 MHz Signal					
Second Harmonic	$V_{S} = +5 V, V_{OUT} = 1 V p - p$		-66		dBc
	$V_{S} = +10 V, V_{OUT} = 1 V p - p$		-67		dBc
Third Harmonic	$V_{S} = +5 V, V_{OUT} = 1 V p - p$		-65		dBc
	$V_{\rm S} = +10 \text{ V}, V_{\rm OUT} = 1 \text{ V} \text{ p-p}$		-70		dBc
Output Second Order Intercept ²	$V_{\rm S} = +5 \text{ V}$		52		dBm
1 1	$V_{S}^{S} = +10 \text{ V}$		52		dBm
Output Third Order Intercept ²	$V_{S}^{J} = +5 V$		22		dBm
	$V_{S} = +10 \text{ V}$		23		dBm
250 MHz Signal			20		ubiii
Second Harmonic	$V_{\rm S} = +5$ V, $V_{\rm OUT} = 1$ V p-p		-48		dBc
	$V_{\rm S} = +10 \text{ V}, V_{\rm OUT} = 1 \text{ V} \text{ p-p}$		-49		dBc
Third Harmonic	$V_{\rm S} = +5$ V, $V_{\rm OUT} = 1$ V p-p		-52		dBc
Third Harmonie	$V_{S} = +10 V, V_{OUT} = 1 V p p$ $V_{S} = +10 V, V_{OUT} = 1 V p - p$		-61		dBc
Output Second Order Intercept ²	$V_{S} = +10 V_{1}, V_{OOT} = 1 V_{P} P$ $V_{S} = +5 V$		33		dBm
Output Second Order Intercept	$V_{\rm S} = +3$ V $V_{\rm S} = +10$ V		34		dBm
Output Third Order Intercept ²	$V_S = +10$ V $V_S = +5$ V		18		dBm
Output Third Order Intercept	$V_{S} = +3 V$ $V_{S} = +10 V$		22		dBm
1 dB Compression Point (RTI) ²	$V_S = +10 V$ $V_S = +5 V$		2		dBm
1 dB Compression Fount (R11)			2 5		dBm
Voltage Moice (DTI)	$V_{\rm S} = +10 {\rm V}_{\rm S}$				nV/\sqrt{Hz}
Voltage Noise (RTI)	f = 150 MHz		1.7 6.8		dB
Noise Figure	f = 150 MHz		0.0		uр
INPUT/OUTPUT CHARACTERISTICS					
Differential Offset Voltage (RTI)	V _{OUT+} - V _{OUT-}		± 1		mV
Differential Offset Drift	T_{MIN} to T_{MAX}		0.02		mV/°C
Input Bias Current			15		μΑ
Input Resistance	Real		200		Ω
Input Capacitance			2		pF
CMRR	f = 50 MHz		-67		dB
Output Resistance	Real		200		Ω
Output Capacitance			2		pF
					1
POWER SUPPLY				11.0	
Operating Range		+4	00	+11.0	V ,
Quiescent Current	Powered Up, $V_S = +5 V$	25	28	32	mA
	Powered Down, $V_S = +5 V$	3	3.8	5.5	mA
	Powered Up, $V_S = +10 V$	27	30	34	mA
	Powered Down, $V_S = +10 V$	3	4	6.5	mA
Power-Up/Down Switching			15		ns
Power Supply Rejection Ratio	$f = 50 \text{ MHz}, V_S \Delta = 1 \text{ V } p-p$		-58		dB

NOTES ¹See Tables I–IV for complete list of S-Parameters. 2 Re: 50 Ω .

Specifications subject to change without notice.

$(@ +25^{\circ}C, V_{S} = +5 V, G = 20 \text{ dB}, \text{ unless otherwise noted}. All specifications refer to differential inputs and differential outputs unless noted.} \\$

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Parameter	Conditions	Min	Тур	Max	Units
DYNAMIC PERFORMANCE					
–3 dB Bandwidth	$V_{\rm S} = +5$ V, $V_{\rm OUT} = 1$ V p-p		0.7		GHz
	$V_{\rm S} = +10$ V, $V_{\rm OUT} = 1$ V p-p		0.9		GHz
Bandwidth for 0.1 dB Flatness	$V_{\rm S} = +5$ V, $V_{\rm OUT} = 1$ V p-p		230		MHz
	$V_{\rm S} = +10 \text{ V}, V_{\rm OUT} = 1 \text{ V} \text{ p-p}$		200		MHz
Slew Rate	$V_{OUT} = 1 V p p$		2000		V/µs
Settling Time	$0.1\%, V_{OUT} = 1 V p-p$		15		ns
Gain (S21) ¹	$V_{\rm S} = +5$ V, f = 50 MHz	19	20	21	dB
Gain Supply Sensitivity	$V_{\rm S} = +5$ V to +10 V, f = 50 MHz	10	0.003	~-	dB/V
Gain Temperature Sensitivity	T_{MIN} to T_{MAX}		-0.002		dB/°C
Isolation (S12) ¹	f = 50 MHz		-22		dB/ C
			~~		uD
NOISE / HARMONIC PERFORMANCE					
50 MHz Signal					
Second Harmonic	$V_{S} = +5 V, V_{OUT} = 1 V p - p$		-65		dBc
	$V_{S} = +10 \text{ V}, V_{OUT} = 1 \text{ V} \text{ p-p}$		-66		dBc
Third Harmonic	$V_{S} = +5 V, V_{OUT} = 1 V p - p$		-66		dBc
	$V_{S} = +10 \text{ V}, V_{OUT} = 1 \text{ V} \text{ p-p}$		-70		dBc
Output Second Order Intercept ²	$V_{\rm S} = +5 \text{ V}$		50		dBm
	$V_{\rm S} = +10 {\rm V}$		50		dBm
Output Third Order Intercept ²	$V_{\rm S} = +5 \text{ V}$		22		dBm
	$V_{\rm S} = +10 {\rm V}$		23		dBm
250 MHz Signal					
Second Harmonic	$V_{\rm S}$ = +5 V, $V_{\rm OUT}$ = 1 V p-p		-45		dBc
	$V_{\rm S} = +10$ V, $V_{\rm OUT} = 1$ V p-p		-46		dBc
Third Harmonic	$V_{\rm S} = +5$ V, $V_{\rm OUT} = 1$ V p-p		-55		dBc
	$V_{\rm S} = +10$ V, $V_{\rm OUT} = 1$ V p-p		-60		dBc
Output Second Order Intercept ²	$V_{\rm S} = +5 \text{ V}$		31		dBm
	$V_{\rm S} = +10 {\rm V}$		32		dBm
Output Third Order Intercept ²	$V_{\rm S} = +5 \text{ V}$		18		dBm
	$V_{\rm S} = +10 {\rm V}$		22		dBm
1 dB Compression Point (RTI) ²	$V_{\rm S} = +5 \text{ V}$		-2.6		dBm
	$V_{\rm S} = +10 {\rm V}$		1.8		dBm
Voltage Noise (RTI)	f = 150 MHz		1.7		nV/\sqrt{H}
Noise Figure	f = 150 MHz		5.6		dB
INPUT/OUTPUT CHARACTERISTICS					
Differential Offset Voltage (RTI)	V _{OUT+} – V _{OUT-}		± 1		mV
Differential Offset Drift	T_{MIN} to T_{MAX}		0.02		mV/°C
Input Bias Current	I MIN TO I MAX		15		
Input Blas Current Input Resistance	Real		200		μΑ Ω
Input Capacitance	iteai		200		pF
CMRR	f = 50 MHz		ر -52		dB
Output Resistance	Real		-32 200		ub Ω
	real		200		
Output Capacitance			2		pF
POWER SUPPLY					
Operating Range		+4		+11.0	V
Quiescent Current	Powered Up, $V_S = +5 V$	25	28	32	mA
	Powered Down, $V_S = +5 V$	3	3.8	5.5	mA
	Powered Up, $V_S = +10 V$	27	30	34	mA
	· · · · · · · · · · · · · · · · · · ·	0	4	6.5	mA
	Powered Down, $V_S = +10 V$	3	4	0.0	шл
Power-Up/Down Switching	Powered Down, $V_{\rm S} = +10$ V	3	4 15	0.5	ns
Power-Up/Down Switching Power Supply Rejection Ratio	Powered Down, $V_S = +10 V$ f = 50 MHz, $V_S \Delta = 1 V p-p$	3		0.0	

NOTES ¹See Tables I–IV for complete list of S-Parameters. ²Re: 50 Ω.

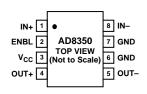
Specifications subject to change without notice.

ABSOLUTE MAXIMUM RATINGS*

Supply Voltage, V _S +11 V
Input Power Differential+8 dBm
Internal Power Dissipation
θ _{JA}
Maximum Junction Temperature+125°C
Operating Temperature Range40°C to +85°C
Storage Temperature Range65°C to +150°C
Lead Temperature Range (Soldering 60 sec) +300°C

*Stresses above those listed under Absolute Maximum Ratings may cause permanent damage to the device. This is a stress rating only; functional operation of the device at these or any other conditions above those indicated in the operational section of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may effect device reliability.

PIN CONFIGURATION



PIN FUNCTION DESCRIPTIONS

Pin	Function	Description
1, 8	IN+, IN-	Differential Inputs. IN+ and IN– should be ac-coupled (pins have a dc bias of midsupply). Differential input impedance is 200 Ω .
2	ENBL	Power-up Pin. A high level (5 V) en- ables the device; a low level (0 V) puts device in sleep mode.
3	V _{CC}	Positive Supply Voltage. +5 V to +10 V.
4, 5	OUT+, OUT-	Differential Outputs. OUT+ and OUT- should be ac-coupled (pins have a dc bias of midsupply). Differential input impedance is 200Ω .
6, 7	GND	Common External Ground Reference.

ORDERING GUIDE

Model	Temperature Range	Package Description	Package Option
AD8350AR15	-40°C to +85°C	8-Lead SOIC	SO-8
AD8350AR15-REEL ¹	-40°C to +85°C	8-Lead SOIC	SO-8
AD8350AR15-REEL7 ²	-40°C to +85°C	8-Lead SOIC	SO-8
AD8350AR15-EVAL		Evaluation Board (15 dB)	
AD8350AR20	-40°C to +85°C	8-Lead SOIC	SO-8
AD8350AR20-REEL ¹	-40°C to +85°C	8-Lead SOIC	SO-8
AD8350AR20-REEL7 ²	-40°C to +85°C	8-Lead SOIC	SO-8
AD8350AR20-EVAL		Evaluation Board (20 dB)	
NOTES	•		•

NOTES ¹13" Reels of 2500 each.

²7" Reels of 750 each.

CAUTION_

ESD (electrostatic discharge) sensitive device. Electrostatic charges as high as 4000 V readily accumulate on the human body and test equipment and can discharge without detection. Although the AD8350 features proprietary ESD protection circuitry, permanent damage may occur on devices subjected to high energy electrostatic discharges. Therefore, proper ESD precautions are recommended to avoid performance degradation or loss of functionality.



Typical Performance Characteristics-AD8350

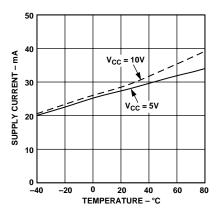


Figure 1. Supply Current vs. Temperature

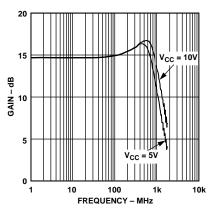


Figure 2. AD8350-15 Gain (S21) vs. Frequency

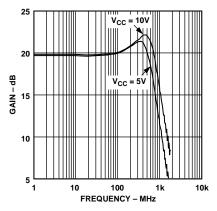


Figure 3. AD8350-20 Gain (S21) vs. Frequency

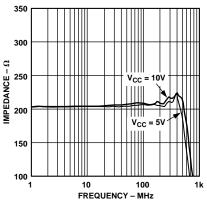


Figure 4. AD8350-15 Input Imped-

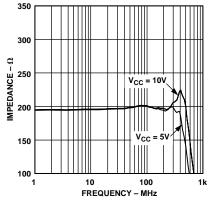


Figure 5. AD8350-20 Input Impedance vs. Frequency

500 400 $\mathsf{IMPEDANCE}-\Omega$ $V_{CC} = 5V$ 300 200 = 10V Vcc 100 0 10 100 1k FREQUENCY - MHz

Figure 7. AD8350-20 Output Impedance vs. Frequency

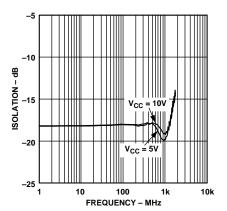


Figure 8. AD8350-15 Isolation (S12) vs. Frequency

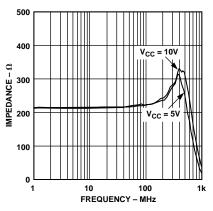


Figure 6. AD8350-15 Output Impedance vs. Frequency

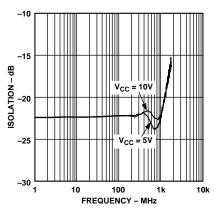


Figure 9. AD8350-20 Isolation (S12) vs. Frequency

ance vs. Frequency

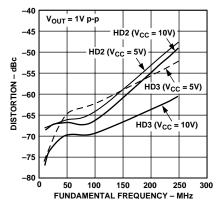


Figure 10. AD8350-15 Harmonic Distortion vs. Frequency

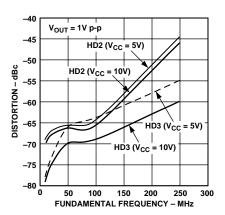


Figure 11. AD8350-20 Harmonic Distortion vs. Frequency

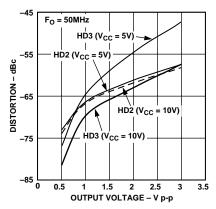


Figure 12. AD8350-15 Harmonic Distortion vs. Differential Output Voltage

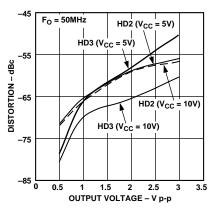


Figure 13. AD8350-20 Harmonic Distortion vs. Differential Output Voltage

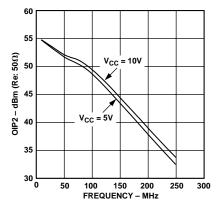


Figure 14. AD8350-15 Output Referred IP2 vs. Frequency

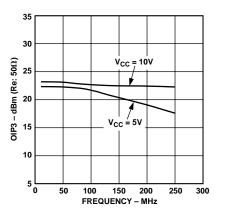


Figure 16. AD8350-15 Output Referred IP3 vs. Frequency

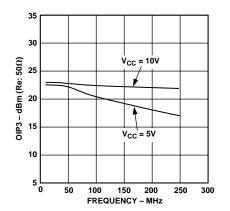


Figure 17. AD8350-20 Output Referred IP3 vs. Frequency

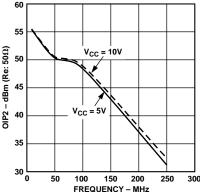


Figure 15. AD8350-20 Output Referred IP2 vs. Frequency

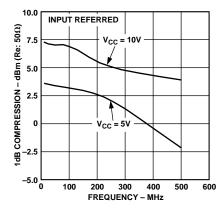
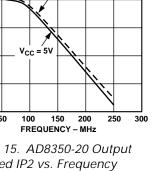


Figure 18. AD8350-15 1 dB Compression vs. Frequency





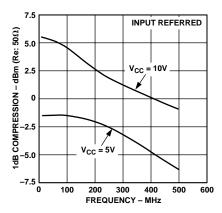


Figure 19. AD8350-20 1 dB Compression vs. Frequency

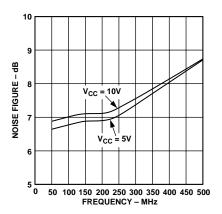


Figure 20. AD8350-15 Noise Figure vs. Frequency

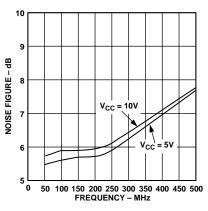


Figure 21. AD8350-20 Noise Figure vs. Frequency

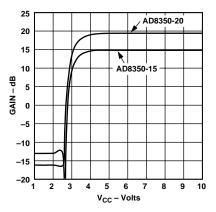


Figure 22. AD8350 Gain (S21) vs. Supply Voltage

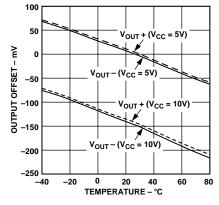


Figure 23. AD8350 Output Offset Voltage vs. Temperature

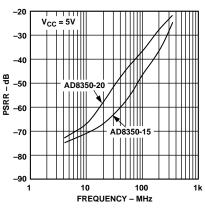


Figure 24. AD8350 PSRR vs. Frequency

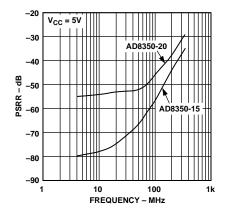


Figure 25. AD8350 CMRR vs. Frequency

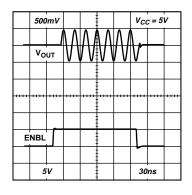


Figure 26. AD8350 Power-Up/Down Response Time

APPLICATIONS Using the AD8350

Figure 27 shows the basic connections for operating the AD8350. A single supply in the range +5 V to +10 V is required. The power supply pin should be decoupled using a 0.1 μ F capacitor. The ENBL pin is tied to the positive supply or to +5 V (when $V_{CC} = +10$ V) for normal operation and should be pulled to ground to put the device in sleep mode. Both the inputs and the outputs have dc bias levels at midsupply and should be ac-coupled.

Also shown, in Figure 27, are the impedance balancing requirements, either resistive or reactive, of the input and output. With an input and output impedance of 200Ω , the AD8350 should be driven by a 200Ω source and loaded by a 200Ω impedance. A reactive match can also be implemented.

Figure 28 shows how the AD8350 can be driven by a singleended source. The unused input should be ac-coupled to ground. When driven single-ended, there will be a slight imbalance in the differential output voltages. This will cause an increase in the second order harmonic distortion (at 50 MHz, with $V_{CC} = +10$ V and $V_{OUT} = 1$ V p-p, -59 dBc was measured for the second harmonic on AD8350-15).

Reactive Matching

In practical applications, the AD8350 will most likely be matched using reactive matching components as shown in Figure 29. Matching components can be calculated using a Smith Chart and the AD8350's S-Parameters (see Tables I and II) along with those of the devices that are driving and loading it. The S-Parameters in Tables I and II assume a differential source and load impedance of 50 Ω . Because the load impedance on the output of the AD8350 affects the input impedance, a simultaneous conjugate match must be performed to correctly match both input and output.

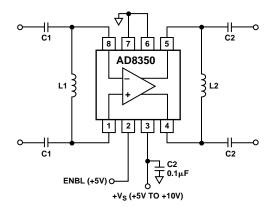


Figure 29. Reactively Matching the Input and Output

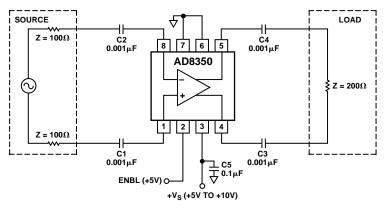


Figure 27. Basic Connections for Differential Drive

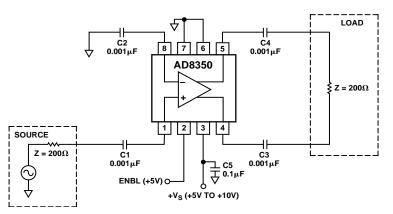


Figure 28. Basic Connections for Single-Ended Drive

Figure 30 shows how the AD8350 input can be matched for a single-ended drive. The unused input is ac-coupled to ground using a low impedance (i.e., high value) capacitance. The S-Parameters for this configuration are shown in Tables III and IV. These values assume a single-ended source impedance of 50 Ω and a differential load impedance of 50 Ω . As in the case of a differential drive, a simultaneous conjugate match must be performed to correctly match both input and output.

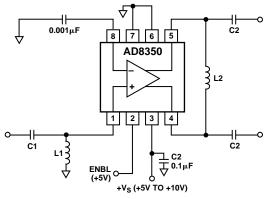


Figure 30. Matching Circuit for Single-Ended Drive

Evaluation Board

Figure 31 shows the schematic of the AD8350 evaluation board as it is shipped from the factory. The board is configured to allow easy evaluation using single-ended 50 Ω test equipment. The input and output transformers have a 4-to-1 impedance ratio and transform the AD8350's 200 Ω input and output impedances to 50 Ω . In this mode, 0 Ω resistors (R1 and R4) are required.

To allow compensation for the insertion loss of the transformers, a calibration path is provided at Test In and Test Out. This consists of two transformers connected back to back.

To drive and load the board differentially, transformers T1 and T2 should be removed and replaced with four 0Ω resistors (0805 size); Resistors R1 and R4 (0Ω) should also be removed. This yields a circuit with a broadband input and output impedance of 200 Ω . To match to impedances other than this, matching components (0805 size) can be placed on pads C1, C2, C3, C4, L1 and L2.

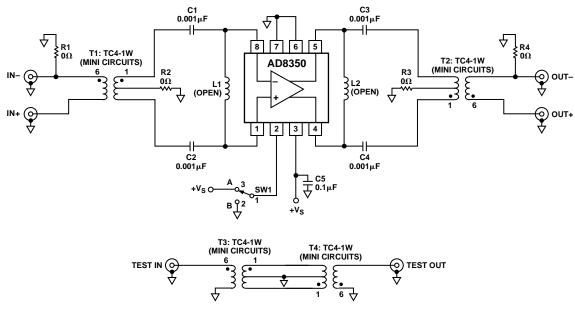


Figure 31. AD8350 Evaluation Board

Frequency (MHz)	S11	S12	S21	S22
50	0.791 ∠ -3°	$0.068 \angle 177^{\circ}$	$2.73 \angle -3^{\circ}$	$0.795 \angle -2^{\circ}$
100	0.787∠–6°	0.071∠174°	$2.79 \angle -7^{\circ}$	0.794 ∠ -5°
150	0.778∠-9°	0.070 ∠ 172°	2.91∠-11°	0.787 ∠ -7°
200	0.766∠-13°	0.072 ∠ 168°	3.06 ∠ −16°	0.779 ∠ -10°
250	$0.749 \angle -17^{\circ}$	$0.074 \angle 165^{\circ}$	$3.24 \angle -21^{\circ}$	$0.768 \angle -12^{\circ}$

Table I. Typical S Parameters AD8350-15: V_{CC} = 5 V, Differential Input Signal. Z_{SOURCE}(diff) = 50 Ω , Z_{LOAD}(diff) = 50 Ω

Table II. Typical S Parameters AD8350-20: V_{CC} = 5 V, Differential Input Signal. Z_{SOURCE}(diff) = 50 Ω , Z_{LOAD}(diff) = 50 Ω

Frequency (MHz)	S11	S12	S21	S22
50	$0.810 \angle -4^{\circ}$	0.046∠176°	$4.82 \angle -2.5^{\circ}$	$0.822 \angle -3^{\circ}$
100	0.795 ∠ -8°	0.043∠173°	$4.99 \angle -6.16^{\circ}$	0.809∠-5°
150	0.790 ∠ -12°	0.045 ∠ 169°	$5.30 \angle -9.82^{\circ}$	0.807∠-8°
200	0.776∠-17°	0.046 ∠ 165°	5.71∠-14.89°	0.795∠-10°
250	$0.757 \angle -22^{\circ}$	$0.048 \angle 162^{\circ}$	$6.25 \angle -21.29^{\circ}$	0.783∠-13°

Table III. Typical S Parameters AD8350-15: V_{CC} = 5 V, Single-Ended Input Signal. Z_{SOURCE}(diff) = 50 Ω , Z_{LOAD}(diff) = 50 Ω

Frequency (MHz)	S11	S12	S21	S22
50	0.718∠-6°	0.068∠177°	$2.62 \angle -4^{\circ}$	0.798∠-3°
100	0.701 ∠ -12°	0.066 ∠ 173°	$2.66 \angle -10^{\circ}$	0.794 ∠ -6°
150	0.683∠-19°	0.067∠167°	$2.76 \angle -15^{\circ}$	0.789 ∠ -10°
200	0.657∠–24°	0.069∠163°	$2.86 \angle -22^{\circ}$	0.776 ∠ -13°
250	$0.625 \angle -31^{\circ}$	$0.070 \angle 159^{\circ}$	$2.98 \angle -28^{\circ}$	0.763 ∠ -16°

Table IV. Typical S Parameters AD8350-20: V_{CC} = 5 V, Single-Ended Input Signal. $Z_{SOURCE}(diff)$ = 50 Ω , $Z_{LOAD}(diff)$ = 50 Ω

Frequency (MHz)	S11	S12	S21	S22
50	0.747∠-7°	0.040 ∠ 175°	4.71∠-4°	0.814∠-3°
100	0.739 ∠ -14°	0.042 ∠ 170°	4.82 ∠ −9°	0.813∠-6°
150	0.728∠-21°	0.044 ∠ 166°	5.08∠-15°	0.804 ∠ -10°
200	0.698 ∠ -29°	0.045∠161°	5.37 ∠ –22°	0.792 ∠ -13°
250	$0.659 \angle -37^{\circ}$	$0.048 \angle 156^{\circ}$	$5.76 \angle -30^{\circ}$	$0.774 \angle -16^{\circ}$

C3577-8-4/99

OUTLINE DIMENSIONS

Dimensions shown in inches and (mm).

8-Lead Plastic SOIC (SO-8)

